

# Data Sheet

## HAL<sup>®</sup> 401

Linear Hall-Effect Sensor IC

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## Linear Hall Effect Sensor IC in CMOS technology

**Release Notes:** Revision bars indicate significant changes to the previous edition.

### 1. Introduction

The HAL 401 is a Linear Hall Effect Sensors produced in CMOS technology. The sensor includes a temperature-compensated Hall plate with choppered offset compensation, two linear output stages, and protection devices (see Fig. 2–1).

The output voltage is proportional to the magnetic flux density through the hall plate. The choppered offset compensation leads to stable magnetic characteristics over supply voltage and temperature.

The HAL 401 can be used for magnetic field measurements, current measurements, and detection of any mechanical movement. Very accurate angle measurements or distance measurements can also be done. The sensor is very robust and can be used in electrical and mechanical hostile environments.

The sensor is designed for industrial and automotive applications and operates in the ambient temperature range from  $-40\text{ }^{\circ}\text{C}$  up to  $150\text{ }^{\circ}\text{C}$  and is available in the SMD-package SOT89B-1.

#### 1.1. Features:

- switching offset compensation at 147 kHz
- low magnetic offset
- extremely sensitive
- operates from 4.8 to 12 V supply voltage
- wide temperature range  $T_A = -40\text{ }^{\circ}\text{C}$  to  $+150\text{ }^{\circ}\text{C}$
- overvoltage protection
- reverse voltage protection of  $V_{DD}$ -pin
- differential output
- accurate absolute measurements of DC and low frequency magnetic fields
- on-chip temperature compensation

### 1.2. Marking Code

Type	Temperature Range	
	A	K
HAL401	401A	401K

### 1.3. Operating Junction Temperature Range

The Hall sensors from Micronas are specified to the chip temperature (junction temperature  $T_J$ ).

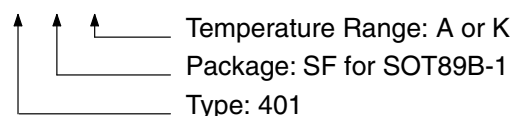
**A:**  $T_J = -40\text{ }^{\circ}\text{C}$  to  $+170\text{ }^{\circ}\text{C}$

**K:**  $T_J = -40\text{ }^{\circ}\text{C}$  to  $+140\text{ }^{\circ}\text{C}$

**Note:** Due to power dissipation, there is a difference between the ambient temperature ( $T_A$ ) and junction temperature. Please refer to section 4.1. on page 18 for details.

### 1.4. Hall Sensor Package Codes

HALXXXPA-T



Example: **HAL401SF-K**

→ Type: 401

→ Package: SOT89B-1

→ Temperature Range:  $T_J = -40\text{ }^{\circ}\text{C}$  to  $+140\text{ }^{\circ}\text{C}$

Hall sensors are available in a wide variety of packaging versions and quantities. For more detailed information, please refer to the brochure: “Hall Sensors: Ordering Codes, Packaging, Handling”.

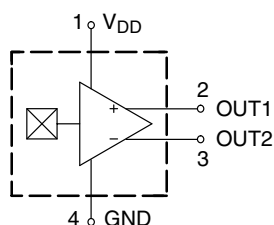
## 1.5. Solderability and Welding

### Soldering

During soldering reflow processing and manual reworking, a component body temperature of 260 °C should not be exceeded.

