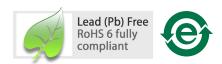
## **ADNS-2051**

# **Optical Mouse Sensor**



## **Data Sheet**





## **Description**

The ADNS-2051 is a low cost optical sensor used to implement a non-mechanical tracking engine for computer mice.

It is based on optical navigation technology, which measures changes in position by optically acquiring sequential surface images (frames) and mathematically determining the direction and magnitude of movement.

The sensor is housed in a 16-pin staggered dual inline package (DIP) that is designed for use with the HDNS-2100 Lens and HDNS-2200 Clip and HLMP-ED80-XX000 (639 nm LED illuminator source). There are no moving parts, and precision optical alignment is not required, facilitating high volume assembly.

The output format is two channel quadrature (X and Y direction) which emulates encoder photo-transistors. The current X and Y information are also available in registers accessed via a serial port.

Default resolution is specified as 400 counts per inch (cpi), with rates of motion up to 14 inches per second (ips).

Resolution can also be programmed to 800 cpi.

The part is programmed via a two wire serial port, through registers.

#### **Theory of Operation**

The ADNS-2051 is based on Optical Navigation Technology. It contains an Image Acquisition System (IAS), a Digital Signal Processor (DSP), a two-channel quadrature output, and a two wire serial port.

The IAS acquires microscopic surface images via the lens and illumination system provided by the HDNS-2100, 2200, and HLMP-ED80-XX000 LED. These images are processed by the DSP to determine the direction and distance of motion. The DSP generates the  $\Delta x$  and  $\Delta y$  relative displacement values that are converted into two channel quadrature signals.

#### **Features**

- Precise optical navigation technology
- No mechanical moving parts
- Complete 2D motion sensor
- Serial interface and/or quadrature interface
- Smooth surface navigation
- Programmable frame speed up to 2300 frames per sec (fps)
- Accurate motion up to 14 ips
- 800 cpi resolution
- High reliability
- High speed motion detector
- No precision optical alignment
- Wave solderable
- Single 5.0 volt power supply
- Shutdown pin for USB suspend mode operation
- Power conservation mode during times of no movement
- On chip LED drive with regulated current
- Serial port registers
  - Programming
  - Data transfer
- 16-pin staggered dual inline package (DIP)

#### **Applications**

- Mice for desktop PCs, workstations, and portable PCs
- Trackballs
- Integrated input devices

# Outline Drawing of ADNS-2051 Optical Mouse Sensor

#### Pinout

Pin	Pin	Description
1	SCLK	Serial port clock (input)
2	XA	XA quadrature output
3	XB	XB quadrature output
4	YB	YB quadrature output
5	YA	YA quadrature output
6	XY_LED	LED control
7	REFA	Internal reference
8	REFB	Internal reference
9	OSC_IN	Oscillator input
10	GND	System ground
11	OSC_OUT	Oscillator output
12	GND	System ground
13	$V_{DD}$	5.0 volt power supply
14	R_BIN	LED current bin resistor
15	PD	Power down pin, active high
16	SDIO	Serial data (input and output)

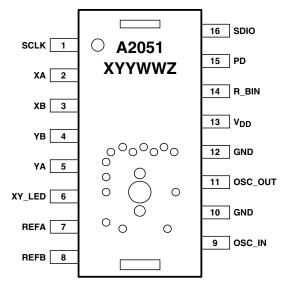
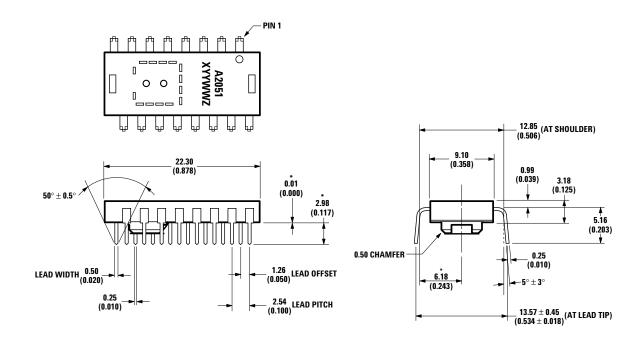
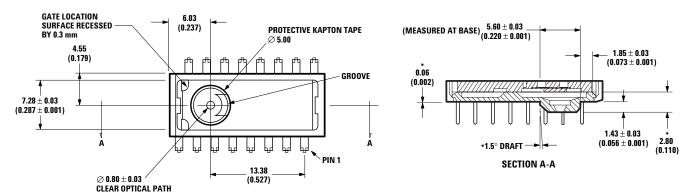


Figure 1. Top view





#### NOTES:

- 1. DIMENSIONS IN MILLIMETERS (INCHES).
- 2. DIMENSIONAL TOLERANCE: ± 0.1 mm.
  3. COPLANARITY OF LEADS: 0.1 mm.
- COPLANARITY OF LEADS: 0.1 mm.
   LEAD PITCH TOLERANCE: ± 0.15 mm.
- 5. CUMULATIVE PITCH TOLERANCE: ± 0.15 mm.
- 6. ANGULAR TOLERANCE: ± 3.0 DEGREES.
- 7. MAXIMUM FLASH + 0.2 mm.
- 8. CHAMFER (25 DEGREES x 2) ON THE TAPER SIDE OF THE LEAD.
- 9. \* THESE DIMENSIONS ARE FOR REFERENCES ONLY AND SHOULD NOT BE USED TO MECHANICALLY REFERENCE THE SENSOR.

Figure 2. Package outline drawing

#### **Overview of Optical Mouse Sensor Assembly**

2D Assembly Drawing of ADNS-2051

Figures 3 and 4, shown with HDNS-2100, HDNS-2200, and HLMP-ED80-XX000.

Avago Technologies provides an IGES file drawing describing the base plate molding features for lens and PCB alignment.

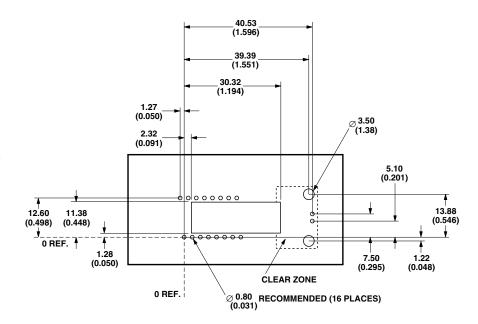
The components interlock as they are mounted onto defined features on the base plate.

The ADNS-2051 sensor is designed for mounting on a through hole PCB, looking down. There is an aperture stop and features on the package that align to the lens (see Figure 3).

The HDNS-2100 lens provides optics for the imaging of the surface as well as illumination of the surface at the optimum angle. Features on the lens align it to the sensor, base plate, and clip with the LED. The lens also has a large round flange to provide a long creepage path for any ESD events that occur at the opening of the base plate (see Figure 4).

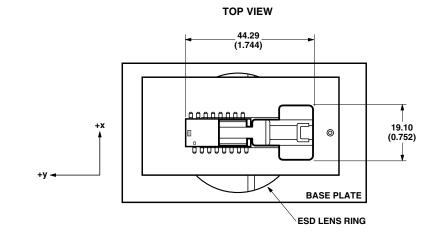
The HDNS-2200 clip holds the LED in relation to the lens. The LED must be inserted into the clip and the LED's leads formed prior to loading on the PCB. The clip interlocks the sensor to the lens, and through the lens to the alignment features on the base plate.

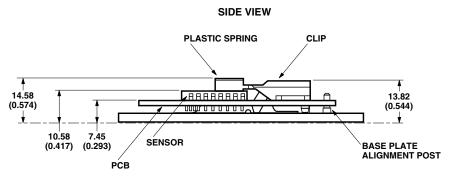
The HLMP-ED80-XX000 LED is recommended for illumination. If used with the bin table, sufficient illumination can be guaranteed.



**DIMENSIONS IN MILLIMETERS (INCHES)** 

Figure 3. Recommended PCB mechanical cutouts and spacing (top view)





DIMENSIONS IN MILLIMETERS (INCHES).

Figure 4. 2D assembly drawing of ADNS-2051 (top and side view)

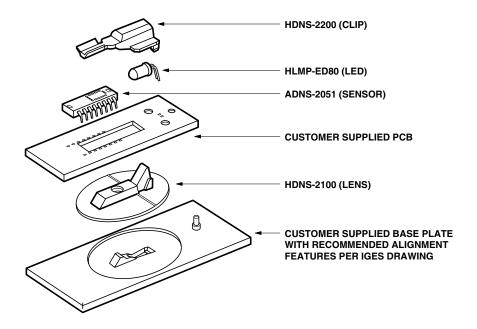


Figure 5. Exploded view drawing

## **PCB Assembly Considerations**

- Insert the sensor and all other electrical components into PCB.
- Bend the LED leads 90° and then insert the LED into the assembly clip until the snap feature locks the LED base.
- 3. Insert the LED/clip assembly into PCB.
- 4. Wave Solder the entire assembly in a no-wash solder process utilizing solder fixture. The solder fixture is needed to protect the sensor during the solder process. The fixture should be designed to expose the sensor leads to solder while shielding the optical aperture from direct solder contact. The solder fixture is also used to set the reference height of the sensor to the PCB top during wave soldering (Note: DO NOT remove the kapton tape during wave soldering).
- 5. Place the lens onto the base plate.

- 6. Remove the protective kapton tape from optical aperture of the sensor. Care must be taken to keep contaminants from entering the aperture. It is recommended not to place the PCB facing up during the entire mouse assembly process. The PCB should be held vertically during the kapton removal process.
- 7. Insert PCB assembly over the lens onto the base plate aligning post to retain PCB assembly. The sensor aperture ring should self-align to the lens.
- 8. The optical position reference for the PCB is set by the base plate and lens. Note that the PCB motion due to button presses must be minimized to maintain optical alignment.
- Install mouse top case. There MUST be a feature in the top case to press down onto the clip to ensure all components are interlocked to the correct vertical height.

