



FEATURES

- Standard quarter-brick package/pinout
- Outputs from 1.5 to 48V up to 125W
- Low profile 0.42" height
- 24 and 48Vdc nominal inputs
- Fully isolated, 2250Vdc (BASIC) insulation
- Designed for RoHS compliance
- Output overvoltage/short-circuit protected
- On/Off control, trim and sense functions
- High efficiency to 92%
- Protected against temp. and voltage limits
- Designed to meet UL/IEC/EN60950-1 safety approvals

ORDERING GUIDE SUMMARY

Model	V _{OUT} Range	I _{OUT} Range	V _{IN} Range	Efficiency
All Models	1.2V to 48V	2.5A to 40A	18-36V or 36-75V	Up to 92%, model dependent

INPUT CHARACTERISTICS

Parameter	Typ. @ 25°C, full load	Notes
Voltage Range	18-36 or 36-75 Volts	24V or 48V nominal
Current, full power	Up to 5.6 Amps	Model dependent
Isolation	2kVdc to 2250V	Model dependent
Remote On/Off Control	Switch or FET control	Positive or negative logic

OUTPUT CHARACTERISTICS

Parameter	Typ. @ 25°C, full load	Notes
Voltage	1.5 to 48 Volts ±10%	Trimable
Current	2.5 to 40 Amps fullscale	No minimum load
Accuracy	Down to 1% of V _{NOM}	Most models
Ripple & Noise (to 20MHz)	Down to 35mVp-p	Model dependent
Line and Load Regulation	Down to ±0.125%/±0.25%	Model dependent
Overcurrent Protection	150% of I _{OUT} max.	With hiccup auto-restart
Overtemperature Protection	+125°C	
Efficiency (minimum)	See Performance Specifications	

GENERAL SPECIFICATIONS

Parameter	Typ. @ 25°C, full load	Notes
Dynamic Load Response	Down to 50µsec	Model dependent
Operating Temperature Range	-40 to +110°C	With baseplate, see derating curve
Safety	UL/IEC/EN 60950-1	and CSA C22.2-No.234

MECHANICAL CHARACTERISTICS

With baseplate	1.45 x 2.30 x 0.5 inches (36.83 x 58.42 x 12.7 mm)
Without baseplate	1.45 x 2.30 x 0.42 inches (36.83 x 58.42 x 10.67 mm)

See Performance Specifications, page 2

PRODUCT OVERVIEW

For efficient, fully isolated DC power in the smallest space, Murata Power Solutions' UVQ series quarter bricks offer output voltages from 1.5 to 48 Volts with currents up to 40 Amps. UVQs operate over a wide temperature range (up to +70°C at 200 lfm airflow) at full-rated power. The optional mounting baseplate extends this to all practical temperature ranges at full power.

UVQ's achieve these impressive specifications while delivering excellent electrical performance. Overall noise is 35mVp-p (3.3V models) with fast step response (down to 50µsec). These converters

offer high stability even with no load and tight output regulation. The unit is fully protected against input over and undervoltage, output overcurrent and short circuit. An on-board temperature sensor shuts down the converter if thermal limits are reached. Protection uses the "hiccup" (auto restart) method.

A convenient remote On/Off control input operates by external digital logic, relay or transistor input. To compensate for longer wiring and to retain output voltage accuracy at the load, UVQ's include a Sense input to dynamically correct for

ohmic losses. A trim input may be connected to a user's adjustment potentiometer or trim resistors for output voltage calibration closer than the standard accuracy.

UVQ's include industry-standard safety certifications and BASIC I/O insulation provides 2250 Volt input/output isolation. Radiation emission testing is performed to widely-accepted EMC standards. The UVQ's may be considered as higher performance replacements for some MPS USQ models.



Performance Specifications and Ordering Guide ^①

ORDERING GUIDE														
Root Models	Output						Input				Efficiency		Package (Case, Pinout)	
	V _{OUT} (Volts)	I _{OUT} (Amps)	Power (Watts)	R/N (mVp-p) ^②		Regulation (Max.) ^③		V _{IN} Nom. (Volts)	Range (Volts)	I _{IN} , No Load (mA)	I _{IN} , Full Load (Amps)	Min.		Typ.
				Typ.	Max.	Line	Load							
UVQ-1.5/40-D24P-C	1.5	40	60	30	60	±0.075%	±0.05%	24	18-36	80	2.84	86.5%	88%	C59, P32
UVQ-2.5/35-D24P-C	2.5	35	87.5	35	60	±0.05%	±0.05%	24	18-36	100	4.14	86%	88%	
UVQ-2.5/40-D48N-C		40	100		60	±0.05%	±0.05%	48	36-75	100	2.37	87%	88%	
UVQ-3.3/30-D24P-C	3.3	30	99		65	±0.1%	±0.25%	24	18-36	180	4.58	88.5%	90%	
UVQ-3.3/35-D48N-C		35	115.5	40	±0.05%	±0.25%	48	36-75	130	2.7	87%	89%		
UVQ-5/20-D24P-C	5	20	100	30	50	±0.05%	±0.05%	24	18-36	180	4.53	91%	92%	
UVQ-5/20-D48N-C				20	25	±0.05%	±0.05%	48	36-75	80	2.31	88.5%	90%	
UVQ-12/8-D24P-C	12	8	96	95	130	±0.1%	±0.1%	24	18-36	90	4.4	89%	91%	
UVQ-12/10-D48N-C		10	120	110	160	±0.075%	±0.05%	48	36-75	60	2.78	88.5%	90%	
UVQ-15/7-D24P-C	15	7	105	85	150	±0.05%	±0.05%	24	18-36	103	4.85	88.5%	90.3%	
UVQ-15/7-D48N-C				120	150	±0.05%	±0.02%	48	36-75	60	2.39	90%	91.5%	
UVQ-18/5.6-D24P-C	18	5.6	100.8	125	185	±0.05%	±0.075%	24	18-36	140	4.69	88%	89.5%	
UVQ-18/6-D48N-C		6		125	185	±0.05%	±0.075%	48	36-75	80	2.5	88.3%	90%	
UVQ-24/4.5-D24P-C	24	4.5	108	60	100	±0.075%	±0.15%	24	18-36	45	5.03	88%	89.5%	
UVQ-24/4.5-D48N-C				75	130	±0.075%	±0.25%	48	36-75	45	2.49	89%	90.5%	
UVQ-48/2.5-D24P-C	48	2.5	120	100	200	±0.1%	±0.2%	24	18-36	45	4.4	89%	91%	
UVQ-48/2.5-D48N-C				250	375	±0.175%	±0.2%	48	36-75	30	2.71	91%	92.3%	

- ① These are partial model numbers. Please refer to the full model number structure for complete ordering part numbers.
- ② Min. I_{OUT} = 3 Amps.
- ③ All specifications are at nominal line voltage and full load, +25°C unless otherwise noted. See detailed specifications.

Output capacitors are 1uF ceramic || 10 uF electrolytic. Input cap is 22 uF, low ESR, except UVQ-24/4.5 is 33uF and UVQ-48/2.5 uses no input cap. I/O caps are necessary for our test equipment and may not be needed for your application.

- ④ I_{OUT} = 14 Amps max. with V_{IN} = 18-19.5 Volts.

UVQ Pin 9 Baseplate Connection

The UVQ series may include an optional installed baseplate for extended thermal management. Various UVQ models (see list below) are also available with an additional pin 9 on special order which connects to the baseplate but is electrically isolated from the rest of the converter. Please refer to the mechanical drawings.

Pin 9 offers a positive method of controlling the electrical potential of the baseplate, independent of the converter. Some baseplate models cannot include pin 9 and in such cases, the baseplate is grounded by the mounting bolts. Or consider adding an external lugged washer with a grounding terminal.

The baseplate may be ordered by adding a “B” to the model number tree and pin 9 will be pre-installed by adding a “9”. The two options are separate. Please refer to the Ordering Guide. Do not order pin 9 without the baseplate. Note that “pin 9” converters may be on limited forecast, requiring minimum order quantities and scheduled deliveries.

Models available with Pin 9:

UVQ-12/10-D48
UVQ-1.5/40-D24

Models which are NOT available with Pin 9:

UVQ-5/20-D24 and -D48
UVQ-3.3/30-D24
UVQ-3.3/35-D48
UVQ-2.5/35-D24
UVQ-2.5/40-D48

Other models which are not listed will be reviewed for future pin 9 accommodation.

PART NUMBER STRUCTURE

U VQ - 3.3 / 20 - D48 N B 9 LX - C

Output Configuration

U = Unipolar/Single

Quarter-Brick Package

Nominal Output Voltage

1.2 to 48 Volts

Maximum Rated Output

Current in Amps

Input Voltage Range:

D24 = 18-36 Volts (24V nominal)

D48 = 36-75 Volts (48V nominal)

RoHS-6 hazardous substance compliant

Alternate Pin Length: (special quantity order)

Blank = Standard pin length

L1 = 0.110 (2.79mm)

L2 = 0.145 (3.68mm)

Baseplate Pin 9, see Mechanical Drawings: (special order)

Blank = No pin 9, standard

9 = Pin 9 installed (see description on pg. 2), optional

Blank = No baseplate, standard

B = Baseplate installed, optional special order

Remote On/Off Control Polarity:

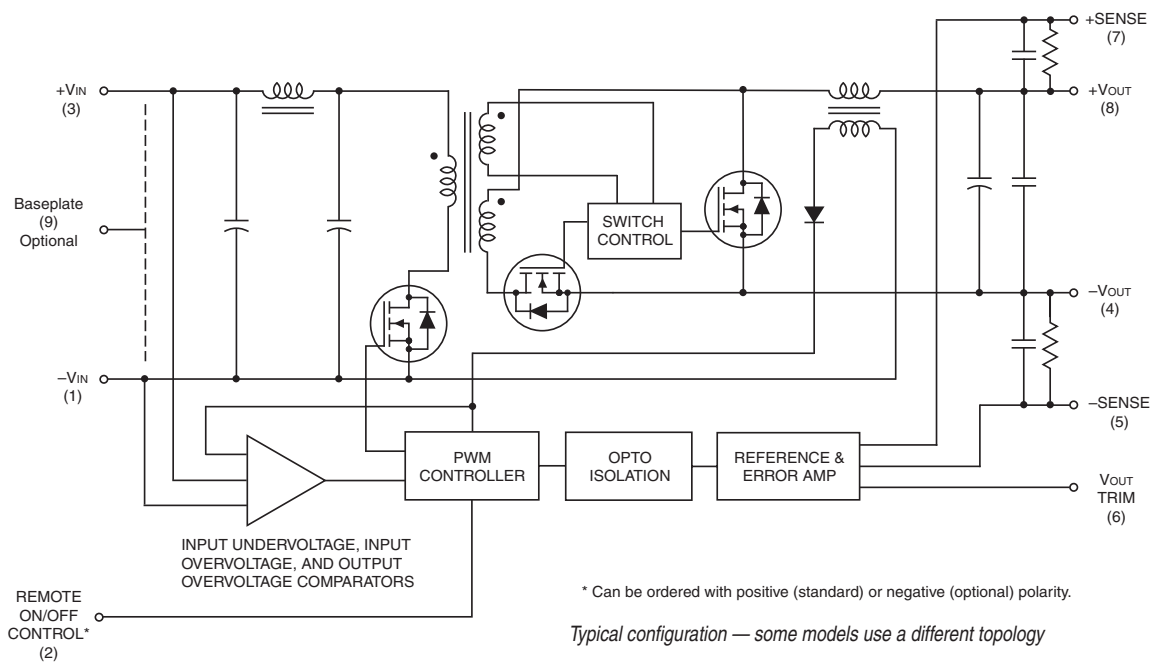
Add "P" for positive polarity

Add "N" for negative polarity

Positive "P" polarity is standard for D24 models and optional special order for D48 models. Negative "N" polarity is standard for D48 models and optional special order for D24 models.

*** Note:**

Some model number combinations may not be available. Contact Murata Power Solutions..



* Can be ordered with positive (standard) or negative (optional) polarity.