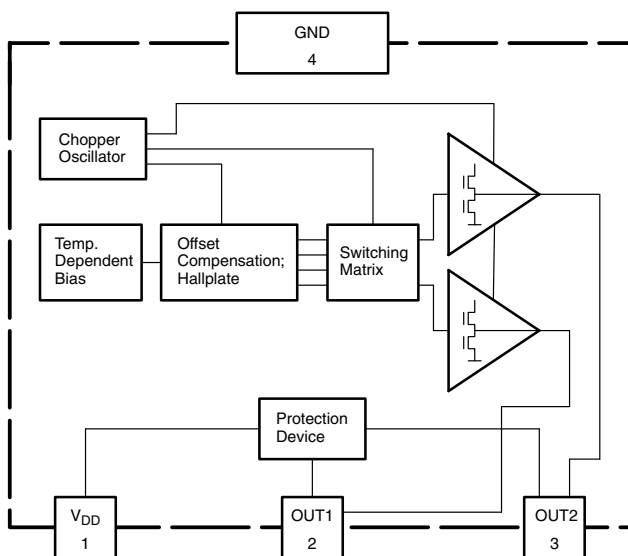
### Welding

Device terminals should be compatible with laser and resistance welding. Please note that the success of the welding process is subject to different welding parameters which will vary according to the welding technique used. A very close control of the welding parameters is absolutely necessary in order to reach satisfying results. Micronas, therefore, does not give any implied or express warranty as to the ability to weld the component.



**Fig. 1-1:** Pin configuration

## 2. Functional Description



**Fig. 2-1:** Block diagram of the HAL401 (top view)

The Linear Hall Sensor measures constant and low frequency magnetic flux densities accurately. The differential output voltage  $V_{OUTDIF}$  (difference of the voltages on pin 2 and pin 3) is proportional to the magnetic flux density passing vertically through the sensitive area of the chip. The common mode voltage  $V_{CM}$  (average of the voltages on pin 2 and pin 3) of the differential output amplifier is a constant 2.2 V.

The differential output voltage consists of two components due to the switching offset compensation technique. The average of the differential output voltage represents the magnetic flux density. This component is overlaid by a differential AC signal at a typical frequency of 147 kHz. The AC signal represents the internal offset voltages of amplifiers and hall plates that are influenced by mechanical stress and temperature cycling.

External filtering or integrating measurement can be done to eliminate the AC component of the signal. Resultingly, the influence of mechanical stress and temperature cycling is suppressed. No adjustment of magnetic offset is needed.

The sensitivity is stabilized over a wide range of temperature and supply voltage due to internal voltage regulation and circuits for temperature compensation.

### Offset Compensation (see Fig. 2-2)

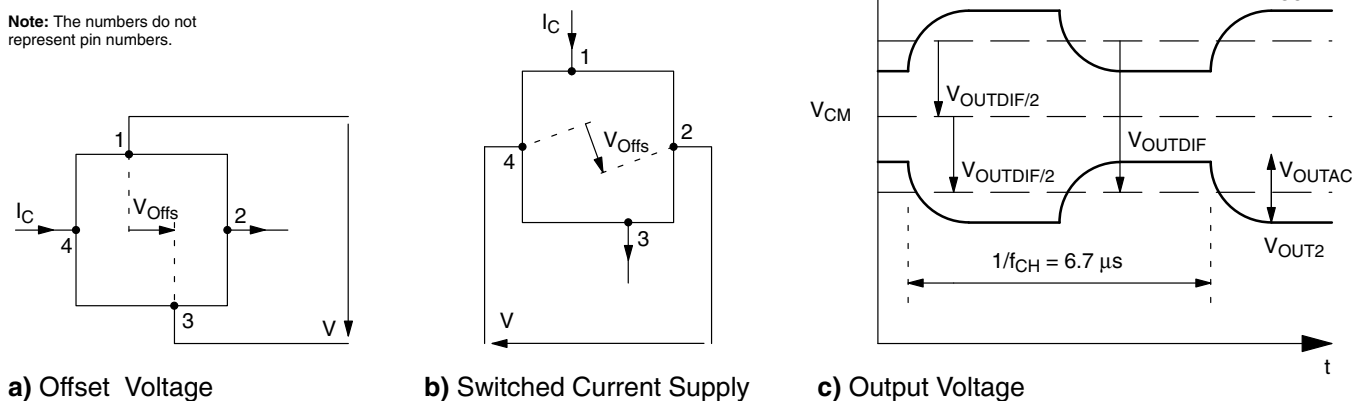
The Hall Offset Voltage is the residual voltage measured in absence of a magnetic field (zero-field residual voltage). This voltage is caused by mechanical stress and can be modeled by a displacement of the connections for voltage measurement and/or current supply.

Compensation of this kind of offset is done by cyclic commutating the connections for current flow and voltage measurement.