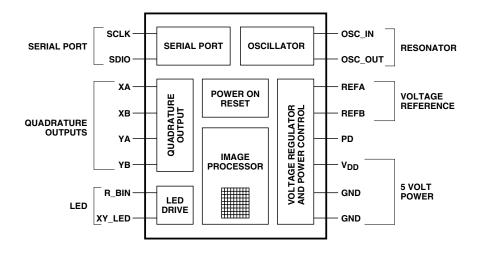


Figure 6. Block diagram of ADNS-2051 optical mouse sensor

#### **Design Considerations for Improving ESD Performance**

The flange on the lens has been designed to increase the creepage and clearance distance for electrostatic discharge. The table on the right shows typical values assuming base plate construction per the Avago supplied IGES file and HDNS-2100 lens flange.

For improved ESD performance, the lens flange can be sealed (i.e. glued) to the base plate. Note that the lens material is polycarbonate and therefore, cyanoacrylate-based adhesives or other adhesives that may damage the lens should NOT be used.

The trimmed lens, HDNS-2100#001, is not recommended for corded applications due to the ESD spec requirement.

Typical Distance	Millimeters
Creepage	16.0
Clearance	2.1

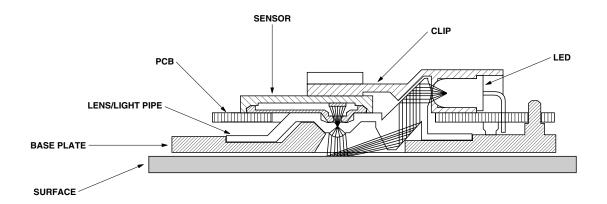


Figure 7. PCB assembly

## **Recommended Typical Application Using SDIO Pins**

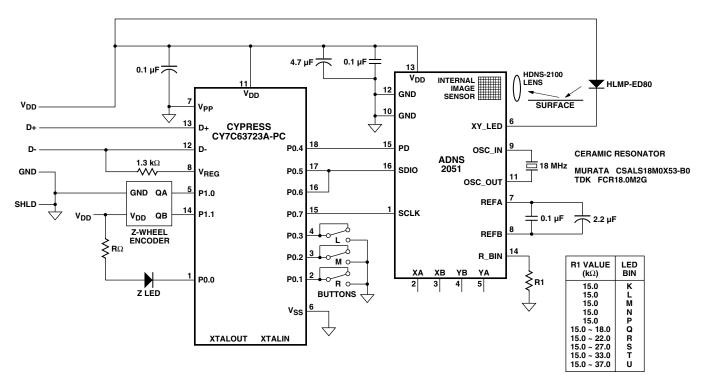


Figure 8. Application using SDIO pins

#### **Notes on Bypass Capacitors:**

- Caps for pins 7, 8 and 12, 13 MUST have trace lengths LESS than 5 mm.
- The 0.1 μF caps must be ceramic.
- Caps should have less than 5 nH of self inductance
- Caps should have less than 0.2 Ω ESR

Surface mount parts are recommended.

SDIO and SCLK pins should be grounded if not used.

## **Regulatory Requirements**

- Passes FCC B and worldwide analogous emission limits when assembled into a mouse with unshielded cable and following Avago recommendations.
- Passes EN61000-4-4/IEC801-4 EFT tests when assembled into a mouse with unshielded cable and following Avago recommendations.
- UL flammability level UL94 V-0.
- Provides sufficient ESD creepage/clearance distance to avoid discharge up to 15 kV when assembled into a mouse according to usage instructions above.
- For eye safety consideration, please refer to the technical report available on the web site,
  - http://www.Avago.com
- The 15.0  $k\Omega$  resistor is determined by the absolute maximum rating of 50 mA for the HLMP-ED80-XX000. The other resistor values for brighter bins will guarantee good signals with reduced power.

## **Alternative Application using Quadrature Output Pins**

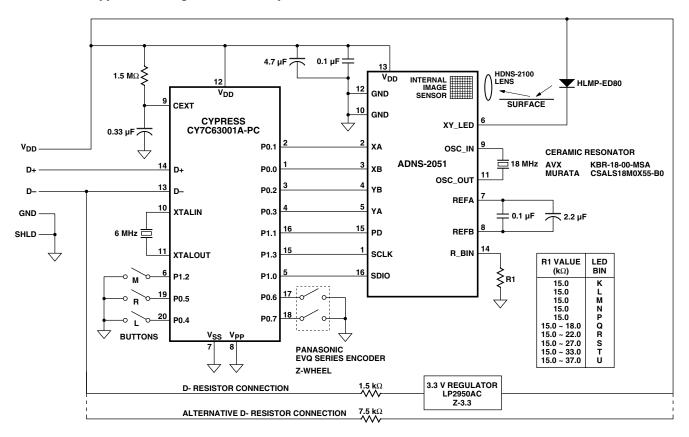


Figure 9. Application using quadrature output pins

#### **Absolute Maximum Ratings**

Absolute maximum nating	,-				
Parameter	Symbol	Min.	Max.	Units	Notes
Storage Temperature	T <sub>S</sub>	-40	85	°C	
Operating Temperature	T <sub>A</sub>	-15	55	°C	
Lead Solder Temperature			260	°C	For 10 seconds, 1.6 mm below seating plane.
Supply Voltage	V <sub>DD</sub>	-0.5	5.5	V	
ESD			2	kV	All pins, human body model MIL 883 Method 3015
Input Voltage	V <sub>IN</sub>	-0.5	V <sub>DD</sub> + 0.5	V R_BIN	PD, SDIO, SCLK, XA, XB, YA, YB, XY_LED,
Input Voltage	V <sub>IN</sub>	-0.5	3.6	V	OSC_IN, OSC_OUT, REF_A

# **Recommended Operating Conditions**

Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
Operating Temperature	T <sub>A</sub>	0		40	°C	
Power Supply Voltage	$V_{DD}$	4.25	5.0	5.5	volts	Register values retained for voltage transients below 4.25 V but greater than 4 V.
Power Supply Rise Time	V <sub>RT</sub>			100	ms	<u> </u>
Supply Noise	V <sub>N</sub>			100	mV	Peak to peak within 0-100 MHz.
Clock Frequency	f <sub>CLK</sub>	17.4	18.0	18.7	MHz	Set by ceramic resonator.
Serial Port Clock Frequency	SCLK			f <sub>CLK</sub> /4	MHz	
Resonator Impendance	X <sub>RES</sub>			55	Ω	
Distance from Lens Reference Plane to Surface	Z	2.3	2.4	2.5	mm	Results in $\pm 0.2$ mm DOF. (See Figure 10.)
Speed	S	0		14	in/sec	@ frame rate = 1500/second.
Acceleration	Α			0.15	g	@ frame rate = 1500/second.
Light Level onto IC	IRR <sub>INC</sub>	80 100		25,000 30,000	mW/m <sup>2</sup>	$\lambda = 639 \text{ nm}$ $\lambda = 875 \text{ nm}$
SDIO Read Hold Time	t <sub>HOLD</sub>	100			μs	Hold time for valid data. (Refer to Figure 28.)
SDIO Serial Write-Write Time	tsww	100			μs	Time between two write commands. (Refer to Figure 31.)
SDIO Serial Write-Read Time	t <sub>SWR</sub>	100			μs	Time between write and read operation. (Refer to Figure 32.)
SDIO Serial Read-Write Time	t <sub>SRW</sub>	120			ns	Time between read and write operation. (Refer to Figure 33.)
SDIO Serial Read-Read Time	t <sub>SRR</sub>	120			ns	Time between two read commands. (Refer to Figure 33.)
Data Delay after PD ↓	tcompute	3.2			ms	After t <sub>COMPUTE</sub> , all registers contain data from first image after PD↓. Note that an additional 75 frames for AGC (shutter) stabilization may be required if mouse movement occurred while PD was high. (Refer to Figure 12.)
SDIO Write Setup Time	t <sub>SETUP</sub>	60			ns	Data valid time before the rising of SCLK. (Refer to Figure 26.)
PD Pulse Width (to power down the chip)	t <sub>PDW</sub>	700			μs	Pulse width to initiate the power down cycle @ 1500 fps. (Refer to Figure 12 and Figure 14.)
PD Pulse Width (to reset the serial port)	t <sub>PDR</sub>	100			μs	Pulse width to reset the serial port @ 1500 fps (but may also initiate a power down cycle.  Normal PD recovery sequence to be followed. (Refer to Figure 15.)
Frame Rate	FR		1500		frames/s	See Frame_Period register section.
Bin Resistor	R1	15 K	15 K	37 K	Ω	Refer to Figure 8.

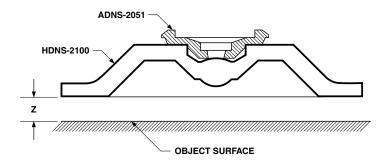


Figure 10. Distance from lens reference plane to surface

# **AC Electrical Specifications**

Electrical Characteristics over recommended operating conditions. Typical values at  $25^{\circ}$ C,  $V_{DD} = 5.0 \text{ V}$ , 1500 fps, 18 MHz.