Typical configuration — some models use a different topology

Figure 1. Simplified Schematic

Performance/Functional Specifications 24V Models

Typical @ T_A = +25°C under nominal line voltage, nominal output voltage, natural air convection, external caps and full-load conditions, unless noted. (1)

	UVQ-1.5/40-D24	UVQ-2.5/35-D24	UVQ-3.3/30-D24	UVQ-5/20-D24	UVQ-12/8-D24	UVQ-15/7-D24	UVQ-18/5.6-D24	UVQ-24/4.5-D24	UVQ-48/2.5-D24
Input									
Input voltage range	See ordering guide								
Start-up threshold, (V) min.	17	17	17	17	17	17	17	17	17
Undervoltage shutdown, (V) ¹⁴	16			16.25		16	16.25	16	16
Overshoot shutdown (V)	none				39			none	
Reflected (back) ripple current ²	10-50 mA pk-pk, model dependent								
Input Current									
Full load conditions	See ordering guide.								
Inrush transient, (A ² sec)	0.5	0.5	0.05	0.5	0.1	1	1	0.05	0.05
Output short circuit, (mA)	40	50			10	320	50	50	50
No load, mA	80	100	180	160	90	103	140	45	30
Low line (V _{IN} = min.), (Amps)	3.79	5.49	6.04	5.57	5.93	6.52	6.29	6.67	3.60
Standby mode, (Off, UV, OT shutdown)	1-4mA, model dependent								
Internal input filter type	L-C						Pi-type		L-C
Reverse polarity protection	See notes.								
Remote On/Off Control⁵									
Positive logic, "P" suffix (specifications are max)	OFF = Ground pin to +0.8V ON = Open or +5V to +V _{IN} max.								
Negative logic, "N" suffix (specifications are max)	OFF = Open or +5V to +V _{IN} max ON = Ground pin to +0.8V max								
Current	1-8 mA, model dependent								

Performance/Functional Specifications 24V Models

Typical @ T_A = +25°C under nominal line voltage, nominal output voltage, natural air convection, external caps and full-load conditions, unless noted. (1)

	UVQ-1.5/40-D24	UVQ-2.5/35-D24	UVQ-3.3/30-D24	UVQ-5/20-D24	UVQ-12/8-D24	UVQ-15/7-D24	UVQ-18/5.6-D24	UVQ-24/4.5-D24	UVQ-48/2.5-D24
Output									
Voltage output range	See ordering guide.								
Voltage output accuracy (50% load)	±1.5% of V _{NOM}			±1.25% of V _{NOM}			±1% of V _{NOM}		
Adjustment range	–20 to +10% of V _{NOM} .								±10% of V _{NOM} .
Temperature coefficient	±0.02% of V _{OUT} range per °C								
Minimum loading	No minimum load	3 amps		No minimum load					
Remote sense compensation	+10%.								
Ripple/noise	See ordering guide.								
Line/Load regulation	See ordering guide.								
Efficiency	See ordering guide.								
Maximum capacitive loading, Low ESR <0.02Ω max., resistive load, (μF)	10,000			5000	4700			2200	
Current limit inception (98% of V _{OUT} , after warmup), (Amps)	45	44	36	24	10	9.5	7	5.8	3.4
Short circuit protection method	Current limiting, hiccup autorestart. Remove overload for recovery.								
Short circuit current, (Amps)	3.6	3	3	3	1.5	15 mA	3	5	2.8
Short circuit duration	Output may be shorted continuously to ground (no damage).								
Overvoltage protection, (via magnetic feedback)	2.3 Volts	3 Volts max	4 Volts max	6.8 Volts max	14.4 Volts max	18.5 Volts	22 Volts max	29 Volts max	59 Volts max
Isolation Characteristics									
Isolation Voltage									
Input to Output, (Volts min)	2000								
Input to baseplate	1500								
Baseplate to output, (Volts min)	1500			1000	1500				
Isolation resistance	100 MΩ								
Isolation capacitance, (pF)	1500				1000	2000	50		
Isolation safety rating	Basic insulation								

Performance/Functional Specifications 24V Models

Typical @ T_A = +25°C under nominal line voltage, nominal output voltage, natural air convection, external caps and full-load conditions, unless noted. (1)

	UVQ-1.5/40-D24	UVQ-2.5/35-D24	UVQ-3.3/30-D24	UVQ-5/20-D24	UVQ-12/8-D24	UVQ-15/7-D24	UVQ-18/5.6-D24	UVQ-24/4.5-D24	UVQ-48/2.5-D24	
Dynamic characteristics										
Dynamic load response (50-75-50% load step)	100 μSec to ±1% of final value	150 μSec to ±1.5% of final value	150 μSec to ±1.5% of final value	100 μSec to ±1.5% of final value	50 μSec to ±1% of final value	40 μSec to ±1.25% of final value	50 μSec to ±1% of final value	100 μSec to ±1% of final value	100 μSec to ±1% of final value	
Start-up time V _{IN} to V _{OUT} regulated, mSec	90msec	50msec	50msec	200msec	40msec	30msec	30msec	290msec	100msec	
Remote On/Off to V _{OUT} regulated, mSec	90msec	50msec	50msec	200msec	30msec	25msec	35msec	200msec	100msec	
Switching frequency, (KHz)	380 ± 30	500 to 650	600	360	290 ± 30	242	240 ± 25	290 ± 30	250 ± 25	
Environmental										
Calculated MTBF ⁴	TBD									
Operating temperature range: see Derating Curves.	–40 to +85°C (with Derating, see Note 15.)									
Operating temperature, with baseplate, no derating required (°C) ³	–40 to +110					–40 to +115		–40 to +110		
Storage temperature (°C)	–55 to +130								–55 to +125	
Thermal protection/ shutdown	+110 to 125°C, model dependent									
Relative humidity	To +85°C/85%, non-condensing									
Physical										
Outline dimensions	See mechanical specs.									
Baseplate material	Aluminum									
Pin material	Brass alloy									
Pin diameter	0.04/0.062 inches (1.016/1.524 mm)									
Weight	1.55 ounce (44 grams)	1 ounce (28 grams)								
Electromagnetic interference (conducted and radiated) (external filter required)	Designed to meet FCC part 15, class B, EN55022									
Safety	Designed to meet UL/cUL 60950-1, CSA C22.2 No.60950-1, IEC/EN 60950-1									

Performance/Functional Specifications 48V Models

Typical @ T_A = +25°C under nominal line voltage, nominal output voltage, natural air convection, external caps and full-load conditions, unless noted. (1)

	UVQ-2.5/40-D48	UVQ-3.3/35-D48	UVQ-5/20-D48	UVQ-12/10-D48	UVQ-15/7-D48	UVQ-18/6-D48	UVQ-24/4.5-D48	UVQ-48/2.5-D48
Input								
Input voltage range	See ordering guide							
Start-up threshold, min (V)	35		34.5		34	34.5		35
Undervoltage shutdown, (V) ¹⁴	33.5			32			33.5	
Overshoot shutdown (V)	none							
Reflected (back) ripple current	10-50 mA pk-pk, model dependent							
Input Current								
Full load conditions	See ordering guide.							
Inrush transient, (A ² sec)	0.05	0.05	1	1	0.05	1	0.05	0.05
Output short circuit, (mA)	50			10	30	50	250	50
No load, mA	100	130	80	60	30	80	45	30
Low line (V _{IN} = min.), (Amps)	3.15	3.56	3.07	3.72	3.21	3.35	3.30	3.60
Standby mode, (Off, UV, OT shutdown)	1-4mA, model dependent							
Internal input filter type	L-C				Pi-type			L-C
Reverse polarity protection	See notes.							
Remote On/Off Control⁵								
Positive logic, "P" suffix (specifications are max)	OFF = Ground pin to +0.8V ON = Open or +5V to +V _{IN} max							
Negative logic, "N" suffix (specifications are max)	OFF = Open or +5V to +V _{IN} max ON = Ground pin to +0.8V max							
Current	1-8 mA, model dependent							

Performance/Functional Specifications 48V Models

Typical @ T_A = +25°C under nominal line voltage, nominal output voltage, natural air convection, external caps and full-load conditions, unless noted. (1)

	UVQ-2.5/40-D48	UVQ-3.3/35-D48	UVQ-5/20-D48	UVQ-12/10-D48	UVQ-15/7-D48	UVQ-18/6-D48	UVQ-24/4.5-D48	UVQ-48/2.5-D48
Output								
Voltage output range	See ordering guide.							
Voltage output accuracy (50% load)	±1.5% of V _{NOM}			±1.25% of V _{NOM}	±1% of V _{NOM}			
Adjustment range	–20 to +10% of V _{NOM} .							
Temperature coefficient	±0.02% of V _{OUT} range per °C							
Minimum loading	No minimum load	3 Amps	No minimum load	No minimum load				
Remote sense compensation	+10%.							
Ripple/noise	See ordering guide.							
Line/Load regulation	See ordering guide.							
Efficiency	See ordering guide.							
Maximum capacitive loading, Low ESR <0.02Ω max., resistive load, (μF)	10,000			4700	2200		1000	
Current limit inception (98% of V _{OUT} , after warmup), (Amps)	46	48	26	12.5	8.5	7	6.5	3.3
Short circuit protection method	Current limiting, hiccup autorestart. Remove overload for recovery.							
Short circuit current, (Amps)	5		0.1	1.5	3	3	3	3.5
Short circuit duration	Output may be shorted continuously to ground (no damage).							
Overvoltage protection, (via magnetic feedback)	3 Volts max	4 Volts max	6 Volts max	14.4 Volts max	18.5 Volts max	22 Volts max	29 Volts max	55 Volts max
Isolation Characteristics								
Isolation Voltage								
Input to Output, (Volts min)	2250							
Input to baseplate	1500							
Baseplate to output, (Volts min)	1500				1500			
Isolation resistance	100 MΩ							
Isolation capacitance, (pF)	1500			1000	50	50	1500	
Isolation safety rating	Basic insulation							

Performance/Functional Specifications 48V Models

Typical @ T_A = +25°C under nominal line voltage, nominal output voltage, natural air convection, external caps and full-load conditions, unless noted. (1)

	UVQ-2.5/40-D48	UVQ-3.3/35-D48	UVQ-5/20-D48	UVQ-12/10-D48	UVQ-15/7-D48	UVQ-18/6-D48	UVQ-24/4.5-D48	UVQ-48/2.5-D48
Dynamic characteristics								
Dynamic load response (50-75-50% load step)	150 µSec to ±1.5% of final value	150 µSec to ±1.5% of final value	90 µSec to ±2% of final value	50 µSec to ±1% of final value	50 µSec to ±1% of final value	50 µSec to ±1% of final value	100 µSec to ±1% of final value	75 µSec to ±1% of final value
Start-up time V _{IN} to V _{OUT} regulated, mSec	50msec	50msec	50msec	40msec	30msec	30msec	100msec	50msec
Remote On/Off to V _{OUT} regulated, mSec	50msec	50msec	50msec	30msec	30msec	30msec	100msec	50msec
Switching frequency, (KHz)	600	600	450 ± 50	290 ± 30	245 ± 20	240 ± 25	290 ± 30	540 ± 40
Environmental								
Calculated MTBF ⁴	TBD							
Operating temperature range: see Derating Curves.	–40 to +85°C (with Derating, see Note 15.)							
Operating temperature, with baseplate, no derating required (°C) ³	–40 to +110			–40 to +115		–40 to +110	–40 to +110	–40 to +120
Storage temperature (°C)	–55 to +125							
Thermal protection/ shutdown	+110 to 125°C, model dependent							
Relative humidity	To +85°C/85%, non-condensing							
Physical								
Outline dimensions	See mechanical specs.							
Baseplate material	Aluminum							
Pin material	Brass alloy							
Pin diameter	0.04/0.062 inches (1.016/1.524 mm)							
Weight	1 ounce (28 grams)							
Electromagnetic interference (conducted and radiated) (external filter required)	Designed to meet FCC part 15, class B, EN55022							
Safety	Designed to meet UL/cUL 60950-1, CSA C22.2 No.60950-1, IEC/EN 60950-1							

PART NUMBER STRUCTURE

	24V models	48V models
Input Voltage	24V models	48V models
Continuous	0 to +36V	0 to +75V
Transient (100 mS)	+50V	+100V
On/Off Control	-0.3 V min to +13.5V max.	
Input Reverse Polarity Protection	See Fuse section	
Output Overvoltage	V _{OUT} +20% max.	
Output Current (Note 7)	Current-limited. Devices can withstand sustained short circuit without damage.	
Storage Temperature	-55 to +125°C	
Lead Temperature	See soldering guidelines	

Absolute maximums are stress ratings. Exposure of devices to greater than any of these conditions may adversely affect long-term reliability. Proper operation under conditions other than those listed in the Performance/Functional Specifications Table is not implied nor recommended.

- (1) All models are tested and specified with 200 LFM airflow, external 1110µF ceramic/tantalum output capacitors. External input capacitance varies according to model type. All capacitors are low ESR types. These capacitors are necessary to accommodate our test equipment and may not be required to achieve specified performance in your applications. All models are stable and regulate within spec under no-load conditions. General conditions for Specifications are +25°C, V_{IN} = nominal, V_{OUT} = nominal, full load.
- (2) Input Ripple Current is tested and specified over a 5-20MHz bandwidth. Input filtering is C_{IN} = 33µF tantalum, C_{BUS} = 220µF electrolytic, L_{BUS} = 12µH.
- (3) Note that Maximum Power Derating curves indicate an average current at nominal input voltage. At higher temperatures and/or lower airflow, the DC/DC converter will tolerate brief full current outputs if the total RMS current over time does not exceed the Derating curve.
- (4) Mean Time Before Failure is calculated using the Telcordia (Belcore) SR-332 Method 1, Case 3, ground fixed conditions, TPCBOARD = +25°C, full output load, natural air convection.
- (5) The On/Off Control may be driven with external logic or by applying appropriate external voltages which are referenced to Input Common. The On/Off Control Input should use either an open collector/open drain transistor or logic gate which does not exceed +13.5V.
- (6) Short circuit shutdown begins when the output voltage degrades approximately 2% from the selected setting.
- (7) The outputs are not intended to sink appreciable reverse current.
- (8) Output noise may be further reduced by adding an external filter. See I/O Filtering and Noise Reduction.
- (9) All models are fully operational and meet published specifications, including “cold start” at -40°C.
- (10) Regulation specifications describe the deviation as the line input voltage or output load current is varied from a nominal midpoint value to either extreme.
- (11) Overvoltage shutdown on 48V input models is not supplied in order to comply with telecom reliability requirements. These requirements attempt continued operation despite significant input overvoltage.
- (12) Do not exceed maximum power specifications when adjusting the output trim.
- (13) Note that the converter may operate up to +110°C with the baseplate installed. However, thermal self-protection occurs near +110°C, and there is a temperature gradient between the hotspot and the baseplate. Therefore, +100°C is recommended to avoid thermal shutdown.
- (14) The converter is guaranteed to turn off at the UV shutdown voltage.
- (15) At full power, the package temperature of all on-board components must not exceed +128°C.