- First cycle:  
The hall supply current flows between points 4 and 2. In the absence of a magnetic field,  $V_{13}$  is the Hall Offset Voltage ( $+V_{Offs}$ ). In case of a magnetic field,  $V_{13}$  is the sum of the Hall voltage ( $V_H$ ) and  $V_{Offs}$ .  
 $V_{13} = V_H + V_{Offs}$
- Second cycle:  
The hall supply current flows between points 1 and 3. In the absence of a magnetic field,  $V_{24}$  is the Hall Offset Voltage with negative polarity ( $-V_{Offs}$ ). In case of a magnetic field,  $V_{24}$  is the difference of the Hall voltage ( $V_H$ ) and  $V_{Offs}$ .  
 $V_{24} = V_H - V_{Offs}$

In the first cycle, the output shows the sum of the Hall voltage and the offset; in the second, the difference of both. The difference of the mean values of  $V_{OUT1}$  and  $V_{OUT2}$  ( $V_{OUTDIF}$ ) is equivalent to  $V_{Hall}$ .

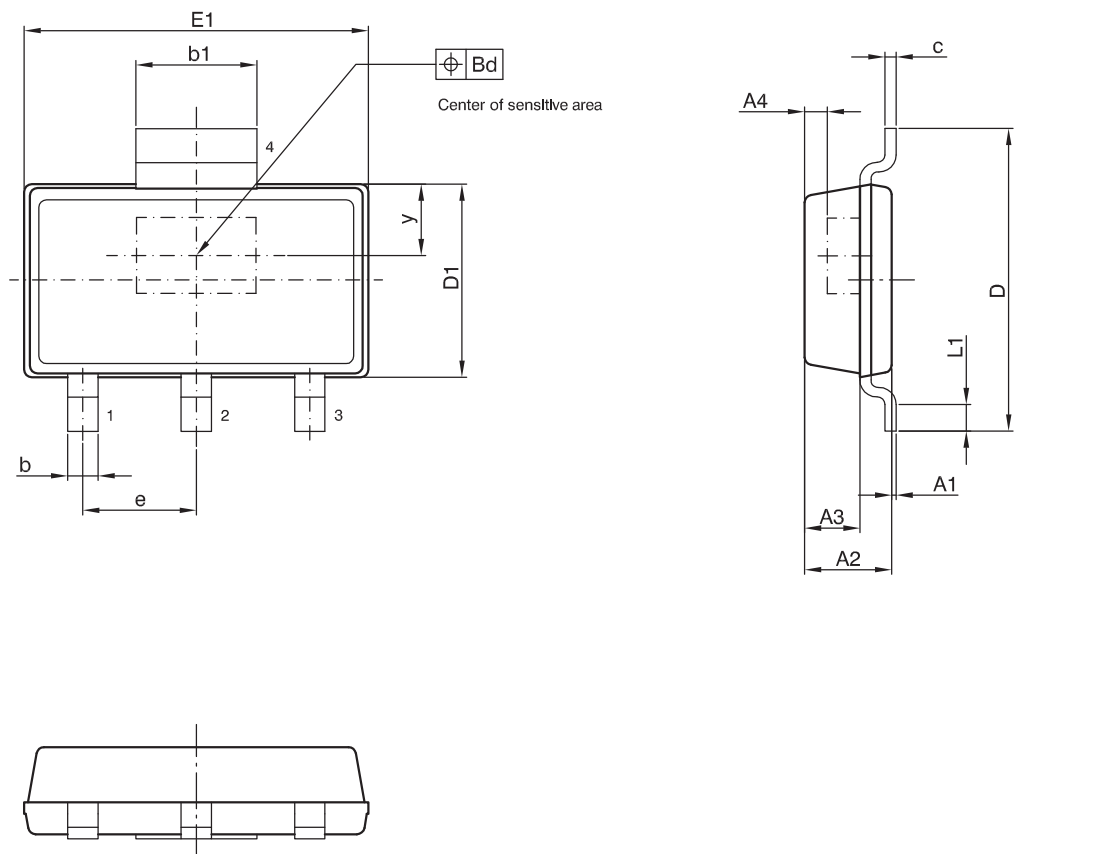
**Note:** The numbers do not represent pin numbers.