Parameter		Symbol	Min.	Тур.	Max.	Units	Notes
Power Down		t <sub>PD</sub>		700		μs	From PD ↑ Time uncertainty due to firm- ware delay. (Refer to Figure 12.)
Power Up from PD ↓		t <sub>PUPD</sub>			50	ms	From PD↓to valid quad signals 705 µsec + 75 frames. (Refer to Figure 12.)
Power Up from V <sub>DD</sub> ↑		t <sub>PU</sub>			30		From V <sub>DD</sub> ↑to valid quad signals 705 µsec + 40 frames
Rise and Fall Times: SDIO  XA, XB, YA, YB		t <sub>r</sub>		30		ns	C <sub>L</sub> = 30 pF (the rise time is between 10% and 90%)
		t <sub>f</sub>		16		ns	$C_L = 30 \text{ pF}$ (the fall time is between 10% and 90%)
		t <sub>r</sub>		50		ns	$C_L = 30 \text{ pF}$ (the rise time is between 10% and 90%)
		t <sub>f</sub>		20		ns	$C_L = 30 \text{ pF}$ (the fall time is between 10% and 90%)
ILED		t <sub>r</sub>		40		ns	With HLMP-ED80-XX000 LED (the rise time is between 10% and 90%)
		t <sub>f</sub>		200		ns	With HLMP-ED80-XX000 LED (the fall time is between 10% and 90%)
Serial Port Transaction Timer		t <sub>SPTT</sub>	0.7	0.9	1.0	S	Serial port will reset if current transaction is not complete within t <sub>SPTT</sub> . (Refer to Figure 36.)
Transient Supply Current		I <sub>DDT</sub>		20	37	mA	Max. supply current during a V <sub>DD</sub> ramp from 0 to 5.0 V with > 500 µs rise time. Does not include charging current for bypass capacitors.

**DC Electrical Specifications**Electrical Characteristics over recommended operating conditions. Typical values at 25°C, V<sub>DD</sub> = 5.0 V, 18 MHz.

Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
DC Supply Current (mouse moving)	I <sub>DD</sub> avg		15	25	mA	No load on XA, XB, YA, YB, SCLK, SDIO. Excluding LED current.
Peak Supply Current (mouse moving)	I <sub>DD</sub> PEAK		20		mA	No load on XA, XB, YA, YB, SCLK, SDIO. Excluding LED current.
DC Supply Current (mouse not moving)	I <sub>DD</sub>		12	25	mA	No load on XA, XB, YA, YB, SCLK, SDIO. Excluding LED current.
DC Supply Current I <sub>DDPD</sub> (power down)			170	240	μΑ	PD = high; SCLK, SDIO = GND or $V_{DD}$ ; $V_{DD}$ = 4.25 V to 5.25 V.
SCLK, SDIO, PD						
Input Low Voltage	V <sub>IL</sub>			0.8	V	
Input High Voltage	V <sub>IH</sub>	0.5 * V <sub>DD</sub>			V	
Output Low Voltage	V <sub>OL</sub>			0.7	V	@ I <sub>OL</sub> = 2 mA (SDIO only)
Output High Voltage	V <sub>OH</sub>	0.6 * V <sub>DD</sub>			V	$@ I_{OH} = 2 \text{ mA (SDIO only)}$
Output Low Voltage (XA, XB, YA, YB)	V <sub>OL</sub>			0.4	V	@ I <sub>OL</sub> = 0.5 mA.
Output High Voltage (XA, XB, YA, YB)	$V_{OH}$	0.6 * V <sub>DD</sub>			V	$@ I_{OH} = 0.5 \text{ mA}.$
Output Low Voltage (XY_LED)	$V_{OL}$			1.1	V	Refer to Figure 11.
XY LED Current	I <sub>LED</sub>	Typ-15%	630/R1	Typ + 15%	Α	Refer to Figure 11, see table below.
XY LED Current (fault mode)	I <sub>LED</sub>			500	μΑ	R1 < 200 Ω.
REF_A (normal mode)	V <sub>REFA</sub>		3.3		V	1.5 K $\Omega$ to 3.0 V or GND, PD = low.
REF_A (power down mode)	V <sub>REFA</sub>		3.3		V	1.5 K $\Omega$ to 3.0 V or GND, PD = high.

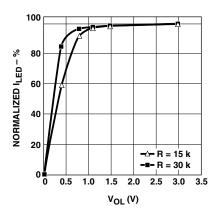


Figure 11. Typical I-V characteristic of ADNS-2051 XY\_LED pin

# **Typical LED Current Table**

R1 Value	kΩ	15	18	22	27	33	37
LED current (typical)	mA	42	35	29	23	19	17

#### **PD Pin Timing**

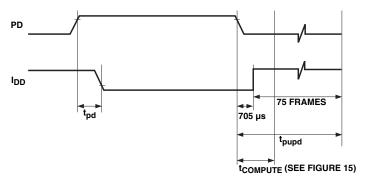
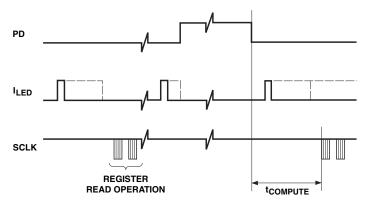


Figure 12. PD timing normal mode



PD

t<sub>PDW</sub>

700 μs

(POWER DOWN)

Figure 13. PD timing sleep mode

Figure 14. PD minimum pulse width

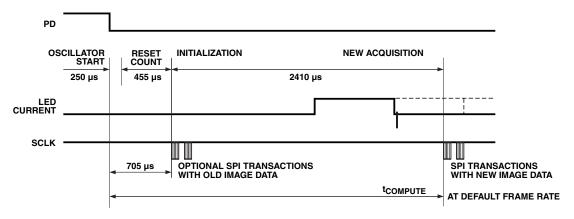


Figure 15. Detail of PD falling edge timing

#### **Quadrature Mode Timing**

The output waveforms emulate the output from encoders. With the resolution set to 400 cpi, from one to five quadrature states can exist within one frame time. The minimum state time is 133  $\mu$ s. If the resolution is 800 cpi, then up to ten quadrature states can exist within a frame time. If the motion within a frame is greater than these values, the extra motion will be reported in the

next frame. The following diagrams (see Figures 16, 17, and 18) show the timing for positive X motion, to the right or positive Y motion, up. If a power down via the PD pin occurs during a transfer, the transfer will resume after PD is de-asserted. The timing for that quadrature state will be increased by the length of the PD time.

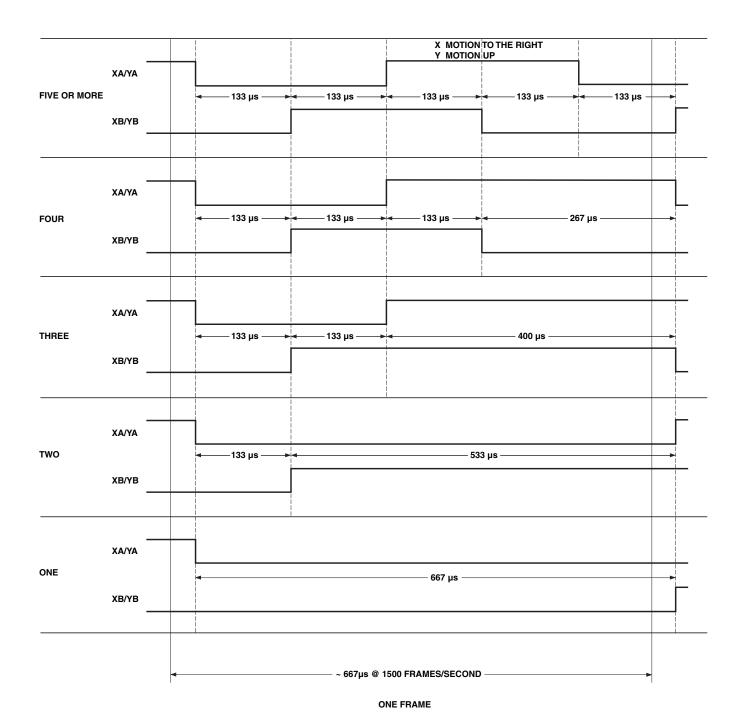


Figure 16. Quadrature states per frame (400 cpi mode)

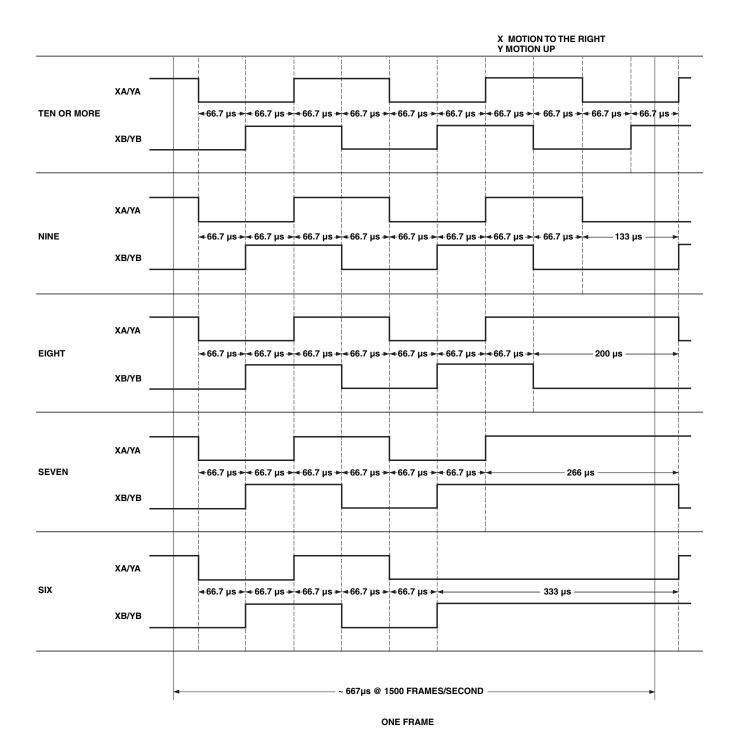


Figure 17. Quadrature states per frame (800 cp imode)