MECHANICAL SPECIFICATIONS

Removal of Soldered UVQ's from Printed Circuit Boards

Should removal of the UVQ from its soldered connection be needed, thoroughly de-solder the pins using solder wicks or de-soldering tools. At no time should any prying or leverage be used to remove boards that have not been properly de-soldered first.

Input Source Impedance

UVQ converters must be driven from a low ac-impedance input source. The DC/DC's performance and stability can be compromised by the use of highly inductive source impedances. The input circuit shown in Figure 2 is a practical solution that can be used to minimize the effects of inductance in the input traces. For optimum performance, components should be mounted close to the DC/DC converter.

I/O Filtering, Input Ripple Current, and Output Noise

All models in the UVQ Series are tested/specified for input ripple current (also called input reflected ripple current) and output noise using the circuits and layout shown in Figures 2 and 3.

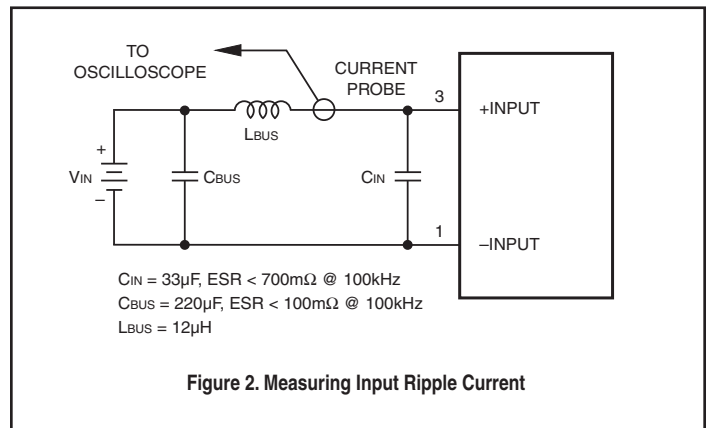
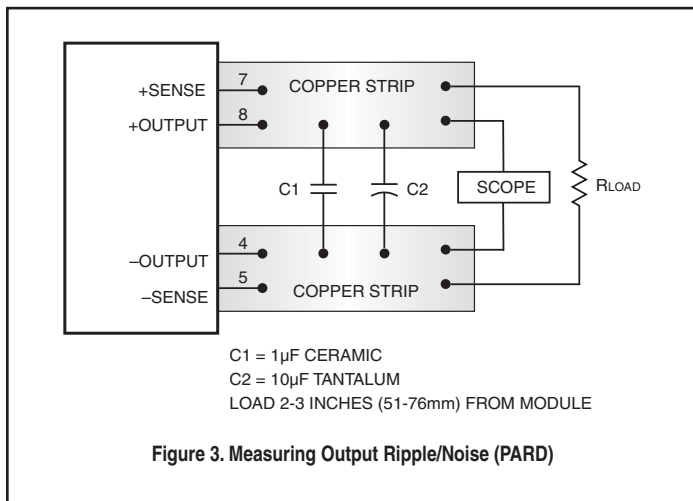


Figure 2. Measuring Input Ripple Current

External input capacitors (C_{IN} in Figure 2) serve primarily as energy-storage elements. They should be selected for bulk capacitance (at appropriate frequencies), low ESR, and high rms-ripple-current ratings. The switching nature of DC/DC converters requires that dc voltage sources have low ac impedance as highly inductive source impedance can affect system stability. In Figure 2, C_{BUS} and L_{BUS} simulate a typical dc voltage bus. Your specific system configuration may necessitate additional considerations.

In critical applications, output ripple/noise (also referred to as periodic and random deviations or PARD) can be reduced below specified limits using filtering techniques, the simplest of which is the installation of additional external output capacitors. Output capacitors function as true filter elements and should be selected for bulk capacitance, low ESR, and appropriate frequency response. In Figure 3, the two copper strips simulate real-world pcb impedances between the power supply and its load. Scope measurements should be made using BNC connectors or the probe ground should be less than ½ inch and soldered directly to the fixture.

All external capacitors should have appropriate voltage ratings and be located as close to the converter as possible. Temperature variations for all relevant parameters should be taken into consideration. OS-CON™ organic semiconductor capacitors (www.sanyo.com) can be especially effective for further reduction of ripple/noise. The most effective combination of external I/O capacitors will be a function of line voltage and source impedance, as well as particular load and layout conditions.



Start-Up Threshold and Undervoltage Shutdown

Under normal start-up conditions, the UVQ Series will not begin to regulate properly until the ramping input voltage exceeds the Start-Up Threshold. Once operating, devices will turn off when the applied voltage drops below the Undervoltage Shutdown point. Devices will remain off as long as the undervoltage condition continues. Units will automatically re-start when the applied voltage is brought back above the Start-Up Threshold. The hysteresis built into this function avoids an indeterminate on/off condition at a single input voltage. See Performance/Functional Specifications table for actual limits.

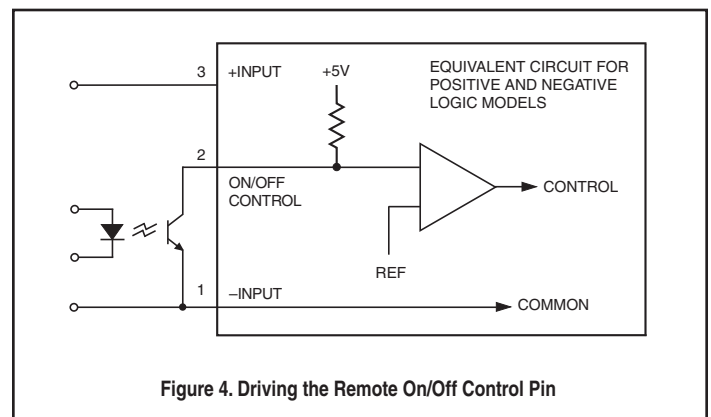
Start-Up Time

The V_{IN} to V_{OUT} Start-Up Time is the interval between the point at which a ramping input voltage crosses the Start-Up Threshold voltage and the point at which the fully loaded output voltage enters and remains within its specified $\pm 1\%$ accuracy band. Actual measured times will vary with input source impedance, external input capacitance, and the slew rate and final value of the input voltage as it appears to the converter. The On/Off to V_{OUT} start-up time assumes that the converter is turned off via the Remote On/Off Control with the nominal input voltage already applied.

On/Off Control

The primary-side, Remote On/Off Control function (pin 2) can be specified to operate with either positive or negative polarity. Positive-polarity devices ("P" suffix) are enabled when pin 2 is left open or is pulled high. Positive-polarity devices are disabled when pin 2 is pulled low (0-0.8V with respect to -Input). Negative-polarity devices are off when pin 2 is high/open and on when pin 2 is pulled low. See Figure 4.

Dynamic control of the remote on/off function is best accomplished with a mechanical relay or an open-collector/open-drain drive circuit (optically isolated if appropriate). The drive circuit should be able to sink appropriate current (see Performance Specifications) when activated and withstand appropriate voltage when deactivated.



Current Limiting

When power demands from the output falls within the current limit inception range for the rated output current, the DC/DC converter will go into a current limiting mode. In this condition the output voltage will decrease proportionately with increases in output current, thereby maintaining a somewhat constant power dissipation. This is commonly referred to as power limiting. Current limit inception is defined as the point where the full-power output voltage falls below the specified tolerance. If the load current being drawn from the converter is significant enough, the unit will go into a short circuit condition. See "Short Circuit Condition."

Short Circuit Condition

When a converter is in current limit mode the output voltages will drop as the output current demand increases. If the output voltage drops too low, the magnetically coupled voltage used to develop primary side voltages will also drop, thereby shutting down the PWM controller. Following a time-out period of about 50 milliseconds, the PWM will restart, causing the output voltages to begin ramping to their appropriate values. If the short-circuit condition persists, another shutdown cycle will be initiated. This on/off cycling is referred to as "hiccup" mode. The hiccup cycling reduces the average output current, thereby preventing internal temperatures from rising to excessive levels. The UVQ is capable of enduring an indefinite short circuit output condition.

Thermal Shutdown

UVQ converters are equipped with thermal-shutdown circuitry. If the internal temperature of the DC/DC converter rises above the designed operating temperature (See Performance Specifications), a precision temperature sensor will power down the unit. When the internal temperature decreases below the threshold of the temperature sensor, the unit will self start.

Output Overvoltage Protection

The output voltage is monitored for an overvoltage condition via magnetic coupling to the primary side. If the output voltage rises to a fault condition, which could be damaging to the load circuitry (see Performance Specifications), the sensing circuitry will power down the PWM controller causing the output voltage to decrease. Following a time-out period the PWM will restart, causing the output voltage to ramp to its appropriate value. If the fault condition persists, and the output voltages again climb to excessive levels, the overvoltage circuitry will initiate another shutdown cycle. This on/off cycling is referred to as "hiccup" mode.

Input Reverse-Polarity Protection

If the input-voltage polarity is accidentally reversed, an internal diode will become forward biased and likely draw excessive current from the power source. If the source is not current limited or the circuit appropriately fused, it could cause permanent damage to the converter.

Input Fusing

Certain applications and/or safety agencies may require the installation of fuses at the inputs of power conversion components. Fuses should also be used if the possibility of a sustained, non-current-limited, input-voltage polarity reversal exists. For MPS UVQ Series DC/DC Converters, slow-blow fuses are recommended with values no greater than twice the maximum input current.

Trimming Output Voltage

UVQ converters have a trim capability (pin 6) that enables users to adjust the output voltage from +10% to -20% (refer to the trim equations). Adjustments to the output voltage can be accomplished with a single fixed resistor as shown in Figures 5 and 6. A single fixed resistor can increase or decrease the output voltage depending on its connection. Resistors should be located close to the converter and have TCR's less than 100ppm/°C to minimize sensitivity to changes in temperature. If the trim function is not used, leave the trim pin open.

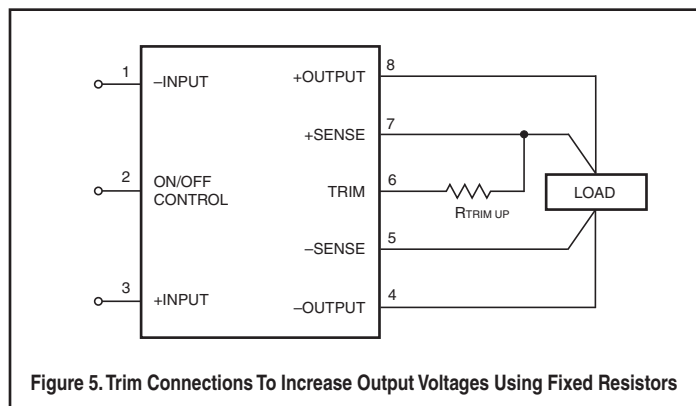


Figure 5. Trim Connections To Increase Output Voltages Using Fixed Resistors

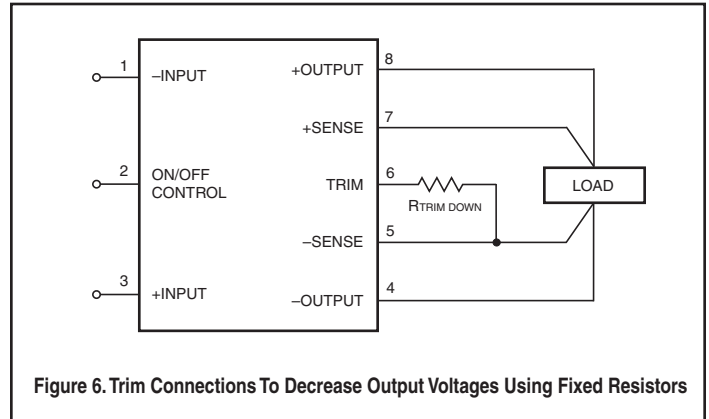


Figure 6. Trim Connections To Decrease Output Voltages Using Fixed Resistors

Standard UVQ's have a "positive trim" where a single resistor connected from the Trim pin (pin 6) to the +Sense (pin 7) will increase the output voltage. A resistor connected from the Trim Pin (pin 6) to the -Sense (pin 5) will decrease the output voltage.

Trim adjustments greater than the specified +10%/-20% can have an adverse affect on the converter's performance and are not recommended. Excessive voltage differences between V_{out} and Sense, in conjunction with trim adjustment of the output voltage, can cause the overvoltage protection circuitry to activate (see Performance Specifications for overvoltage limits).