**Fig. 2-2:** Hall Offset Compensation

3. Specifications

3.1. Outline Dimensions



physical dimensions do not include moldflash.  
A4, y = these dimensions are different for each sensor type and are specified in the data sheet.

UNIT	A1	A2	A3	b	b1	Bd	c	D	D1	e	E1	L1
mm	0.10 0.02	1.20 1.10	0.73	0.4	1.7	0.2	0.15	4.0	2.6 2.5	1.5	4.6 4.5	0.25 min.

JEDEC STANDARD		ANSI	ISSUE DATE YY-MM-DD	DRAWING-NO.	ZG-NO.
ISSUE	ITEM NO.				
-	-		07-07-02	06610.0001.4	ZG001010_Ver.03

**Fig. 3-1:**  
**SOT89B-1:** Plastic Small Outline Transistor package, 4 leads  
Weight approximately 0.034 g

### 3.2. Dimensions of Sensitive Area

0.37 mm x 0.17 mm

### 3.3. Position of Sensitive Area

	<b>SOT89B-1</b>
y	0.95 mm nominal

### 3.4. Absolute Maximum Ratings

Stresses beyond those listed in the “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only. Functional operation of the device at these conditions is not implied. Exposure to absolute maximum rating conditions for extended periods will affect device reliability.

This device contains circuitry to protect the inputs and outputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid application of any voltage higher than absolute maximum-rated voltages to this circuit.

All voltages listed are referenced to ground (GND).

Symbol	Parameter	Pin No.	Min.	Max.	Unit
$V_{DD}$	Supply Voltage	1	−12	12	V
$V_O$	Output Voltage	2, 3	−0.3	12	V
$I_O$	Continuous Output Current	2, 3	−5	5	mA
$T_J$	Junction Temperature Range		−40	170	°C
$T_A$	Ambient Temperature at $V_{DD} = 5\text{ V}$ at $V_{DD} = 12\text{ V}$		− −	150 125	°C °C

#### 3.4.1. Storage and Shelf Life

The permissible storage time (shelf life) of the sensors is unlimited, provided the sensors are stored at a maximum of 30 °C and a maximum of 85% relative humidity. At these conditions, no Dry Pack is required.

Solderability is guaranteed for one year from the date code on the package.

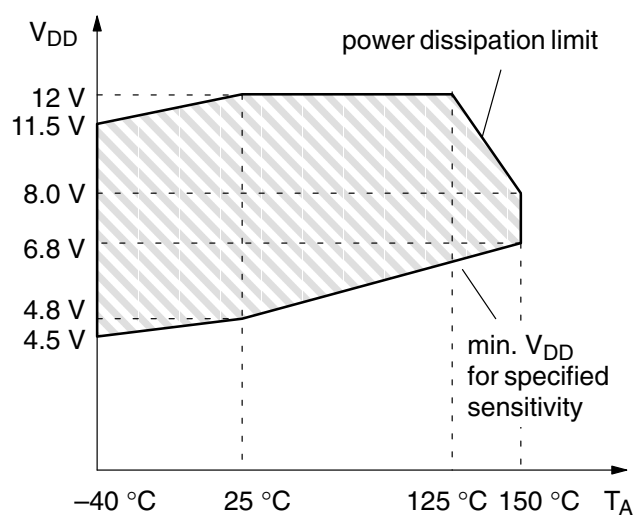


### 3.5. Recommended Operating Conditions

Functional operation of the device beyond those indicated in the “Recommended Operating Conditions” of this specification is not implied, may result in unpredictable behavior of the device and may reduce reliability and lifetime.

All voltages listed are referenced to ground (GND).