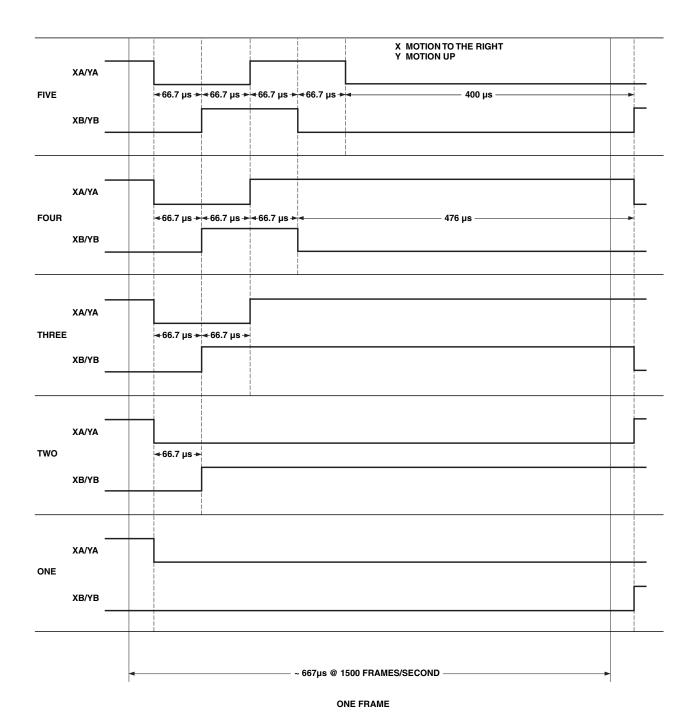


Figure 18. Quadrature states per frame (800 cpi mode)

#### **Quadrature State Machine**

The following state machine shows the states of the quadrature pins. The two things to note are that while the PD pin is asserted, the state machine is halted. Once PD is de-asserted, the state machine picks up from where it left off. State 0 is entered after a power up reset.

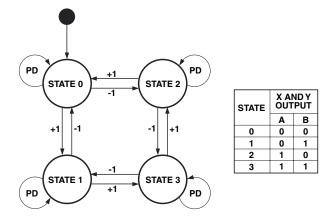


Figure 19. Quadrature state machine

## **Quadrature Output Waveform**

The two channel quadrature outputs are 5.0 volt CMOS outputs. The  $\Delta x$  count is used to generate the XA and XB signals, and  $\Delta y$  count is used for the YA and YB signals.

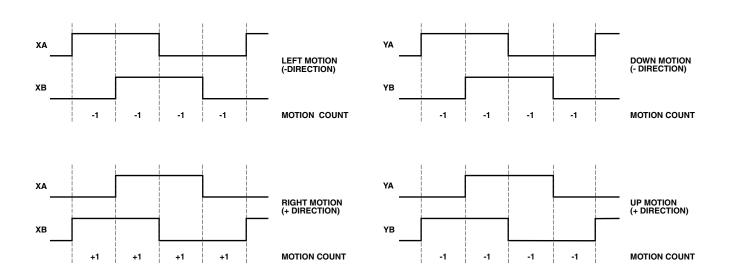


Figure 20. Quadrature output waveform

#### **Typical Performance Characteristics**

Performance characteristics over recommended operating conditions. Typical values at 25°C, V<sub>DD</sub> = 5.0 V, 18 MHz.

Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
Path Error (Deviation)	P <sub>ERROR</sub>		0.5		%	Path Error (Deviation) is the error from the
						ideal cursor path. It is expressed as a
						percentage of total travel and is measured
						over standard surfaces.

The following graphs (Figures 21, 22, 23, and 24) are the typical performance of the ADNS-2051 sensor, assembled as shown in the 2D assembly drawing with the HDNS-2100 Lens/Prism, the HDNS-2200 clip, and the HLMP-ED80-XX000 LED (page 3, Figure 4).

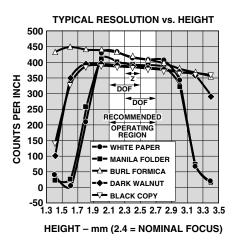


Figure 21. Typical resolution vs. Z (comparative surfaces) $^{\left[2,3\right]}$ 

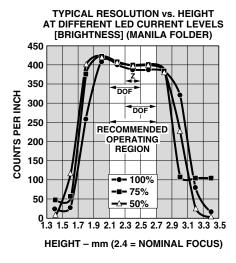


Figure 23. Typical resolution vs. z (manila folder and LED variation)<sup>[2,3]</sup>

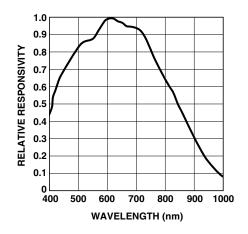


Figure 22. Wavelength responsitivity<sup>[1]</sup>

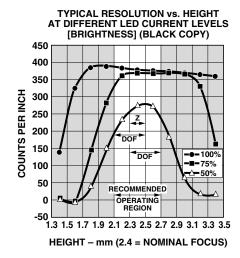


Figure 24. Typical resolution vs. z (black copy and LED variation) $^{[2,3]}$ 

#### Note

- 1. The ADNS-2051 is designed for optimal performance when used with the HLMP-ED80-XX000 (red LED 639 nm). For use with other LED colors (i.e., blue, green), please consult factory. When using alternate LEDs, there may also be performance degradation and additional eye safety considerations.
- 2. Z = Distance from Lens Reference plane to Surface.
- 3. DOF = Depth of Field.

#### **Synchronous Serial Port**

The synchronous serial port is used to set and read parameters in the ADNS-2051, and can be used to read out the motion information instead of the quadrature data pins.

The port is a two wire, half duplex port. The host microcontroller always initiates communication; the ADNS-2051 never initiates data transfers.

SCLK: The serial port clock. It is always generated by the master (the micro-controller).

SDIO: The data line.

PD: A third line is sometimes involved. PD (Power Down) is usually used to place the ADNS-2051 in a low power mode to meet USB suspend specification. PD can also be used to force resynchronization between the micro-controller and the ADNS-2051 in case of an error.

#### **Write Operation**

Write operations, where data is going from the micro-controller to the ADNS-2051, is always initiated by the micro-controller and consists of two bytes. The first byte contains the address (seven bits) and has a "1" as its MSB to indicate data direction. The second byte contains the data. The transfer is synchronized by SCLK. The micro-controller changes SDIO on falling edges of SCLK. The ADNS-2051 reads SDIO on rising edges of SCLK.

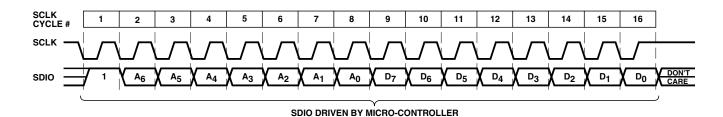


Figure 25. Write operation

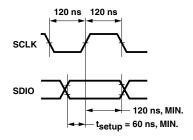


Figure 26. SDIO setup and hold times SCLK pulse width

#### **Read Operation**

A read operation, which means that data is going from the ADNS-2051 to the micro-controller, is always initiated by the micro-controller and consists of two bytes. The first byte contains the address, is written by the micro-controller, and has a "0" as its MSB to indicate data direction. The second byte contains the data and is driven by the ADNS-2051. The transfer is synchronized by SCLK. SDIO is changed on falling edges of SCLK and read on every rising

edge of SCLK. The micro-controller must go to a high Z state after the last address data bit. The ADNS-2051 will go to the high Z state after the last data bit (see detail "B" in Figure 28). One other thing to note during a read operation is that SCLK will need to be delayed after the last address data bit to ensure that the ADNS-2051 has at least 100  $\mu$ s to prepare the requested data. This is shown in the timing diagrams below.

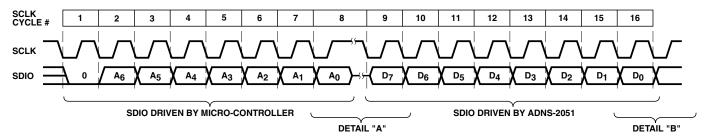


Figure 27. Read operation

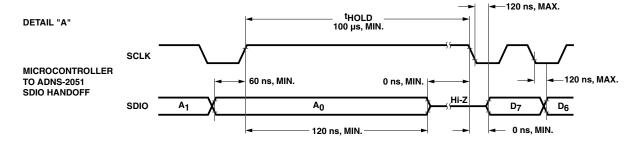


Figure 28. Microcontroller to ADNS-2051 SDIO handoff

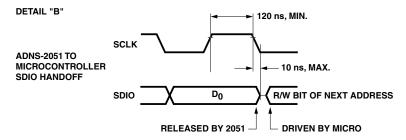


Figure 29. ADNS-2051 to microcontroller SDIO handoff

#### Note:

The 120 ns high state of SCLK is the minimum data hold time of the ADNS-2051. Since the falling edge of SCLK is actually the start of the next read or write command, the ADNS-2051 will hold the state of  $D_0$  on the SDIO line until the falling edge of SCLK. In both write and read operations, SCLK is driven by the micro-controller.

Serial port communications is not allowed while PD (power down) is high. See "Error Detection and Recovery" regarding resynchronizing via PD.

#### Forcing the SDIO Line to the Hi-Z State

There are times when the SDIO line from the ADNS-2051 should be in the Hi-Z state. If the microprocessor has completed a write to the ADNS-2051, the SDIO line is Hi-Z, since the SDIO pin is still configured as an input. However, if the last operation from the microprocessor was a read, the ADNS-2051 will hold the D0 state on SDIO until a falling edge of SCLK.

To place the SDIO pin into the Hi-Z state, raise the PD pin for 100  $\mu$ s (min). The PD pin can stay high, with the ADNS-2051 in the shutdown state, or the PD pin can be lowered, returning the ADNS-2051 to normal operation. The SDIO line will now be in the Hi-Z state.