Temperature/power derating is based on maximum output current and voltage at the converter's output pins. Use of the trim and sense functions can cause output voltages to increase, thereby increasing output power beyond the UVQ's specified rating, or cause output voltages to climb into the output overvoltage region. Therefore:

$$(V_{out \text{ at pins}}) \times (I_{out}) \leq \text{rated output power}$$

The Trim pin (pin 6) is a relatively high impedance node that can be susceptible to noise pickup when connected to long conductors in noisy environments.

Soldering Guidelines

Murata Power Solutions recommends the specifications below when installing these converters. These specifications vary depending on the solder type. Exceeding these specifications may cause damage to the product. Be cautious when there is high atmospheric humidity. We strongly recommend a mild pre-bake (100° C. for 30 minutes). Your production environment may differ; therefore please thoroughly review these guidelines with your process engineers.

Wave Solder Operations for through-hole mounted products (THMT)

For Sn/Ag/Cu based solders:

Maximum Preheat Temperature	115° C.
Maximum Pot Temperature	270° C.
Maximum Solder Dwell Time	7 seconds

For Sn/Pb based solders:

Maximum Preheat Temperature	105° C.
Maximum Pot Temperature	250° C.
Maximum Solder Dwell Time	6 seconds

Trim Equations

Trim Up

Trim Down

UVQ-1.5/40-D24	
$R_{TUP} (k\Omega) = \frac{6.23(V_O - 1.226)}{V_O - 1.5} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{7.64}{1.5 - V_O} - 10.2$
UVQ-2.5/40-D48, UVQ-2.5/35-D24	
$R_{TUP} (k\Omega) = \frac{10(V_O - 1.226)}{V_O - 2.5} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{12.26}{2.5 - V_O} - 10.2$
UVQ-3.3/35-D48	
$R_{TUP} (k\Omega) = \frac{13.3(V_O - 1.226)}{V_O - 3.3} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{16.31}{3.3 - V_O} - 10.2$
UVQ-5/25-D24, UVQ-5/20-D48	
$R_{TUP} (k\Omega) = \frac{20.4(V_O - 1.226)}{V_O - 5} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{25.01}{5 - V_O} - 10.2$
UVQ-12/8-D24, -12/10-D48	
$R_{TUP} (k\Omega) = \frac{49.6(V_O - 1.226)}{V_O - 12} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{60.45}{12 - V_O} - 10.2$
UVQ-15/7-D24, -D48	
$R_{TUP} (k\Omega) = \frac{62.9(V_O - 1.226)}{V_O - 15} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{76.56}{15 - V_O} - 10.2$
UVQ-18/5.6-D24, -18/6-D48	
$R_{TUP} (k\Omega) = \frac{75.5(V_O - 1.226)}{V_O - 18} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{92.9}{18 - V_O} - 10.2$
UVQ-24/4.5-D24, -D48	
$R_{TUP} (k\Omega) = \frac{101(V_O - 1.226)}{V_O - 24} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{124.2}{24 - V_O} - 10.2$
UVQ-48/2.5-D24, -D48	
$R_{TUP} (k\Omega) = \frac{210.75(V_O - 1.226)}{V_O - 48} - 10.2$	$R_{TDOWN} (k\Omega) = \frac{250}{48 - V_O} - 10.2$

Note: Higher output 24V and 48V converters require larger, low-tempco, precision trim resistors. An alternative is a low-TC multi-turn potentiometer (20kΩ typical) connected between +V_{OUT} and -V_{OUT} with the wiper to the Trim pin.

FEATURES AND OPTIONS

Remote Sense

Note: The Sense and V_{OUT} lines are internally connected through low-value resistors. Nevertheless, if the sense function is not used for remote regulation the user must connect the +Sense to +V_{OUT} and -Sense to -V_{OUT} at the DC/DC converter pins.

UVQ series converters employ a sense feature to provide point of use regulation, thereby overcoming moderate IR drops in pcb conductors or cabling. The remote sense lines carry very little current and therefore require minimal cross-sectional-area conductors. The sense lines, which are capacitively coupled to their respective output lines, are used by the feedback control-loop to regulate the output. As such, they are not low impedance points and must be treated with care in layouts and cabling. Sense lines on a pcb should be run adjacent to dc signals, preferably ground. In cables and discrete wiring applications, twisted pair or other techniques should be implemented.

UVQ series converters will compensate for drops between the output voltage at the DC/DC and the sense voltage at the DC/DC provided that:

$$[V_{OUT(+)} - V_{OUT(-)}] - [Sense(+) - Sense(-)] \leq 10\% V_{OUT}$$

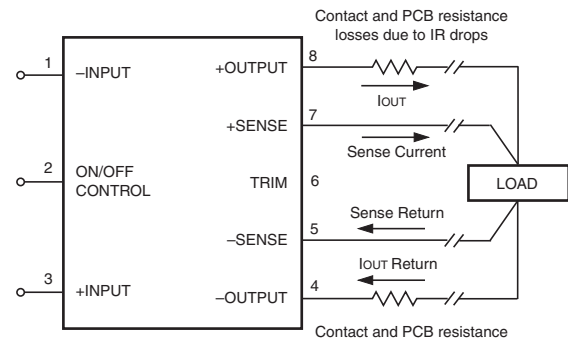
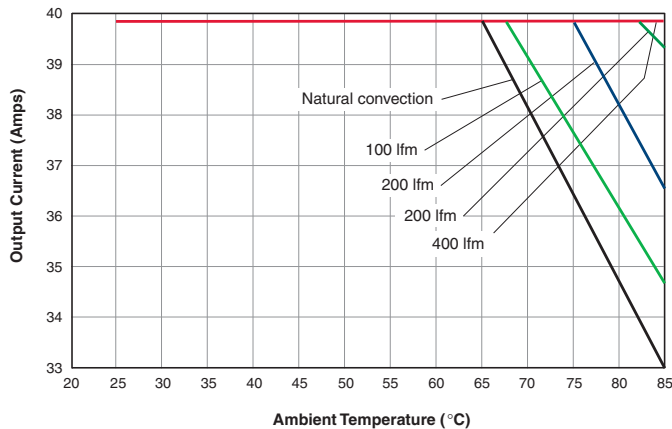


Figure 8. Remote Sense Circuit Configuration

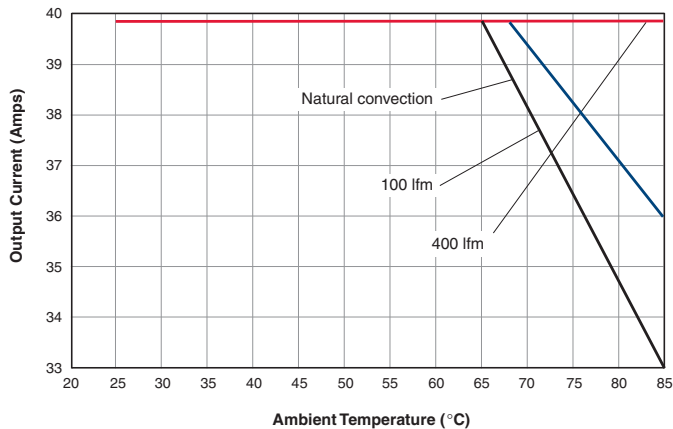
Output overvoltage protection is monitored at the output voltage pin, not the Sense pin. Therefore, excessive voltage differences between V_{OUT} and Sense in conjunction with trim adjustment of the output voltage can cause the overvoltage protection circuitry to activate (see Performance Specifications for overvoltage limits). Power derating is based on maximum output current and voltage at the converter's output pins. Use of trim and sense functions can cause output voltages to increase, thereby increasing output power beyond the converter's specified rating, or cause output voltages to climb into the output overvoltage region. Therefore, the designer must ensure:

$$(V_{OUT \text{ at pins}}) \times (I_{OUT}) \leq \text{rated output power}$$

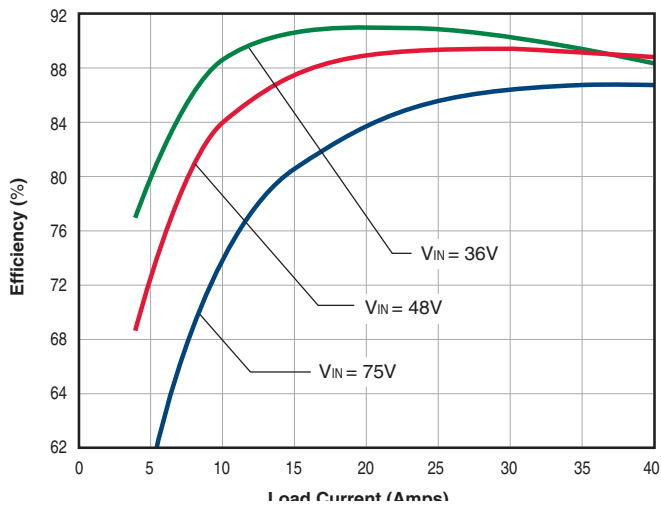
UVQ-1.5/40-D24N: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 24V$, transverse air flow)



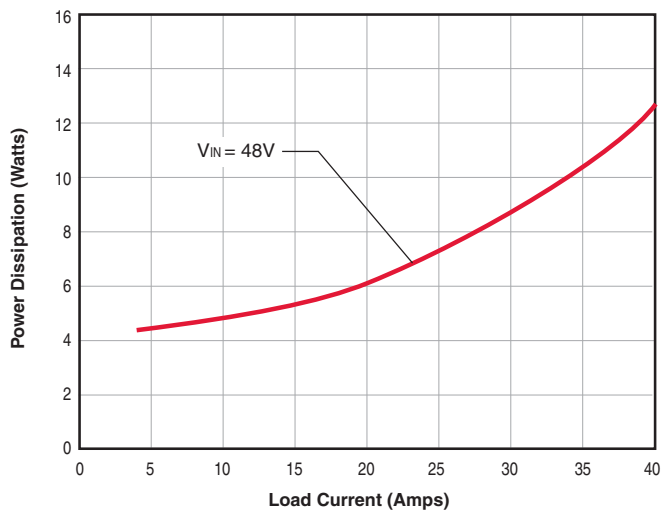
UVQ-1.5/40-D24N: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 24V$, transverse air flow)



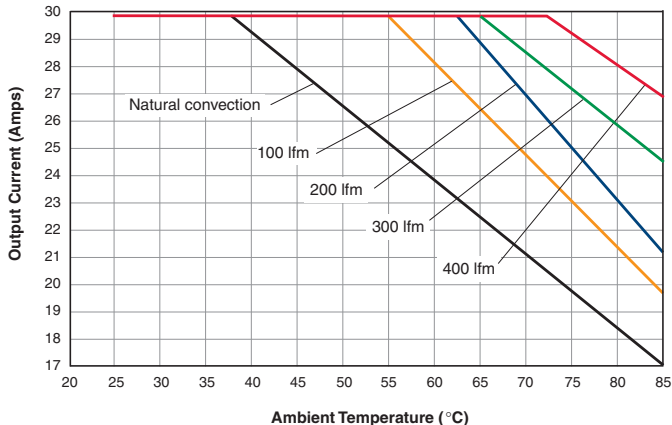
UVQ-2.5/40-D48N
Efficiency vs. Line Voltage and Load Current @ 25°C



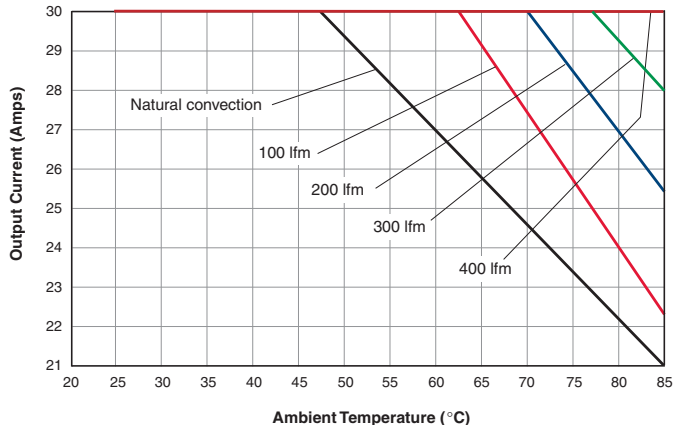
UVQ-2.5/40-D48
Power Dissipation vs. Load Current @ 25°C



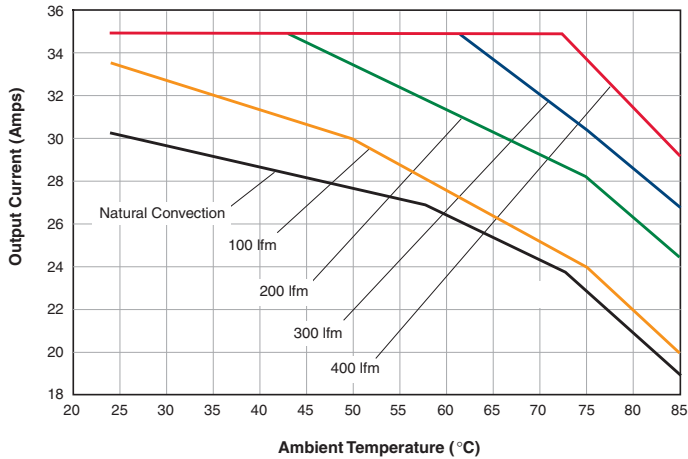
UVQ-3.3/30-D24N: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 24V$, transverse air flow at sea level)