Symbol	Parameter	Pin No.	Min.	Max.	Unit	Remarks
$I_O$	Continuous Output Current	2, 3	-2.25	2.25	mA	$T_J = 25\text{ }^\circ\text{C}$
$I_O$	Continuous Output Current	2, 3	-1	1	mA	$T_J = 170\text{ }^\circ\text{C}$
$C_L$	Load Capacitance	2, 3	—	1	nF	
$V_{DD}$	Supply Voltage	1	4.8	12	V	see Fig. 3-2
B	Magnetic Field Range		-50	50	mT	

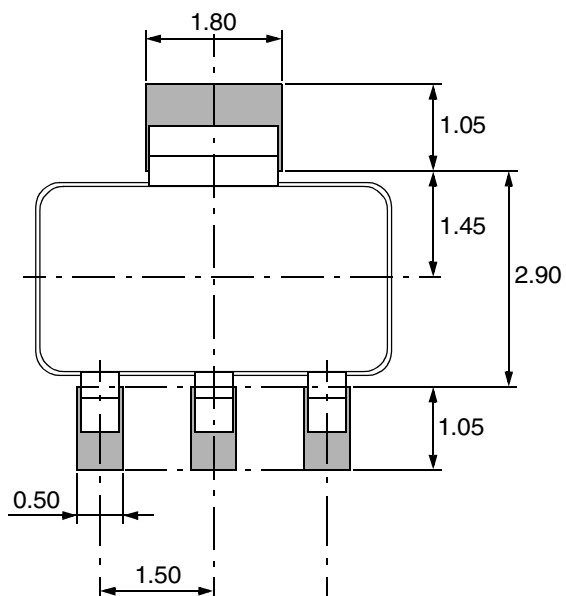


**Fig. 3-2:** Recommended Operating Supply Voltage

**3.6. Characteristics** at  $T_J = -40\text{ }^{\circ}\text{C}$  to  $+170\text{ }^{\circ}\text{C}$ ,  $V_{DD} = 4.8\text{ V}$  to  $12\text{ V}$ ,  $GND = 0\text{ V}$   
at Recommended Operation Conditions (Fig. 3–2 for  $T_A$  and  $V_{DD}$ ) as not otherwise specified in the column “Conditions”.  
Typical characteristics for  $T_J = 25\text{ }^{\circ}\text{C}$ ,  $V_{DD} = 6.8\text{ V}$  and  $-50\text{ mT} < B < 50\text{ mT}$