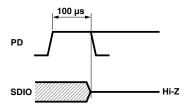


Figure 30. SDIO Hi-Z state and timing

#### Required Timing between Read and Write Commands (tsxx)

There are minimum timing requirements between read and write commands on the serial port.

If the rising edge of the SCLK for the last data bit of the

second write command occurs before the 100 microsecond required delay, then the first write command may not complete correctly.

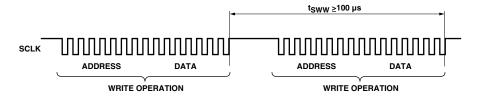


Figure 31. Timing between two write commands

If the rising edge of SCLK for the last address bit of the read command occurs before the 100 microsecond required delay, then the write command may not complete correctly.

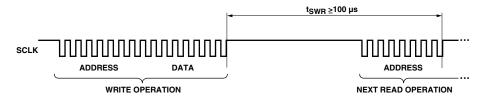


Figure 32. Timing between write and read commands

The falling edge of SCLK for the first address bit of either the read or write command must be at least 120 ns after the last SCLK rising edge of the last data bit of the previous read operation.

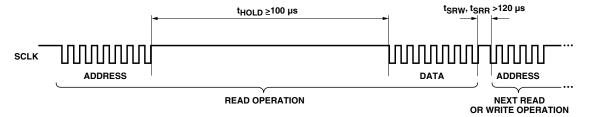


Figure 33. Timing between read and either write or subsequent read commands

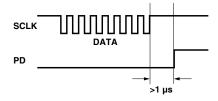


Figure 34. Timing between SCLK and PD rising edge

#### **Error Detection and Recovery**

- 1. The ADNS-2051 and the micro-controller might get out of synchronization due to ESD events, power supply droops or micro-controller firmware flaws. In such a case, the micro-controller should raise PD for 100  $\mu$ s. The ADNS-2051 will reset the serial port but will not reset the registers and be prepared for the beginning of a new transmission.
- 2. The ADNS-2051 has a transaction timer for the serial port. If the sixteenth SCLK rising edge is spaced more than approximately 0.9 seconds from the first SCLK edge of the current transaction, the serial port will reset.
- Invalid addresses:
  - Writing to an invalid address will have no effect.
     Reading from an invalid address will return all zeros.
- 4. Collision detection on SDIO
  - The only time that the ADNS-2051 drives the SDIO line is during a READ operation. To avoid data collisions, the micro-controller should relinquish SDIO before the falling edge of SCLK after the last address bit. The ADNS-2051 begins to drive SDIO after the next rising edge of SCLK. The ADNS-2051 relinquishes SDIO within 120 ns of the falling SCLK edge after the last data bit. The micro-controller can begin driving SDIO any time afterthat. In order to maintain low power consumption in nor mal operation or when the PD pin is pulled high, the micro-controller should not leave SDIO floating un til the next transmission (although that will not cause any communication difficulties).

- In case of synchronization failure, both the ADNS-2051 and the micro-controller may drive SDIO. The ADNS-2051 can withstand 30 mA of short circuit current and will withstand infinite duration short circuit conditions.
- 6. Termination of a transmission by the micro-controller may sometimes be required (for example, due to a USB suspend interrupt during a read operation). To accomplish this the micro-controller should raise PD. The ADNS-2051 will not write to any register and will reset the serial port (but nothing else) and be prepared for the beginning of future transmissions after PD goes low.
- 7. The micro-controller can verify success of write operations by issuing a read command to the same address and comparing written data to read data.
- 8. The micro-controller can verify the synchronization of the serial port by periodically reading the product ID register.

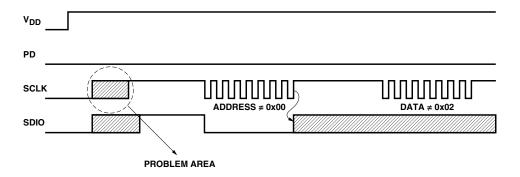


Figure 35. Power up serial port watchdog timer sequence

#### Notes on Power up and the Serial Port

The sequence in which  $V_{DD}$ , PD, SCLK, and SDIO are set during powerup can affect the operation of the serial port. The diagram below shows what can happen shortly after powerup when the microprocessor tries to read data from the serial port.

This diagram shows the  $V_{DD}$  rising to valid levels, at some point the microcontroller starts its program, sets the SCLK and SDIO lines to be outputs, and sets them high. It then waits to ensure that the ADNS-2051 has powered up and is ready to communicate. The microprocessor then tries to read from location 0x00, Product\_ID, and is expecting a value of 0x02. If it receives this value, it then knows that the communication to the ADNS-2051 is operational.

The problem occurs if the ADNS-2051 powers up before the microprocessor sets the SCLK and SDIO lines to be

outputs and high. The ADNS-2051 sees the raising of the SCLK as a valid rising edge, and clocks in the state of the SDIO as the first bit of the address (sets either a read or a write depending upon the state).

In the case of SDIO low, then a read operation has started. When the microprocessor begins to actually send the address, the ADNS-2051 already has the first bit of an address. When the seventh bit is sent by the micro, the ADNS-2051 has a valid address, and drives the SDIO line high within 120 ns (see detail "A" in Figure 27 and Figure 28). This results in a bus fight for SDIO. Since the address is wrong, the data sent back will be incorrect.

In the case of SDIO high, a write operation is started. The address and data are out of synchronization, and the wrong data will be written to the wrong address.

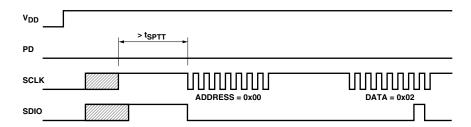


Figure 36. Power up serial port watchdog timer sequence

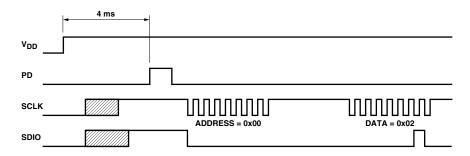


Figure 37. Power up serial port PD sync sequence

#### **Two Solutions**

There are two different ways to solve the problem, waiting for the serial port watchdog timer to time out, or using the PD line to reset the serial port.

#### 1. Serial Port Watchdog Timer Timeout

If the microprocessor waits at least  $t_{SPTT}$  from  $V_{DD}$  valid, it will ensure that the ADNS-2051 has powered up and the watchdog timer has timed out. This assumes that the microprocessor and the ADNS-2051 share the same power supply. If not, then the microprocessor must wait  $t_{SPTT}$  from ADNS-2051  $V_{DD}$  valid. Then when the SCLK toggles for the address, the ADNS-2051 will be in sync with the microprocessor.

### 2. PD Sync

The PD line can be used to resync the serial port. If the microprocessor waits for 4 ms from  $V_{DD}$  valid, and then outputs a valid PD pulse (see Figure 15), then the serial port will be ready for data.

#### **Resync Note**

If the microprocessor and the ADNS-2051 get out of sync, then the data either written or read from the registers will be incorrect. An easy way to solve this is to output a PD pulse to resync the parts after an incorrect read.

#### SPI communication code for the Cypress CY7C63000 or CY7C63001

(Please consult factory for the CY7C63722 or CY7C63723 codes.)

**Note:** This programming sequence is not covered in Avago's product warranty. It is only a recommended example when using the mentioned Cypress microcontrollers. For the latest updates on Cypress microcontrollers, please contact Cypress at email: usbapps@cypress.com or call (858) 613-7929 (US).

The following code can be used to implement the SPI data communications. See the schematic in Figure 9.

```
Notes:
   CY7C63001 20pin package
   ADNS-2051
   SDIO line connected to pin5 (P1.0)
   PD connected to pin 16 (P1.1)
   SCLK line connected to pin15 (P1.3)
   I/O port
Port1_Data:
                             01h
                                                      GPIO data port 1
                      equ
Port1_Interrupt:
                                                      Interrupt enable for port 1
                      equ
                             05h
Port1_Pullup:
                                                      Pullup resistor control for port 1
                      equ 09h
   Port bit definitions
SDIO:
                           01h
                                                   ; bit 0
PD:
                      equ 02h
                                                      bit 1
SCLK:
                      equ 08h
                                                      bit 3
Pt1_Current:
                      equ
                            00h
                                                   ; port1 current setting
   GPIO Isink registers
Port1_Isink:
                             38h
                      equ
Port1 Isink0:
                             38h
                      equ
Port1 Isink1:
                             39h
                      equ
Port1_Isink3:
                      equ
                             3Bh
   data memory variables
                             40h
                                                      address of spi writes
spi_addr:
                      equ
spi_data:
                             41h
                                                      data of spi writes
                      equ
bit_counter:
                      equ
                             44h
                                                      SPI bit counter
port1 wrote:
                      equ 45h
                                                      what we wrote last
   initialize Port 1
                             mov A, Pt1_Current ; select DAC setting
                      iowr Port1_Isink0 ; isink current Port 1 bit[0]
                      iowr Port1_Isink1
                                                      isink current Port 1 bit[1]
                      iowr Port1_Isink3
                                                      isink current Port 1 bit[3]
                      mov A, 0h
                                                      enable Port 1 bit [7:0] pullups
                      iowr Port1_Pullup
                      mov A, ~(PD|SDIO)
                                                   ; turn on the ADNS-2051
                      mov [port1_wrote], A
                      mov A, [port1_wrote]
                      iowr Port1 Data
                                                   ; PD low, SCLK, SDIO
                      mov A, 0
                      iowr Port1_Interrupt
                                                   ; disable port 1 interrupts
   There are possible problems with the SPI port if the microcontroller starts executing
   instructionsbefore the ADNS-2051 sensor has powered up. See page 18 for details.
   It is assumed that power to the microcontroller is OK if the next instructions can be executed.
   These instructions will reset the SPI port of the sensor.
```