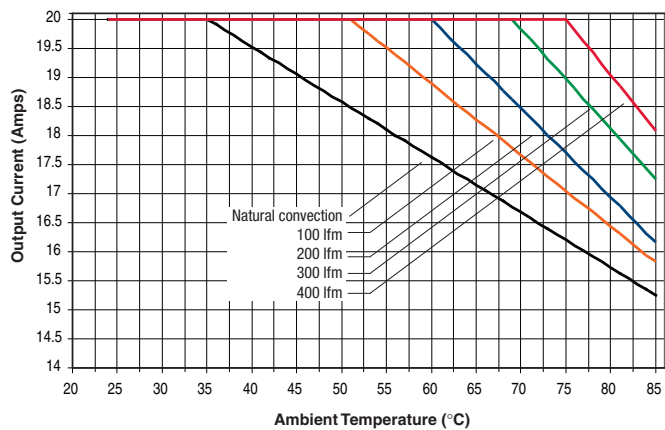
UVQ-3.3/30-D24N: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 24V$, transverse air flow at sea level)



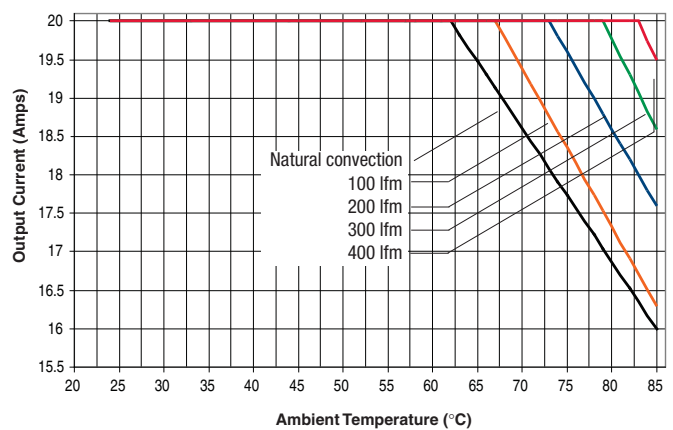
UVQ-3.3/35-D48 Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 48V$, transverse air flow at sea level)



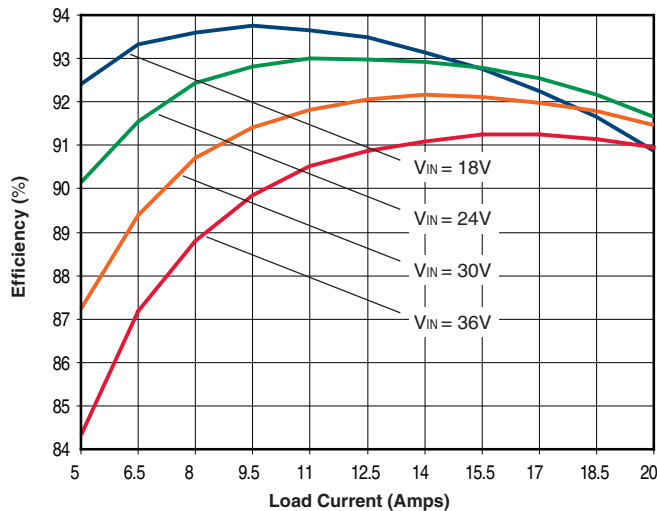
UVQ-5/20-D24P: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 24V$, transverse air flow)



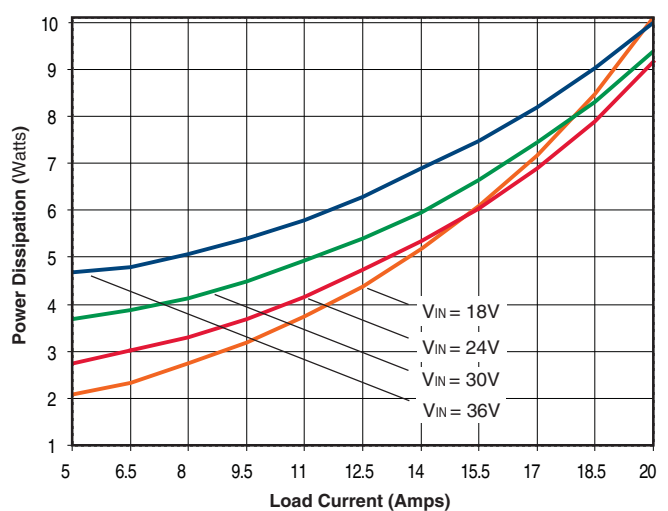
UVQ-5/20-D24PB: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 24V$, transverse air flow)



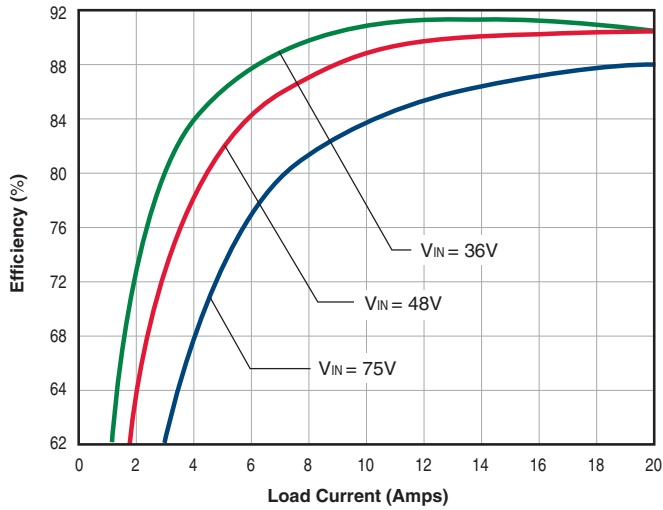
UVQ-5/20-D24P Efficiency vs. Line Voltage and Load Current @ +25°C



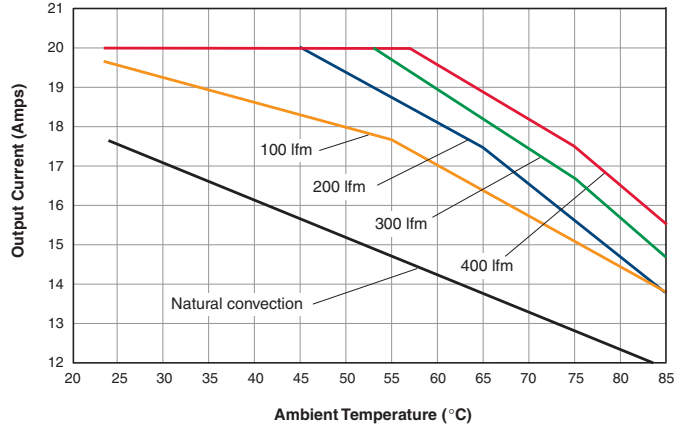
UVQ-5/20-D24 Power Dissipation vs. Load Current @ +25°C



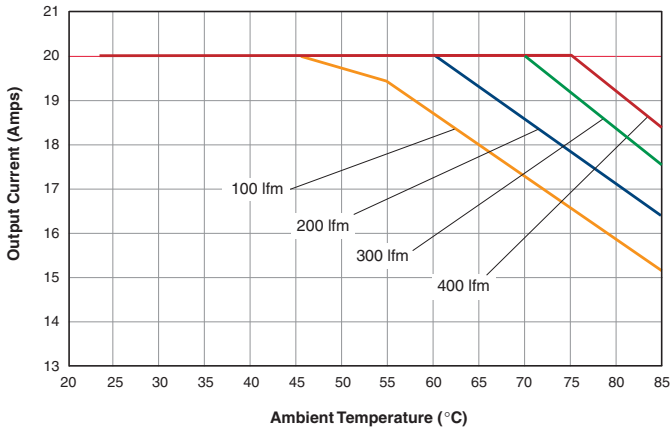
UVQ-5/20-D48
Efficiency vs. Line Voltage and Load Current @ 25°C



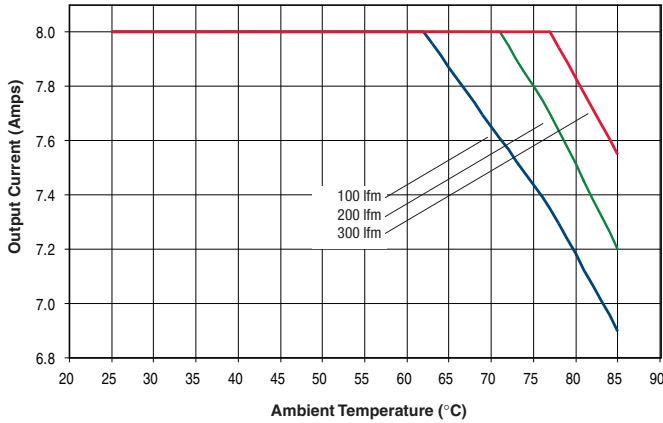
UVQ-5/20-D48P: Maximum Current Temperature Derating
(No baseplate, VIN = 48V, transverse air flow at sea level)



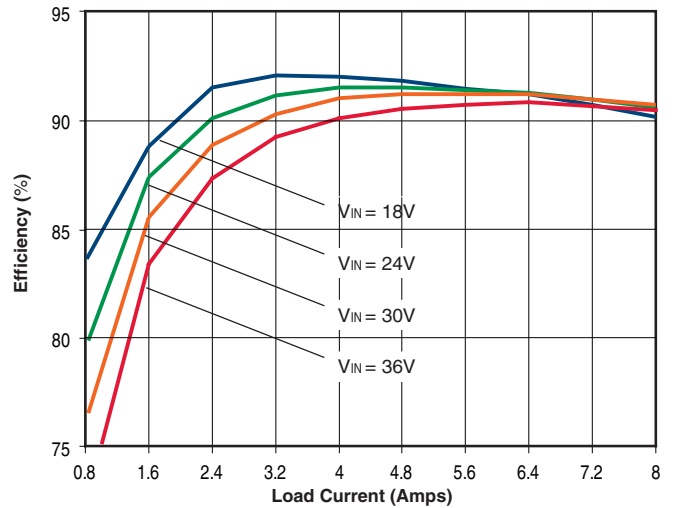
UVQ-5/20-D48PB: Maximum Current Temperature Derating
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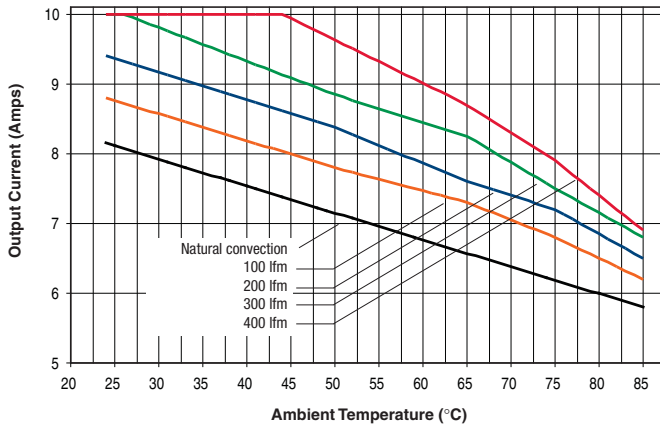
UVQ-12/8-D24P: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 24V$, transverse air flow)



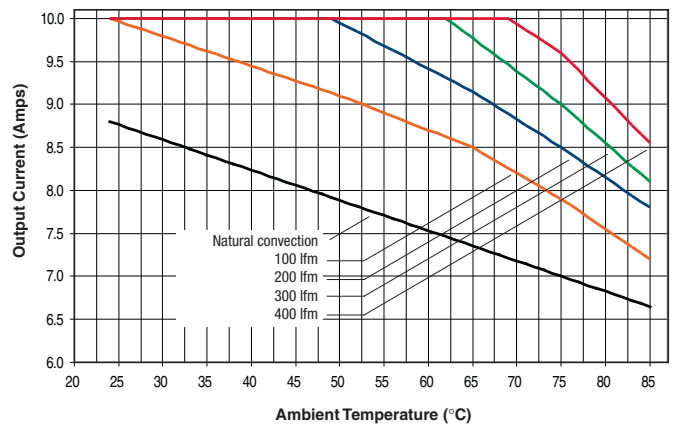
UVQ-12/8-D24P Efficiency vs. Line Voltage and Load Current @ +25°C



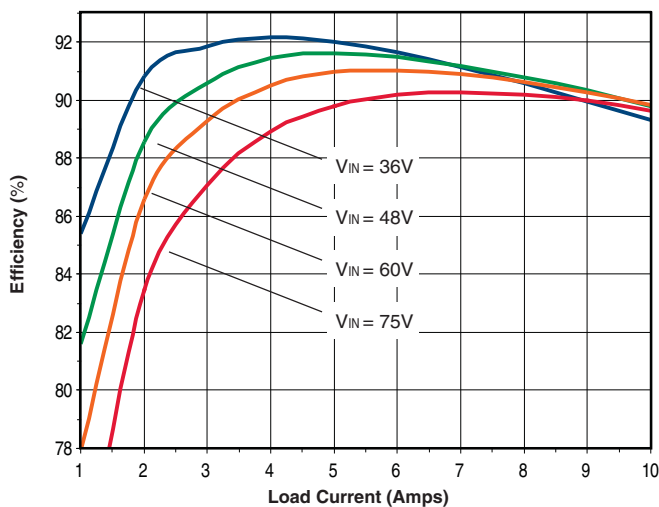
UVQ-12/10-D48N: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 48V$, transverse air flow)