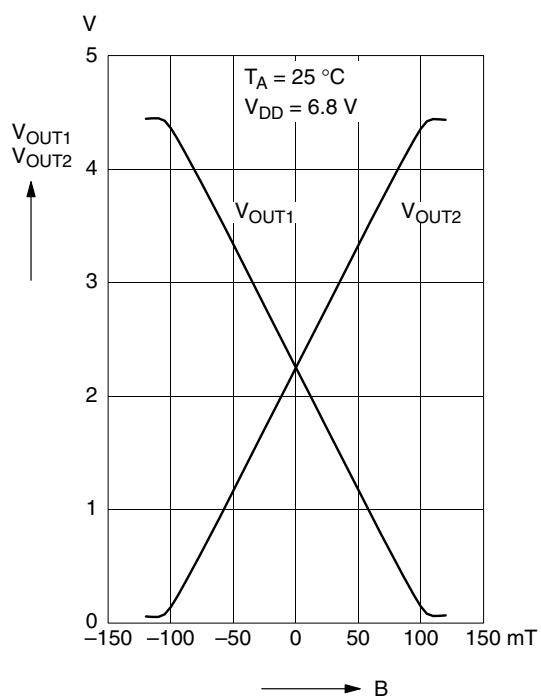
Symbol	Parameter	Pin No.	Min.	Typ.	Max.	Unit	Conditions
$I_{DD}$	Supply Current	1	11	14.5	17.1	mA	$T_J = 25\text{ }^{\circ}\text{C}$ , $I_{OUT1,2} = 0\text{ mA}$
$I_{DD}$	Supply Current over Temperature Range	1	9	14.5	18.5	mA	$I_{OUT1,2} = 0\text{ mA}$
$V_{CM}$	Common Mode Output Voltage $V_{CM} = (V_{OUT1} + V_{OUT2}) / 2$	2, 3	2.1	2.2	2.3	V	$I_{OUT1,2} = 0\text{ mA}$ ,
CMRR	Common Mode Rejection Ratio	2, 3	-2.5	0	2.5	mV/V	$I_{OUT1,2} = 0\text{ mA}$ , CMRR is limited by the influence of power dissipation.
$S_B$	Differential Magnetic Sensitivity	2–3	42	48.5	55	mV/mT	$-50\text{ mT} < B < 50\text{ mT}$ $T_J = 25\text{ }^{\circ}\text{C}$
$S_B$	Differential Magnetic Sensitivity over Temperature Range	2–3	37.5	46.5	55	mV/mT	$-50\text{ mT} < B < 50\text{ mT}$
$B_{offset}$	Magnetic Offset over Temperature	2–3	-1.5	-0.2	1.5	mT	$B = 0\text{ mT}$ , $I_{OUT1,2} = 0\text{ mA}$
$\Delta B_{OFFSET} / \Delta T$	Magnetic Offset Change		-25	0	25	$\mu\text{T/K}$	$B = 0\text{ mT}$ , $I_{OUT1,2} = 0\text{ mA}$
BW	Bandwidth (-3 dB)	2–3	–	10	–	kHz	without external Filter <sup>1)</sup>
$NL_{dif}$	Non-Linearity of Differential Output	2–3	–	0.5	2	%	$-50\text{ mT} < B < 50\text{ mT}$
$NL_{single}$	Non-Linearity of Single Ended Output	2, 3	–	2	–	%	
$f_{CH}$	Chopper Frequency over Temp.	2, 3	–	147	–	kHz	
$V_{OUTACpp}$	Peak-to-Peak AC Output Voltage	2, 3	–	0.6	1.3	V	
$n_{meff}$	Magnetic RMS Differential Broadband Noise	2–3	–	10	–	$\mu\text{T}$	$BW = 10\text{ Hz}$ to $10\text{ kHz}$
$f_{Cflicker}$	Corner Frequency of 1/f Noise	2–3	–	10	–	Hz	$B = 0\text{ mT}$
$f_{Cflicker}$	Corner Frequency of 1/f Noise	2–3	–	100	–	Hz	$B = 50\text{ mT}$
$R_{OUT}$	Output Impedance	2, 3	–	30	50	$\Omega$	$I_{OUT1,2} \leq 2.5\text{ mA}$ , $T_J = 25\text{ }^{\circ}\text{C}$ , $V_{DD} = 6.8\text{ V}$
$R_{OUT}$	Output Impedance over Temperature	2, 3	–	30	150	$\Omega$	$I_{OUT1,2} \leq 2.5\text{ mA}$
$R_{thJSB}$ case SOT89B-1	Thermal Resistance Junction to Substrate Backside		–	150	200	K/W	Fiberglass Substrate 30 mm x 10 mm x 1.5 mm pad size (see Fig. 3–3)

<sup>1)</sup> with external 2 pole filter ( $f_{3db} = 5\text{ kHz}$ ),  $V_{OUTAC}$  is reduced to less than 1 mV by limiting the bandwidth