mov A,~(SCLK|SDIO|PD) Resync\_sensor: ; set the SCLK, SDIO and PD lines low and [port1\_wrote], A mov A, [port1\_wrote] iowr Port1 Data ; If the power to the sensor needs more time ; to stabilize, insert a delay here ; wait about 4 milliseconds for the sensor call delay700us call delay700us ; oscillator to stabilize call delay700us call delay700us call delay700us call delay700us mov A, (SCLK|SDIO|PD ; set the SCLK, SDIO and PD lines high or [port1\_wrote], A ; this shuts down the oscillator and mov A, [port1\_wrote] ; resets the SPI port iowr Port1\_Data ; wait for the PD to reset the part call delay700us ; set the PD line low to put the sensor mov A, ~PD and [port1\_wrote], A ; back into normal operation mov A, [port1\_wrote] iowr Port1\_Data call delay700us wait about 4 milliseconds for the sensor call delay700us oscillator to stabilize call delay700us call delay700us call delay700us call delay700us ; sensor SPI port now in sync ReadSPI routine Includes delays for long traces or cables between the uP and ADNS-2051 Has correct timing of SCLK and SDIO spi\_addr = Address of SPI register in the ADNS-2051 On entry: spi data = undefined spi\_addr = undefined On exit spi data = register contents from ADNS-2051 ReadSPI: mov A. 64 ; wait 200us (optional) mov [bit\_counter], A ; (about 3us per loop) Waitrspi: nop nop nop nop nop nop dec [bit\_counter] jnz Waitrspi ; read address mov A,~80h and [spi\_addr], A lower MSB of address (read) call writeaddr wait 200us (about 3us per loop)(100us minimum required) mov A,64 wait for data to be ready mov [bit\_counter], A Waitrspi2: nop nop nop nop nop nop dec [bit\_counter] jnz Waitrspi2 mov A, 0h clear the data

```
mov [spi_data], A
                       mov A, 08h
                       mov [bit_counter], A
                       mov A, SDIO
                       or [port1 wrote], A
                       mov A, [port1_wrote]
                                                    ; write a 1 to SDIO
                       iowr Port1_Data
nextr:
                       mov A, ~SCLK
                                                    ; lower SCLK
                       and [port1_wrote], A
                       mov A, [port1_wrote]
                       iowr Port1_Data
                                                      wait for cable to settle
                       nop
                                                      if ADNS-2051 is connected to
                       nop
                       nop
                                                      IC via short PCB traces,
                                                      then the number of NOPs can
                       nop
                       nop
                                                      reduced or eliminated
                       nop
                       nop
                       mov A,[spi_data]
                                                    ; shift next bit
                                                      shift next bit
                       mov [spi_data], A
                       iord Port1 Data
                                                      read SDIO
                       and A, SDIO
                       jz rdx
rd1:
                       mov A, 01h
                       or [spi_data], A
                       mov A, SCLK
                                                    ; raise SCLK
rdx:
                       or [port1_wrote], A
                       mov A, [port1 wrote]
                       iowr Port1_Data
                                                    ; wait for cable to settle
                       nop
                       nop
                       nop
                       nop
                       nop
                       nop
                       nop
                       dec [bit_counter]
                       inz nextr
                       ret
   WriteSPI routine
   Includes delays for long traces or cables between the uP and ADNS-2051.
; Has correct timing of SCLK and SDIO
                       spi_addr = Address of SPI register in the ADNS-2051
   spi_data = Data to be written to the SPI register
   On exit
                       spi_addr = undefined
   spi data = undefined
WriteSPI:
                       mov A, 64
                                                    ; wait 200us (optional)
                       mov [bit_counter], A
                                                    ; about 3us per loop
Waitspi:
                       nop
                       nop
                       nop
                       nop
                       nop
                       nop
                       dec [bit_counter]
                       jnz Waitspi
                                                    ; write address
```

mov A, 80h or [spi\_addr], A ; set MSB of address (write) call writeaddr jmp wrdata writeaddr: mov A, 08h 8 bits to shift out mov [bit\_counter], A mov A, ~SCLK ; lower SCLK nexta: and [port1\_wrote], A mov A, [port1\_wrote] iowr Port1\_Data mov A, [spi\_addr] ; shift next bit mov [spi\_addr], A inc addr0 addr1: mov A, SDIO or [port1\_wrote], A ; raise SDIO jmp addrx addr0: mov A, ~SDIO and [port1\_wrote], A lower SDIO addrx: mov A, [port1\_wrote] iowr Port1\_Data ; wait for cable to settle nop nop nop nop nop nop nop mov A, SCLK ; raise SCLK or [port1\_wrote], A mov A, [port1\_wrote] iowr Port1\_Data ADNS-2051 reads the address bit nop wait for cable to settle nop nop nop nop nop nop dec [bit\_counter] jnz nexta ret wrdata: mov A, 08h ; 8 bits of data mov [bit\_counter], A mov A, ~SCLK ; lower SCLK nextw: and [port1\_wrote], A mov A, [port1\_wrote] iowr Port1\_Data ; shift next bit mov A, [spi\_data] mov [spi\_data], A jnc wr0 wr1: mov A, SDIO or [port1\_wrote], A ; raise SDIO jmp wrx wr0: mov A, ~SDIO lower SDIO and [port1\_wrote], A mov A, [port1\_wrote] wrx: iowr Port1\_Data ; wait for cable to settle nop nop nop nop nop nop nop

```
mov A, SCLK
                                                   ; raise SCLK
                      or [port1_wrote], A
                      mov A, [port1_wrote]
                      iowr Port1_Data
                                                   ; ADNS-2051 reads the data bit
                      nop
                                                     wait for cable to settle
                      nop
                      nop
                      nop
                      nop
                      nop
                      dec [bit_counter]
                      jnz nextw
                      ret
delay700us:
                                                  ; wait for 710us
                      mov A, ffh
                      mov [bit_counter], A
                                                   ; reuse bit_counter
waitd0:
                      nop
                      nop
                      nop
                      nop
                      nop
                      nop
                                                   ; 2us
                      dec [bit_counter]
                      jnz waitd0
Example calling syntax
   WriteSPI
   Set register 0a to 40h, LED blink mode
                                                  ; move address into A
                      mov A, 0ah
                      mov [spi_addr], A
                                                 ; move address into spi_addr
                                                  ; move data into A
                      mov A, 40h
                      mov [spi_data], A
                                                  ; move data into spi_data
                      call WriteSPI
                                                     call WriteSPI routine, on return,
                                                   ; spi_addr and spi_data will be undefined
   ReadSPI
   Read register 02h, the motion register
                      mov A, 02h
                                                   ; move address into A
                      mov [spi_addr], A
                                                  ; move address into spi_addr
                      call ReadSPI
                                                  ; call ReadSPI, on return, data is in spi_data,
                                                   ; spi_addr is undefined
```

#### Registers

The ADNS-2051 can be programmed through registers, via the serial port, and configuration and motion data can be read from these registers.

Address	Register
0x00	Product_ID
0x01	Revision_ID
0x02	Motion
0x03	Delta_X
0x04	Delta_Y
0x05	SQUAL

Address	Register				
0x06	Average_Pixel				
0x07	Maximum_Pixel				
0x08	Reserved				
0x09	Reserved				
0x0a	Configuration_bits				
0x0b	Reserved				

Address	Register
0x0c	Data_Out_Lower
0x0d	Data_Out_Upper
0x0e	Shutter_Lower
0x0f	Shutter_Upper
0x10	Frame_Period_Lower
0x11	Frame_Period_Upper

Product_ID Access: Reac	k				Address: 0x00 Reset Value: 0x02					
Bit	7	6	5	4	3	2	1	0		
Field	PID <sub>7</sub>	PID <sub>6</sub>	PID <sub>5</sub>	PID <sub>4</sub>	PID <sub>3</sub>	PID <sub>2</sub>	PID <sub>1</sub>	PID <sub>0</sub>		

Data Type: Eight bit number with the product identifier.

USAGE: The value in this register does not change, it can be used to verify that the serial communications link is OK.

Revision_ID Access: Reac	ł				ress: 0x0 et Value:				
Bit 7 6 5 4 3 2 1 0									
Field	RID <sub>7</sub>	RID <sub>6</sub>	RID <sub>5</sub>	RID <sub>4</sub>	RID <sub>3</sub>	RID <sub>2</sub>	RID <sub>1</sub>	RID <sub>0</sub>	

Data Type: Eight bit number with current revision of the IC.

USAGE: NN is a value between 00 and FF which represent the current design revision of the device.

Motion											
Access: Reac	Reset Value: 0x00										
Bit	7	7 6 5 4 3 2 1									
Field	MOT	Reserved	FAULT	OVFY	OVFX	Reserved	Reserved	RES			

Data Type: Bit field

USAGE: Register 0x02 allows the user to determine if motion has occurred since the last time it was read. If so, then the user should read registers 0x03 and 0x04 to get the accumulated motion. It also tells if the motion buffers have overflowed and whether or not an LED fault occurred since the last reading. The current resolution is also shown.