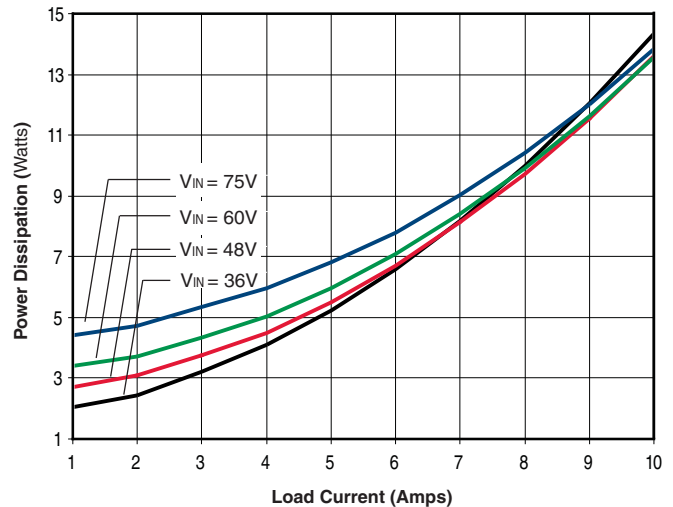
UVQ-12/10-D48N: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 48V$, transverse air flow)



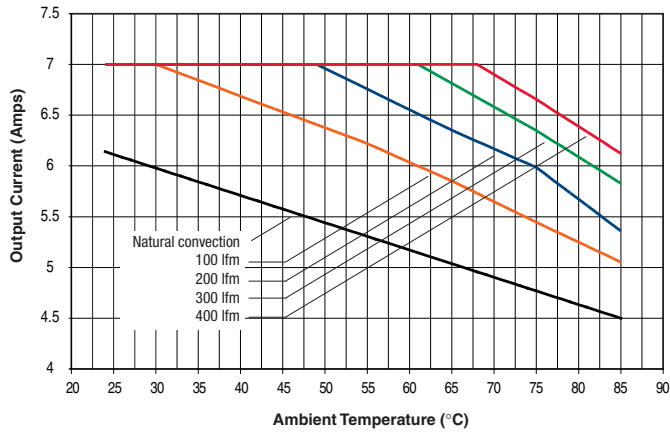
UVQ-12/10-D48N Efficiency vs. Line Voltage and Load Current @ +25°C



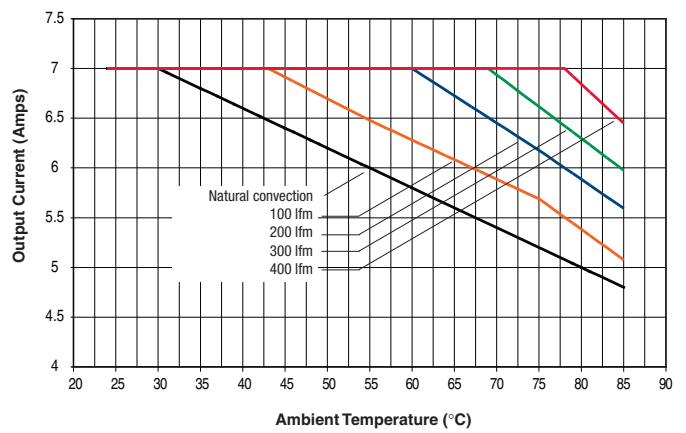
UVQ-12/10-D48N Power Dissipation vs. Load Current @ +25°C



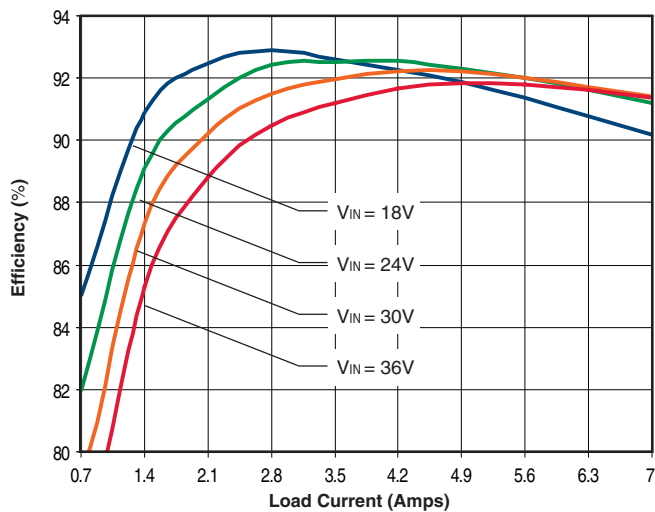
UVQ-15/7-D24N: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 24V$, transverse air flow)



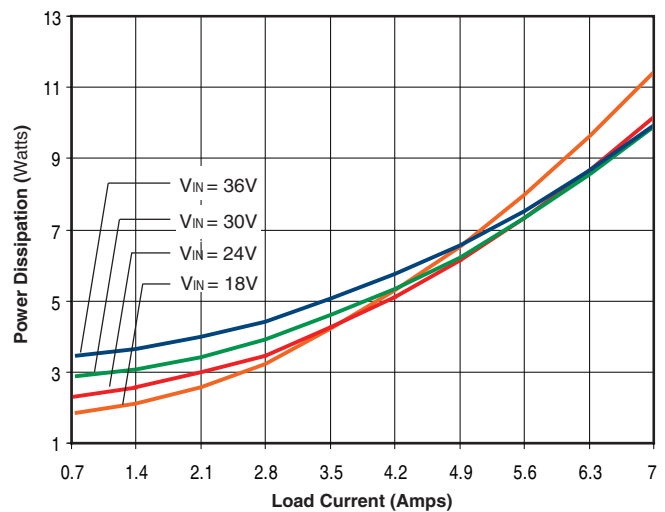
UVQ-15/7-D24N: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 24V$, transverse air flow)



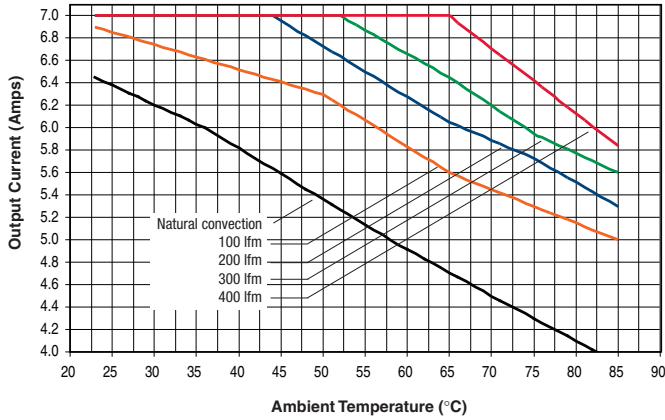
UVQ-15/7-D24N
Efficiency vs. Line Voltage and Load Current @ +25°C



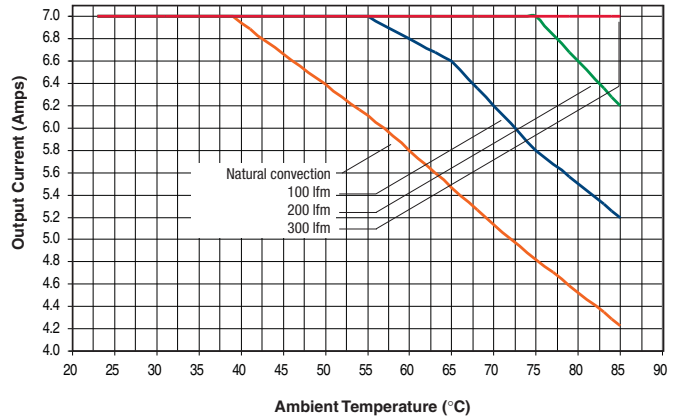
UVQ-15/7-D24N
Power Dissipation vs. Load Current @ +25°C



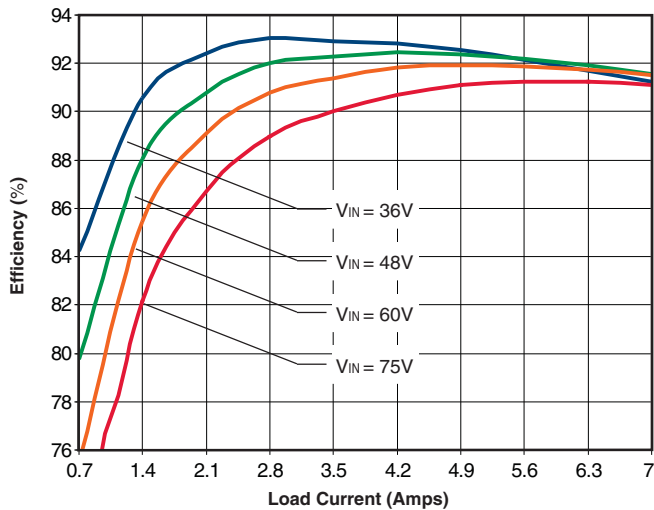
UVQ-15/7-D48N: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 48V$, transverse air flow)



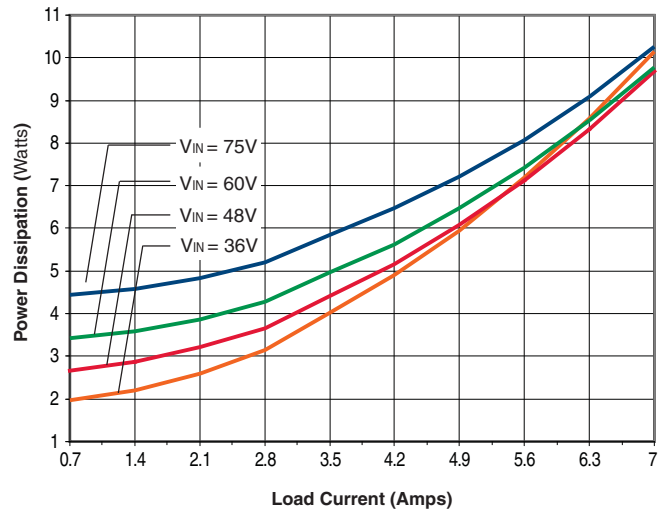
UVQ-15/7-D48N: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 48V$, transverse air flow)



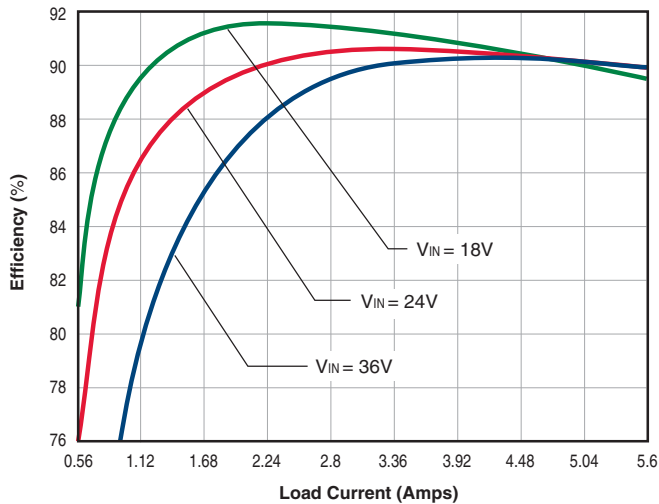
UVQ-15/7-D48N
Efficiency vs. Line Voltage and Load Current @ +25°C



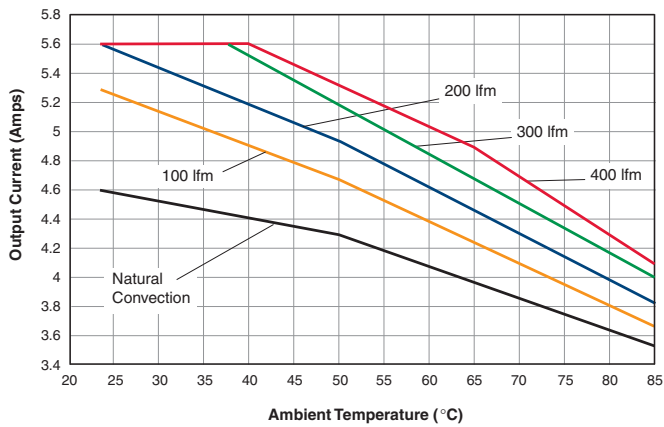
UVQ-15/7-D48N
Power Dissipation vs. Load Current @ +25°C