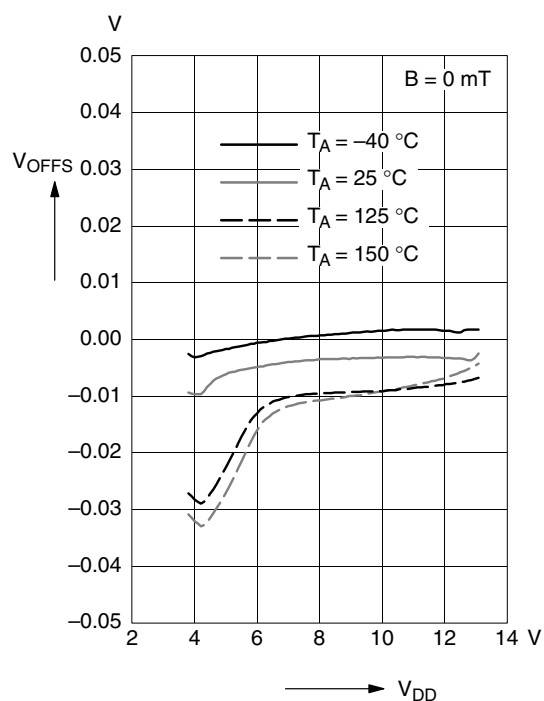


**Fig. 3-3:** Recommended footprint SOT89B,  
Dimensions in mm

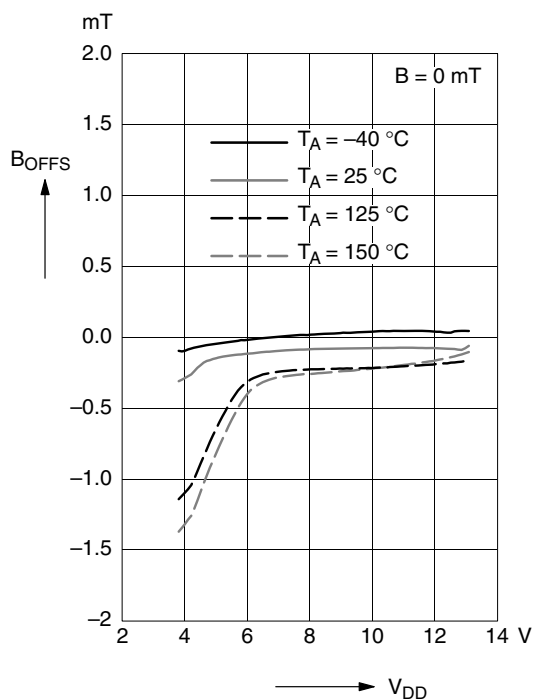
**Note:** All dimensions are for reference only. The pad size may vary depending on the requirements of the soldering process.



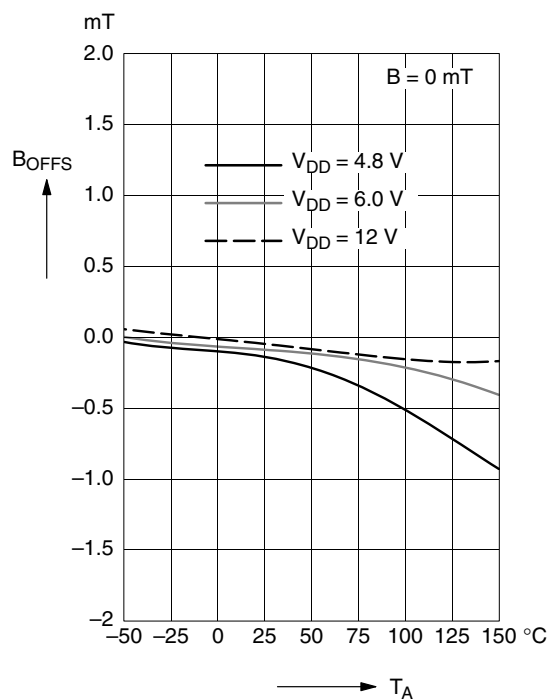
**Fig. 3-4:** Typical output voltages versus magnetic flux density



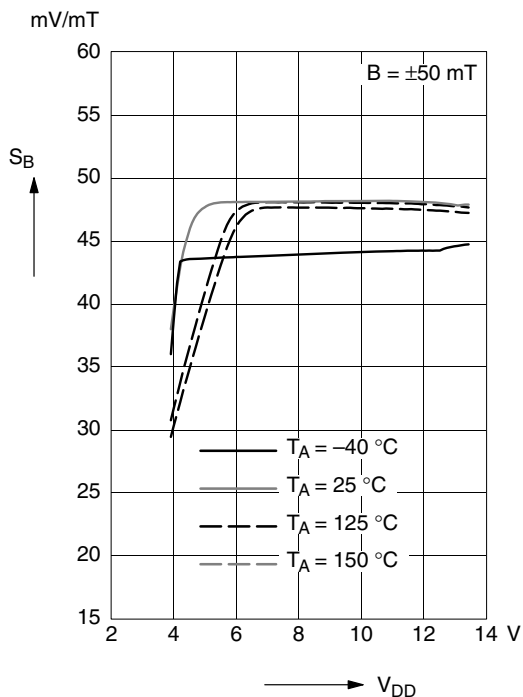
**Fig. 3-6:** Typical differential output offset voltage versus supply voltage