Field Name	Description
MOT	Motion since last report or PD
	0 = No motion
	1 = Motion occurred, data ready for reading in Delta_X and Delta_Y registers
Reserved	Reserved for future
FAULT	LED Fault detected – set when R_BIN is too low or too high, shorts to V <sub>DD</sub> or Ground
	0 = No fault
	1 = Fault detected
OVFY	Motion overflow Y, $\Delta Y$ buffer has overflowed since last report
	0 = No overflow
	1 = Overflow has occurred
OVFX	Motion overflow X, $\Delta X$ buffer has overflowed since last report
	0 = No overflow
	1 = Overflow has occurred
Reserved	Reserved for future
Reserved	Reserved for future
RES	Resolution in counts per inch
	0 = 400
	1 = 800

#### **Notes for Motion:**

- 1. Reading this register freezes the Delta\_X and Delta\_Y register values. Read this register before reading the Delta\_X and Delta\_Y registers. If Delta\_X and Delta\_Y are not read before the motion register is read a second time, the data in Delta\_X and Delta\_Y will be lost.
- 2. Avago RECOMMENDS that registers 0x02, 0x03 and 0x04 be read sequentially.
- 3. Internal buffers can accumulate more than eight bits of motion for X or Y. If either one of the internal buffers overflows, then absolute path data is lost, and the OVFX or OVFY bit is set. These bits (OVFX and OVFY) are cleared once some motion has been read from the Delta\_X and Delta\_Y registers, and if the buffers are not at full scale. Since more data is present in the buffers, the cycle of reading the Motion, Delta\_X and Delta\_Y registers should be repeated until the motion bit (MOT) is cleared. Until MOT is cleared, either the Delta\_X or Delta\_Y registers will read either positive or negative full scale. If the motion register has not been read for a long time, at 400 cpi it may take up to 16 read cycles to clear the buffers, at 800 cpi, up to 32 cycles.
- 4. FAULT is a sticky bit that is cleared by reading the Motion register. It signifies that an LED fault has occurred since the last time the motion register was read. Once an LED fault has cleared, the hardware will drive the LED normally.

Delta_X				Add	ress: 0x0	)3					
Access: Read	ł	Reset Value: 0x00									
Bit	7	6	5	4	3	2	1	0			
Field	X <sub>7</sub>	X <sub>6</sub>	X <sub>5</sub>	X <sub>4</sub>	Х3	X <sub>2</sub>	X <sub>1</sub>	X <sub>0</sub>			

Data Type: Eight bit 2's complement number.

USAGE: X movement is counts since last report. Absolute value is determined by resolution. Reading clears the register.

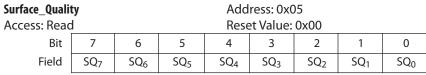


Delta Y Address: 0x04 Access: Read Reset Value: 0x00 7 5 Bit 6 4 1 0 Field  $Y_7$  $Y_6$  $Y_5$  $Y_4$  $Y_3$  $Y_2$  $Y_1$  $Y_0$ 

Data Type: Eight bit 2's complement number.

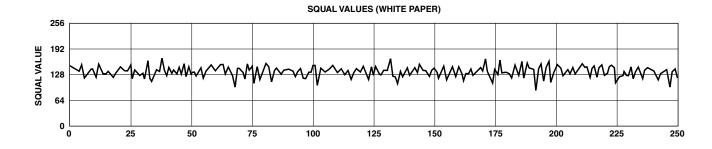
USAGE: Y movement is counts since last report. Absolute value is determined by resolution. Reading clears the register.





Data Type: Eight bit number.

USAGE: SQUAL is a measure of the number of features visible by the sensor in the current frame. The maximum value is 255. Since small changes in the current frame can result in changes in SQUAL, variations in SQUAL when looking at a surface are expected. The graph below shows 250 sequentially acquired SQUAL values, while a sensor was moved slowly over white paper. SQUAL is nearly equal to zero, if there is no surface below the sensor.



The focus point is important and could affect the squal value, the graph below showing another setup with various z-height. The graph clearly shows that the squal count is dependent on focus distance.

#### Note:

This graph is obtained by getting multiple readings over different heights.

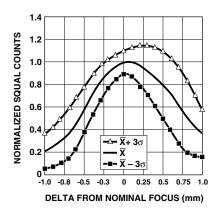
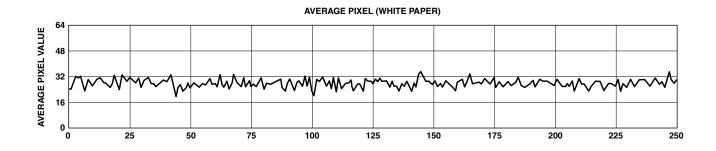


Figure 38. Typical mean squal vs. Z (white paper)

Average_Pixel				Add	ress: 0x0	)6		
Access: Reac	k			Rese	et Value:	0x00		
Bit	7	6	5	4	3	2	1	0
Field	0	0	AP <sub>5</sub>	AP <sub>4</sub>	AP <sub>3</sub>	AP <sub>2</sub>	AP <sub>1</sub>	AP <sub>0</sub>

Data Type: Six bit number.

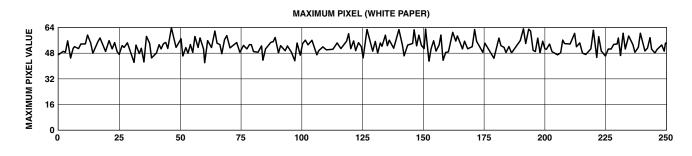
USAGE: Average Pixel value in current frame. Minimum value = 0, maximum = 63. The average pixel value can be adjusted every frame. Shown below is a graph of 250 sequentially acquired average pixel values, while the sensor was moved slowly over white paper.



Maximum_Pixel Address: 0x07										
Access: Read Reset Value: 0x00										
Bit	7	6	5	4	3	2	1	0		
Field	0	0	MP <sub>5</sub>	MP <sub>4</sub>	MP <sub>3</sub>	MP <sub>2</sub>	MP <sub>1</sub>	MP <sub>0</sub>		

Data Type: Six bit number.

USAGE: Maximum Pixel value in current frame. Minimum value = 0, maximum value = 63. The maximum pixel value can be adjusted every frame. Shown below is a graph of 250 sequentially acquired maximum pixel values, while the sensor was moved slowly over white paper.



**Reserved** Address: 0x08 **Reserved** Address: 0x09

# ${\bf Configuration\_bits}$

Address: 0x0a Reset Value: 0x00

Access: Read/Write

Bit	7	6	5	4	3	2	1	0
Field	RESET	LED_MODE	Sys Test	RES	PixDump	Reserved	Reserved	Sleep

Data Type: Bit field

USAGE: Register 0x0a allows the user to change the configuration of the sensor. Shown below are the bits, their default values, and optional values.

Field Name	Description
RESET	Power up defaults (bit always reads 0)
	0 = No effect
	1 = Reset registers and bits to power up default settings (bold entries)
LED_MODE	LED Shutter Mode
	0 = Shutter mode off (LED always on) (even if no motion up to 1 sec.)
	1 = Shutter mode on (LED only on when the electronic shutter is open)
Sys Test	System Tests (bit always reads 0)
	0 = No tests
	1 = perform all system tests, output 16 bit CRC via Data_Out_Upper and Data_Out_Lower registers.
	Note: Since part of the system test is a RAM test, the RAM will be overwritten with the default values
	when the test is done. If any configuration changes from the default are needed for operation, make
	the changes AFTER the system test is run. This operation requires substantially more time to
	complete than other register transactions.
RES	Resolution in counts per inch
	0 = 400
	1 = 800
Pix Dump	Dump the pixel array through Data_Out_Upper and Data_Out_Lower, 256 bytes
	0 = disabled
	1 = dump pixel array
Reserved	Reserved
Reserved	Reserved
Sleep	Sleep Mode
	0 = Normal, fall asleep after one second of no movement (1500 frames/s)
	1 = Always awake

**Reserved** Address: 0x0b

Data_Out_Lower Address: 0x0c										
Access: Read	ł	Reset Value: undefined								
Bit 7 6 5 4 3 2 1										
Field	DO <sub>7</sub>	DO <sub>6</sub>	DO <sub>5</sub>	DO <sub>4</sub>	DO <sub>3</sub>	DO <sub>2</sub>	DO <sub>1</sub>	DO <sub>0</sub>		

Data_Out_Upper Address: 0x0d										
Access: Read Reset Value: undefined										
Bit	7	6	5	4	3	2	1	0		
Field DO <sub>15</sub> DO <sub>14</sub> DO <sub>13</sub> DO <sub>12</sub> DO <sub>11</sub> DO <sub>10</sub> DO <sub>9</sub>										

Data Type: Sixteen bit word.

USAGE: Data can be written to these registers from the system self test, or the pixel dump command. The data can be read out 0x0d, or 0x0d first, then 0x0c.

	Data_Out_Upper	Data_Out_Lower	Note
System test result 1:	FE	D4	One of two results returned. These
System test result 2:	4D	10	values are subject to change with
			each device design revision.
Pixel Dump command	Pixel Address	Pixel Data (Lower 6 bits)	

Once the pixel dump command is given, the sensor writes the address and the value for the first pixel into the Data\_Out\_Upper and Data\_Out\_Lower registers. The MSB of Data\_Out\_Lower is the status bit for the data. If the bit is high, the data are NOT valid. Once the MSB is low, the data for that particular read are valid and should be saved. The pixel address and data will then be incremented on the next frame. Once the pixel dump is complete, the PixDump bit in register 0x0a should be set to zero. To obtain an accurate image, the LED needs to be turned on by changing the sleep mode of the configuration register 0x0a to always awake.