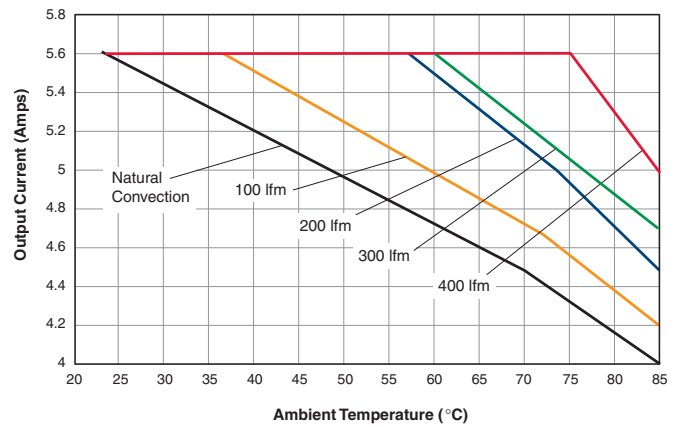
UVQ-18/5.6-D24
Efficiency vs. Line Voltage and Load Current @ 25°C



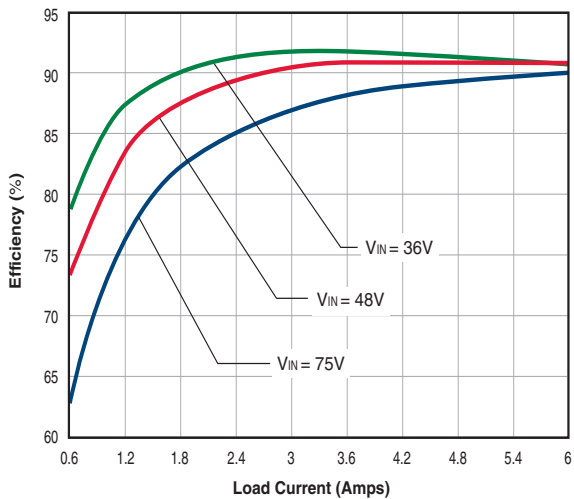
UVQ-18/5.6-D24: Maximum Current Temperature Derating
(No baseplate, V_{IN} = 24V, transverse air flow)



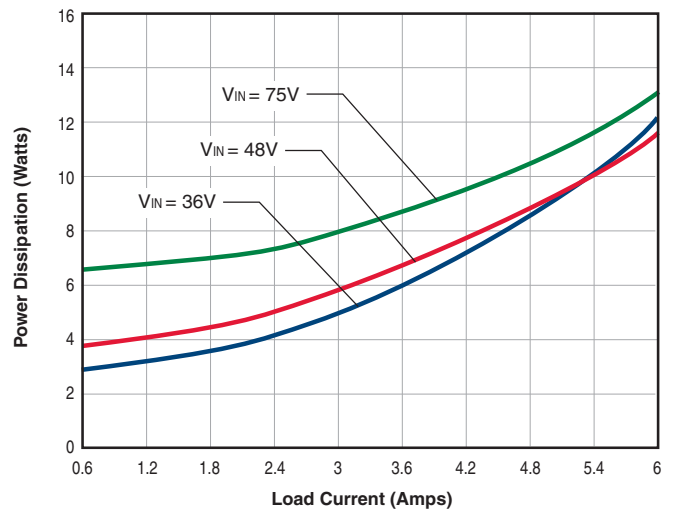
UVQ-18/5.6-D24: Maximum Current Temperature Derating
(With baseplate, V_{IN} = 24V, transverse air flow)



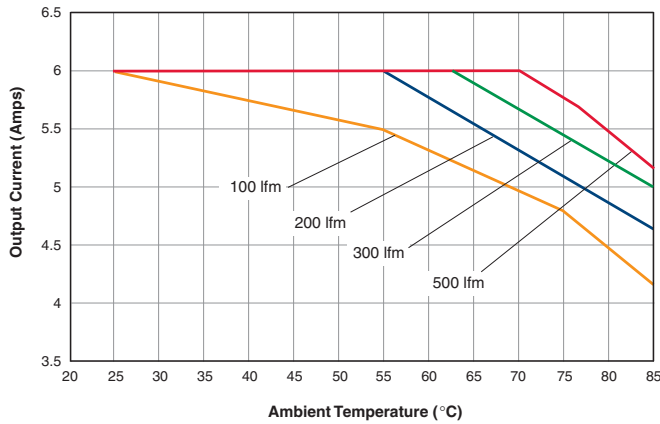
UVQ-18/6-D48N
Efficiency vs. Line Voltage and Load Current @ 25°C



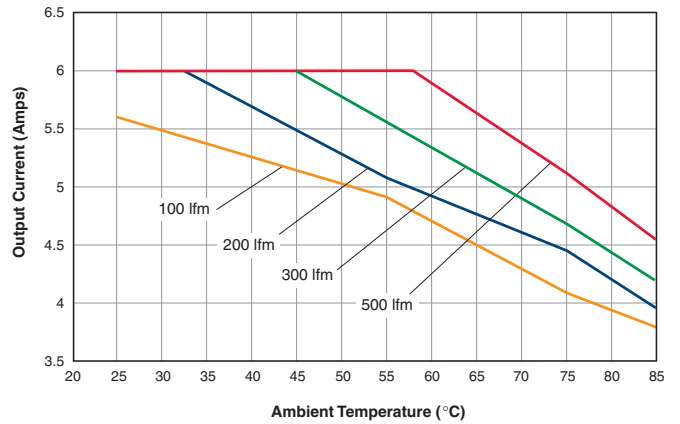
UVQ-18/6-D48
Power Dissipation vs. Load Current @ 25°C



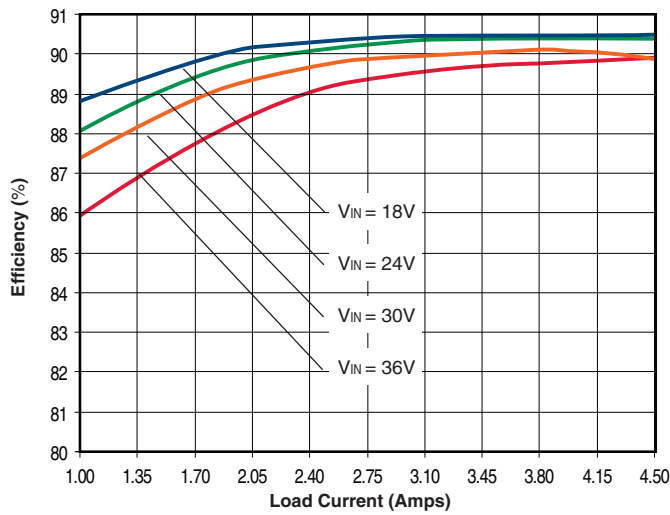
UVQ-18/6-D48: Maximum Current Temperature Derating
(With baseplate, $V_{IN} = 48V$, transverse air flow)



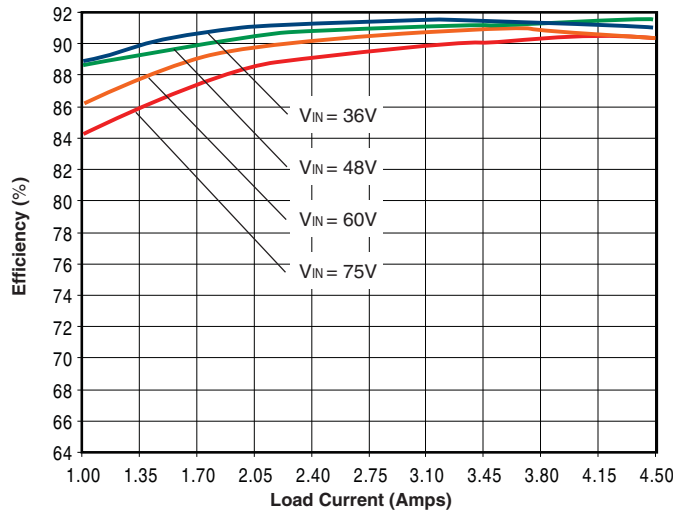
UVQ-18/6-D48: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 48V$, transverse air flow)



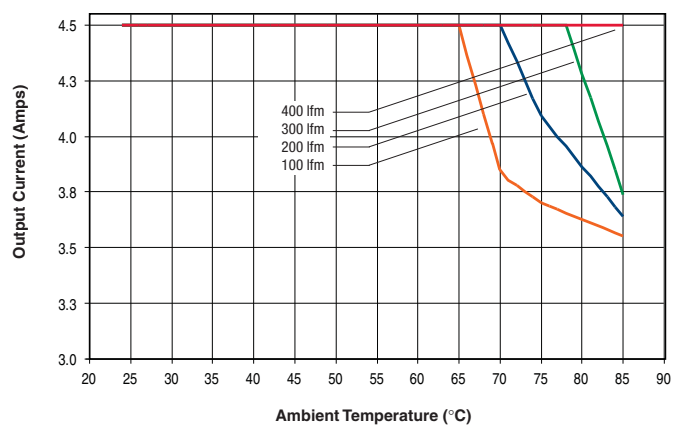
UVQ-24/4.5-D24N
Efficiency vs. Line Voltage and Load Current @ +25°C



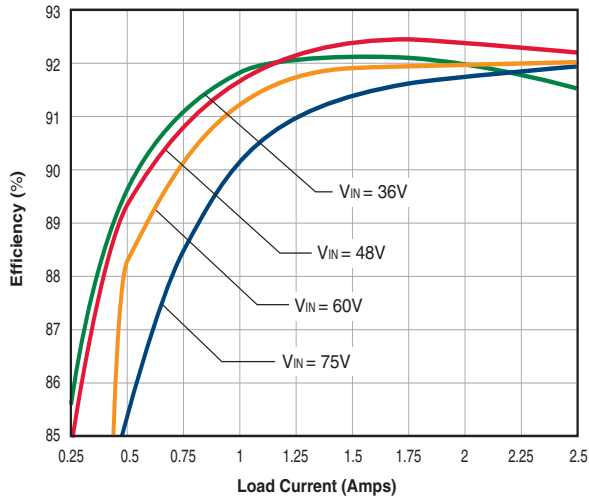
UVQ-24/4.5-D48N
Efficiency vs. Line Voltage and Load Current @ +25°C



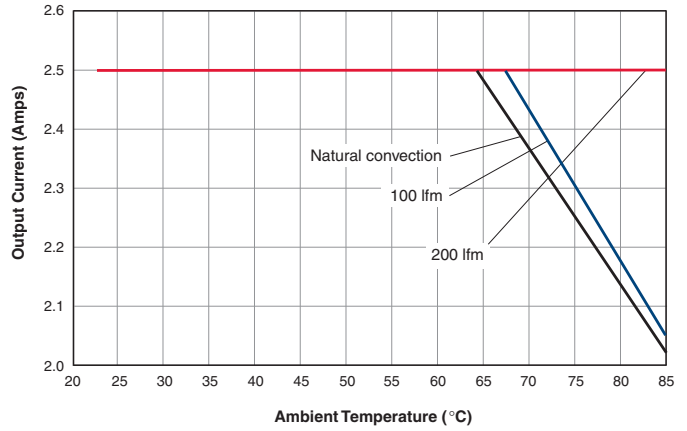
UVQ-24/4.5-D48N: Maximum Current Temperature Derating
(No baseplate, $V_{IN} = 48V$, transverse air flow)



UVQ-48/2.5-D48N
Efficiency vs. Line Voltage and Load Current @ 25°C



UVQ-48/2.5-D48N: Maximum Current Temperature Derating
(With baseplate, VIN = 48V, transverse air flow)



UVQ Series Aluminum Heatsink

The UVQ series converter baseplate can be attached either to an enclosure wall or a heatsink to remove heat from internal power dissipation. The discussion below concerns only the heatsink alternative. The UVQ's are available with a low-profile extruded aluminum heatsink kit, models HS-QB25-UVQ, HS-QB50-UVQ, and HS-QB100-UVQ. This kit includes the heatsink, thermal mounting pad, screws and mounting hardware. See the assembly diagram below. Do not overtighten the screws in the tapped holes in the converter. This kit adds excellent thermal performance without sacrificing too much component height. See the Mechanical Outline Drawings for assembled dimensions. If the thermal pad is firmly attached, no thermal compound (“thermal grease”) is required.

Thermal Performance

The HS-QB25-UVQ heatsink has a thermal resistance of 12 °C/Watt of internal heat dissipation with “natural convection” airflow (no fans or other mechanical airflow) at sea level altitude. This thermal resistance assumes that the heatsink is firmly attached using the supplied thermal pad and that there is no nearby wall or enclosure surface to inhibit the airflow. The thermal pad adds a negligible series resistance of approximately 0.5°C/Watt so that the total assembled resistance is 12.5°C/Watt.

Be aware that we need to handle only the internal heat dissipation, not the full power output of the converter. This internal heat dissipation is related to the efficiency as follows:

$$\text{Power Dissipation [Pd]} = \text{Power In} - \text{Power Out} \quad [1]$$

$$\text{Power Out} / \text{Power In} = \text{Efficiency [in \%]} / 100 \quad [2]$$

$$\text{Power Dissipation [Pd]} = \text{Power In} \times (1 - \text{Efficiency\%/100}) \quad [3]$$

$$\text{Power Dissipation [Pd]} = \text{Power Out} \times (1 / (\text{Efficiency\%/100}) - 1) \quad [4]$$

Efficiency of course varies with input voltage and the total output power. Please refer to the Performance Curves.

Since many applications do include fans, here is an approximate equation to calculate the net thermal resistance:

$$R_{\Theta} \text{ [at airflow]} = R_{\Theta} \text{ [natural convection]} / (1 + (\text{Airflow in LFM}) \times \text{[Airflow Constant]}) \quad [5]$$

Where,

R_{Θ} [at airflow] is the net thermal resistance (in °C/W) with the amount of airflow available and,

R_{Θ} [natural convection] is the still air total path thermal resistance or in this case 12.5°C/Watt and,

“Airflow in LFM” is the net air movement flow rate immediately at the converter.

This equation simplifies an otherwise complex aerodynamic model but is a useful starting point. The “Airflow Constant” is dependent on the fan and enclosure geometry. For example, if 200 LFM of airflow reduces the effective natural convection thermal resistance by one half, the airflow constant would be 0.005. There is no practical way to publish a “one size fits all” airflow constant because of variations in airflow direction, heatsink orientation, adjacent walls, enclosure geometry, etc. Each application must be determined empirically and the equation is primarily a way to help understand the cooling arithmetic.

This equation basically says that small amounts of forced airflow are quite effective removing the heat. But very high airflows give diminishing returns. Conversely, no forced airflow causes considerable heat buildup. At zero airflow, cooling occurs only because of natural convection over the heatsink. Natural convection is often well below 50 LFM, not much of a breeze.

While these equations are useful as a conceptual aid, most users find it very difficult to measure actual airflow rates at the converter. Even if you know the velocity specifications of the fan, this does not usually relate directly to the enclosure geometry. Be sure to use a considerable safety margin doing thermal analysis. If in doubt, measure the actual heat sink temperature with a calibrated thermocouple, RTD or thermistor. Safe operation should keep the heat sink below 100°C.

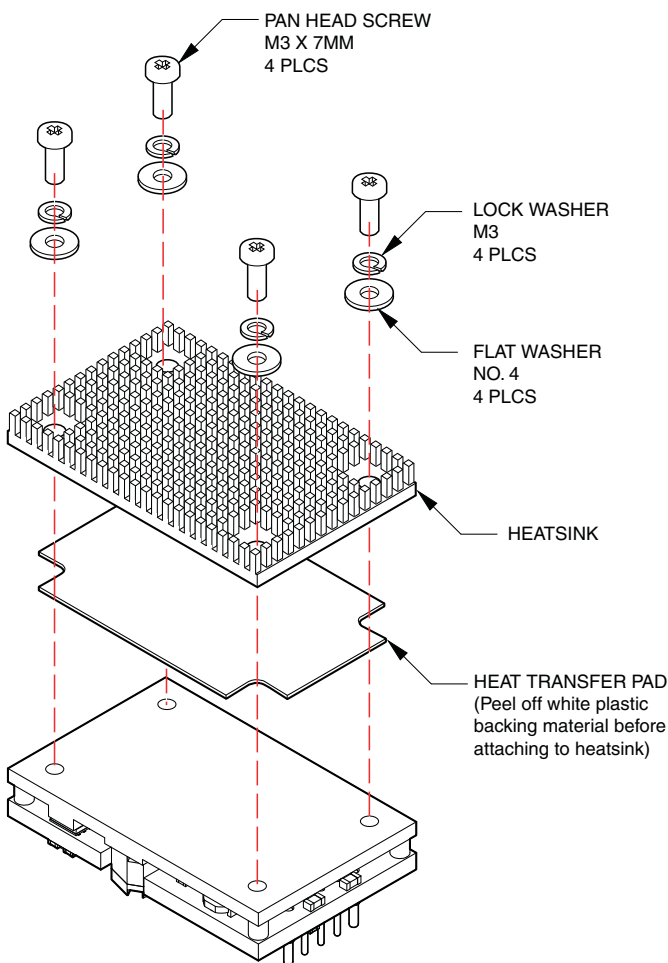


Figure 7. Model UVQ Heatsink Assembly Diagram

When assembling these kits onto the converter, include ALL kit hardware to assure adequate mechanical capture and proper clearances. Thread relief is 0.090" (2.3mm).

Calculating Maximum Power Dissipation

To determine the maximum amount of internal power dissipation, find the ambient temperature inside the enclosure and the airflow (in Linear Feet per Minute – LFM) at the converter. Determine the expected heat dissipation using the Efficiency curves and the converter Input Voltage. You should also compensate for lower atmospheric pressure if your application altitude is considerably above sea level.

The general procedure is to compute the expected temperature rise of the heatsink. If the heatsink exceeds +100°C. either increase the airflow and/or reduce the power output. Start with this equation:

$$\text{Internal Heat Dissipation [Pd in Watts]} = (Ts - Ta) / R\Theta \text{ [at airflow]} \text{ [6]}$$

where “Ta” is the enclosure ambient air temperature and,

where “Ts” is the heatsink temperature and,

where “RΘ [at airflow]” is a specific heat transfer thermal resistance (in degrees Celsius per Watt) for a particular heat sink at a set airflow rate. We have already estimated RΘ [at airflow] in the equations above.

Note particularly that Ta is the air temperature inside the enclosure at the heatsink, not the outside air temperature. Most enclosures have higher internal temperatures, especially if the converter is “downwind” from other heat-producing circuits. Note also that this “Pd” term is only the internal heat dissipated inside the converter and not the total power output of the converter.

We can rearrange this equation to give an estimated temperature rise of the heatsink as follows:

$$Ts = (Pd \times R\Theta \text{ [at airflow]}) + Ta \text{ [7]}$$

Heatsink Kit * Model Number	Still Air (Natural convection) thermal resistance	Heatsink height (see drawing)
HS-QB25-UVQ	12°C/Watt	0.25" (6.35mm)
HS-QB50-UVQ	10.6°C/Watt	0.50" (12.7mm)
HS-QB100-UVQ	8°C/Watt	1.00" (25.4mm)

* Kit includes heatsink, thermal pad and mounting hardware.

Heat Sink Example

Assume an efficiency of 92% and power output of 100 Watts. Using equation [4], Pd is about 8.7 Watts at an input voltage of 48 Volts. Using +30°C ambient temperature inside the enclosure, we wish to limit the heat sink temperature to +90°C maximum baseplate temperature to stay well away from thermal shut-down. The +90°C. figure also allows some margin in case the ambient climbs above +30°C or the input voltage varies, giving us less than 92% efficiency. The heat sink and airflow combination must have the following characteristics:

$$8.7 \text{ W} = (90-30) / R\Theta[\text{airflow}] \text{ or,}$$

$$R\Theta[\text{airflow}] = 60/8.7 = 6.9^\circ\text{C/W}$$

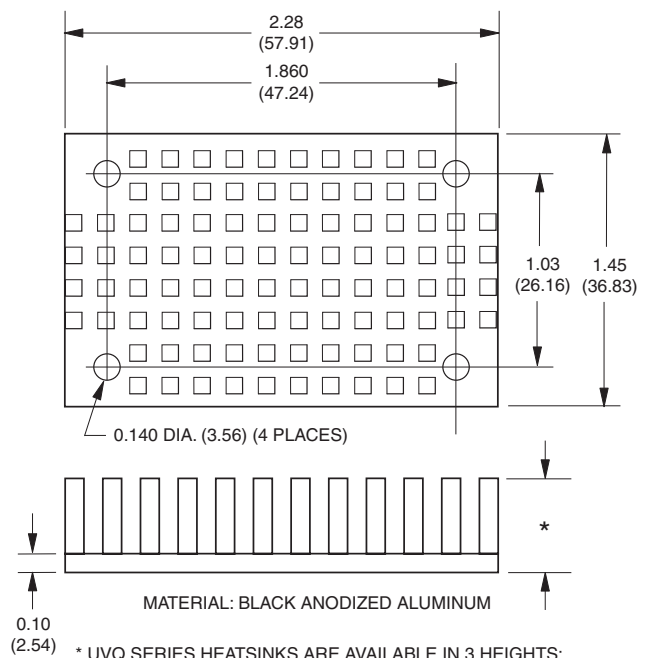
Since the ambient thermal resistance of the heatsink and pad is 12.5°C/W, we need additional forced cooling to get us down to 6.9°C/W. Using a hypothetical airflow constant of 0.005, we can rearrange equation [5] as follows:

$$(\text{Required Airflow, LFM}) \times (\text{Airflow Constant}) = R\Theta[\text{Nat.Convection}] / R\Theta[\text{at airflow}] - 1$$

or, (Required Airflow, LFM) x (Airflow Constant) = 12.5/6.9 – 1 = 0.81 and, rearranging again,

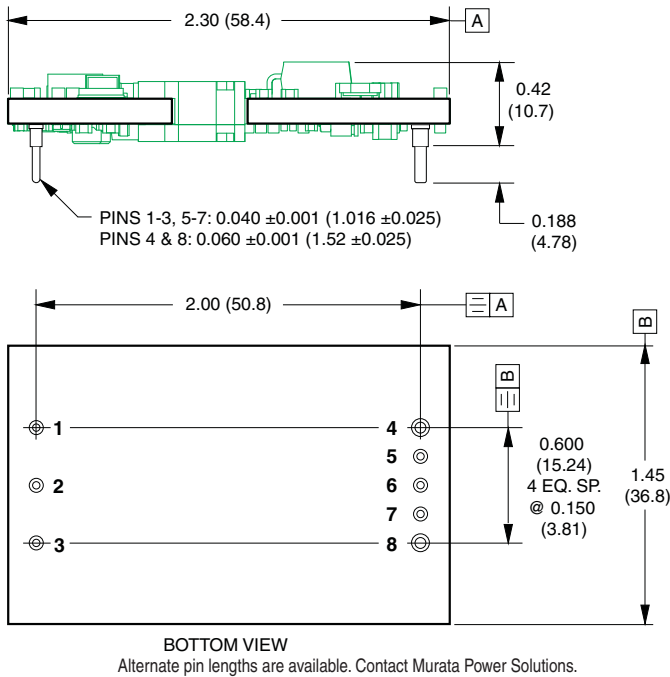
$$(\text{Required Airflow, LFM}) = 0.81/0.005 = 162 \text{ LFM}$$

162 LFM is the minimum airflow to keep the heatsink below +90°C. Increase the airflow to several hundred LFM to reduce the heatsink temperature further and improve life and reliability.

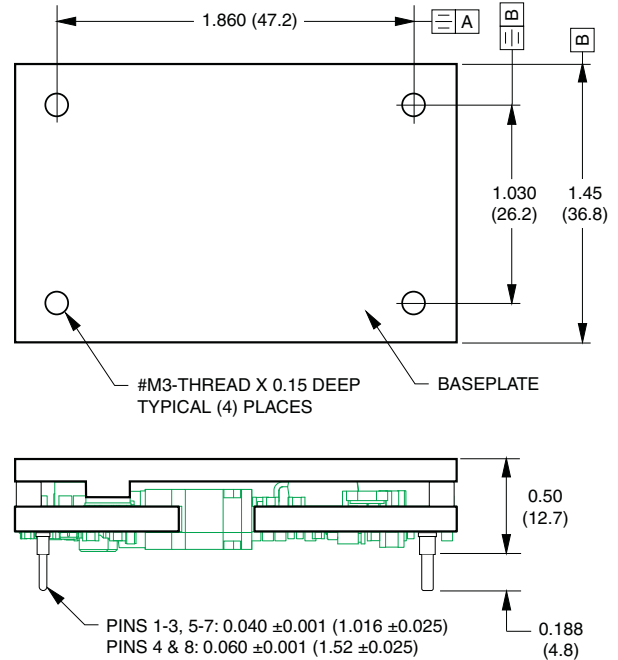


Dimensions in inches (mm)

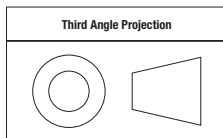
Case C59



Case C59 with Baseplate



Dimensions are in inches (mm shown for ref. only).



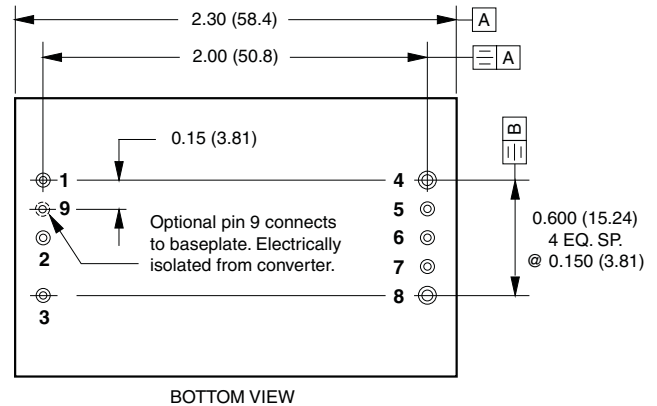
Tolerances (unless otherwise specified):
 .XX ± 0.02 (0.5)
 .XXX ± 0.010 (0.25)
 Angles ± 2°

Components are shown for reference only.

I/O CONNECTIONS	
Pin	Function P32
1	-Input
2	On/Off Control
3	+Input
4	-Output
5	-Sense
6	Output Trim
7	+Sense
8	+Output

Important: If sense inputs are not connected to a remote load, connect them to their respective V_{OUT} pins at the converter.

Optional baseplate pin is special order. Contact Murata Power Solutions.



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03/05/09
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