**Pixel Address Map** (Looking through the HDNS-2100 Lens)

#### LAST PIXEL

FF	EF	DF	CF	BF	AF	9F	8F	7F	6F	5F	4F	3F	2F	1F	0F
FE	EE	DE	CE	BE	AE	9E	8E	7E	6E	5E	4E	3E	2E	1E	0E
FD	ED	DD	CD	BD	AD	9D	8D	7D	6D	5D	4D	3D	2D	1D	0D
FC	EC	DC	СС	вс	AC	9C	8C	7C	6C	5C	4C	зС	2C	1C	0C
FB	ЕВ	DB	СВ	вв	АВ	9B	8B	7B	6B	5B	4B	3В	2B	1B	0В
FA	EA	DA	CA	ВА	АА	9A	8A	7A	6A	5A	4A	3 <b>A</b>	2A	1A	0А
F9	E9	D9	C9	В9	Α9	99	89	79	69	59	49	39	29	19	09
F8	E8	D8	C8	B8	A8	98	88	78	68	58	48	38	28	18	08
F7	E7	D7	<b>C</b> 7	В7	Α7	97	87	77	67	57	47	37	27	17	07
F6	<b>E</b> 6	D6	C6	В6	<b>A</b> 6	96	86	76	66	56	46	36	26	16	06
F5	<b>E</b> 5	D5	<b>C</b> 5	<b>B</b> 5	<b>A</b> 5	95	85	75	65	55	45	35	25	15	05
F4	E4	D4	C4	В4	<b>A</b> 4	94	84	74	64	54	44	34	24	14	04
F3	<b>E</b> 3	DЗ	СЗ	В3	А3	93	83	73	63	53	43	33	23	13	03
F2	E2	D2	C2	B2	<b>A2</b>	92	82	72	62	52	42	32	22	12	02
F1	E1	D1	C1	В1	<b>A</b> 1	91	81	71	61	51	41	31	21	11	01
F0	E0	D0	C0	В0	Α0	90	80	70	60	50	40	30	20	10	00

FIRST PIXEL

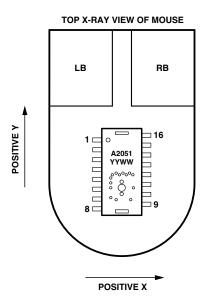
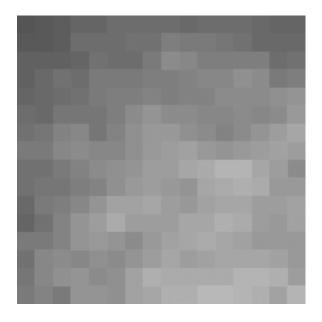
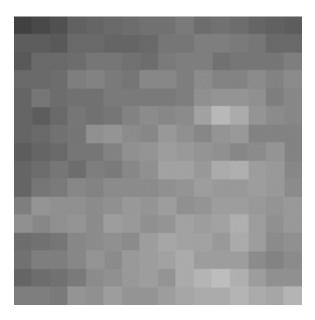


Figure 39. Directions are for a complete mouse, with the HDNS-2100 lens

## **Pixel Dump Pictures**

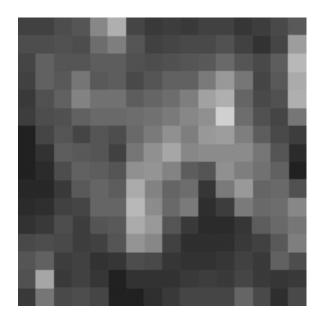
The following images (Figure 40) are the output of the pixel dump command. The data ranges from zero for complete black, to 63 for complete white. An internal AGC circuit adjusts the shutter value to keep the brightest feature (max. pixel) in the mid 50s.





(a) White Paper

(b) Manila Folder



(c) Neoprene Mouse Pad (Gray)

(d) USAF Test Chart Group 3, Element 1 8 line pairs per mm

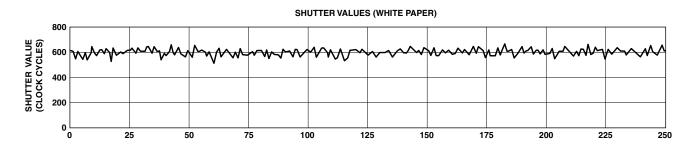
Figure 40. Pixel dump pictures

Shutter_Lowe		Address: 0x0e						
Access: Read				Rese	et Value:	0x64		
Bit	7	6	5	4	3	2	1	0
Field	S <sub>7</sub>	S <sub>6</sub>	S <sub>5</sub>	S <sub>4</sub>	S <sub>3</sub>	S <sub>2</sub>	S <sub>1</sub>	S <sub>0</sub>

Shutter_Uppe Access: Read			Address: 0x0f Reset Value: 0x00					
Bit	7	6	5	4	3	2	1	0
Field	S <sub>15</sub>	S <sub>14</sub>	S <sub>13</sub>	S <sub>12</sub>	S <sub>11</sub>	S <sub>10</sub>	S <sub>9</sub>	S <sub>8</sub>

Data Type: Sixteen bit word.

USAGE: Units are clock cycles; default value is 64. Read Shutter\_Upper first, then Shutter\_Lower. They should be read consecutively. The shutter is adjusted to keep the average and maximum pixel values within normal operating ranges. The shutter value can be adjusted to a new value on every frame. When the shutter adjusts, it changes by  $\pm 1/16$  of the current value. Shown below is a graph of 250 sequentially acquired shutter values, while the sensor was moved slowly over white paper.



The focus point is important and could affect the shutter value. The graph below shows another setup with various z-height. This graph clearly shows that the shutter value is dependent on focus distance.

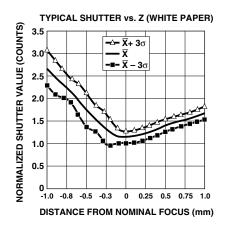


Figure 41. Typical shutter vs. Z (white paper)

**Note:** This graph shows average readings over different heights.

The maximum value of the shutter is dependent upon the frame rate and clock frequency. The formula for the maximum shutter value is:

Max. Shutter Value = 
$$\frac{\text{Clock Frequency}}{\text{Frame Rate}}$$
 -2816

For a clock frequency of 18 MHz, the following table shows the maximum shutter value. 1 clock cycle is 55.56 nsec.

	Max Shutter		Shutter		
Frames/second	Decimal	Hex	Upper	Lower	
2300	5010	0x1392	13	92	
2000	6184	0x1828	18	28	
1500	9184	0x23E0	23	E0	
1000	15184	0x3B50	3B	50	
500	33184	0x81A0	81	A0	

← Default Max. Shutter

Frame_Period	Lower							
Access: Read		Reset Value: 0x20						
Bit	7	6	5	4	3	2	1	0
Field	FP <sub>7</sub>	FP <sub>6</sub>	FP <sub>5</sub>	FP <sub>4</sub>	FP <sub>3</sub>	FP <sub>2</sub>	FP <sub>1</sub>	FP <sub>0</sub>

Frame_Period	Address: 0x11							
Access: Reac			Rese	et Value:	0xd1			
Bit	7	6	5	4	3	2	1	0
Field	FP <sub>15</sub>	FP <sub>14</sub>	FP <sub>13</sub>	FP <sub>12</sub>	FP <sub>11</sub>	FP <sub>10</sub>	FP <sub>9</sub>	FP <sub>8</sub>

Data Type: Sixteen bit 2's complement word.

USAGE: The frame period counter counts up until it overflows. Units are clock cycles. The formula is:

 $\frac{\text{Clock Rate}}{\text{Frame Rate}} = \text{Counts (decimal)} \rightarrow \text{Counts (hex)} \rightarrow \text{Counts (2's complement hex)}$ 

For an 18 MHz clock, here are the Frame\_Period values for popular frame rates.

	Counts		Frame_Pe	Frame_Period		
Frames/second	Decimal	Нех	2's Comp	Upper	Lower	
2300*	7826	0x1E92	0xE16E	E1	6E	_
2000*	9000	0x2328	0xDCD8	DC	D8	– –  ← Defa
1500	12000	0x2EE0	0xD120	D1	20	— ← Deia
1000	18000	0x4650	0xB9B0	B9	В0	– ← Minir
500	36000	0x8CA0	0x7360	73	60	_

← Default Frame Time

← Minimum Frame Time

#### \*Note:

To optimize tracking performance on dark surfaces, it is recommended that an adaptive frame rate based on shutter value be implemented, for frame rates greater than 1500. Changing the frame rate results in changes in the maximum speed, acceleration limits, and dark surface performance.

To read from the registers, read Frame\_Period\_Upper first followed by Frame\_Period\_Lower. To write to the registers, write Frame\_Period\_Lower first followed by Frame\_Period\_Upper.

# IC Register State after Reset (power up or setting bit 7, register 0x0a)

Address	Register	Value	Meaning
0x0	Product_ID	0x02	Product ID = 2 (Fixed value)
0x01	Revision_ID	0xNN	Revision of IC (Fixed value). (For each device design revision.)
0x02	Motion	0x00	No Motion
			LED = No Fault
			No X data overflow
			No Y data overflow
			Resolution is 400 counts per inch
0x03	Delta_X	0x00	No X motion
0x04	Delta_Y	0x00	No Y motion
0x05	SQUAL	0x00	No image yet to measure
0x06	Average_Pixel	0x00	No image yet to measure
0x07	Maximum_Pixel	0x00	No image yet to measure
0x08	Reserved	_	
0x09	Reserved	_	
0x0a	Configuration_bits	0x00	Part is not Reset
			LED Shutter Mode is off
			No System tests
			Resolution = 400 counts per inch
			Pixel Dump is disabled
			Sleep mode is enabled
0x0b	Reserved	_	
0x0c	Data_Out_Lower	undefined	No data to read
0x0d	Data_Out_Upper	undefined	No data to read
0x0e	Shutter_Lower	0x64	Initial shutter value
0x0f	Shutter_Upper	0x00	Initial shutter value
0x10	Frame_Period_Lower	0x20	Initial frame period value (corresponds to 1500 fps)
0x11	Frame_Period_Upper	0xd1	Initial frame period value (corresponds to 1500 fps)

## **Optical Mouse Design References**

**Application Note AN1179** 

Eye Safety calculation AN1228

### **Ordering Information**

Specify part number as follows:

ADNS-2051 = Sensor IC in a 16-pin staggered DIP, 20 per tube.

HDNS-2100 = Round Optical Mouse Lens

HDNS-2100#001 = Trimmed Optical Mouse Lens

HDNS-2200 = LED Assembly Clip (Black)

HDNS-2200#001 = LED Assembly Clip (Clear)

HLMP-ED80-XX000 = LED