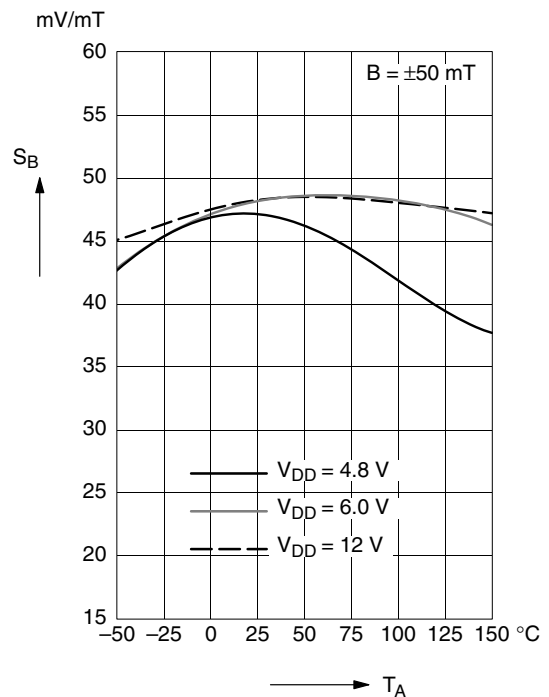
**Fig. 3-5:** Typical magnetic offset of differential output versus supply voltage



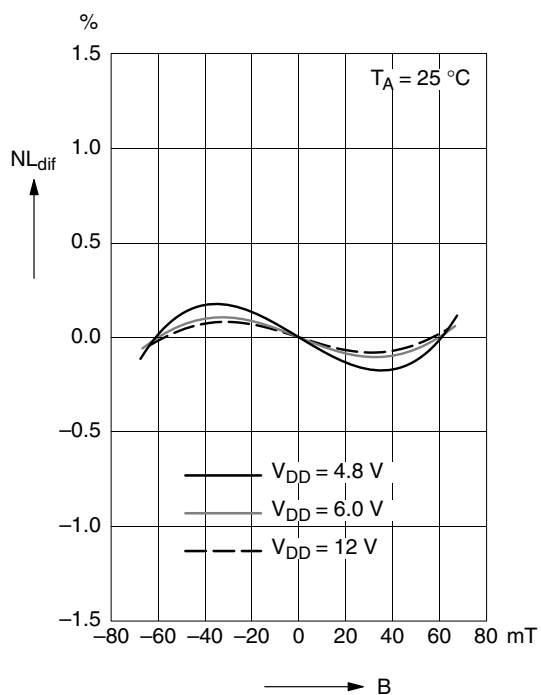
**Fig. 3-7:** Typical magnetic offset of differential output versus ambient temperature



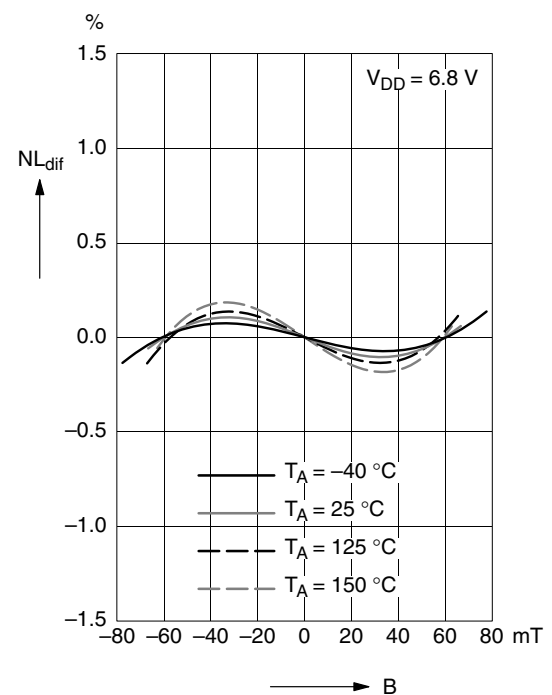
**Fig. 3-8:** Typical differential magnetic sensitivity versus supply voltage



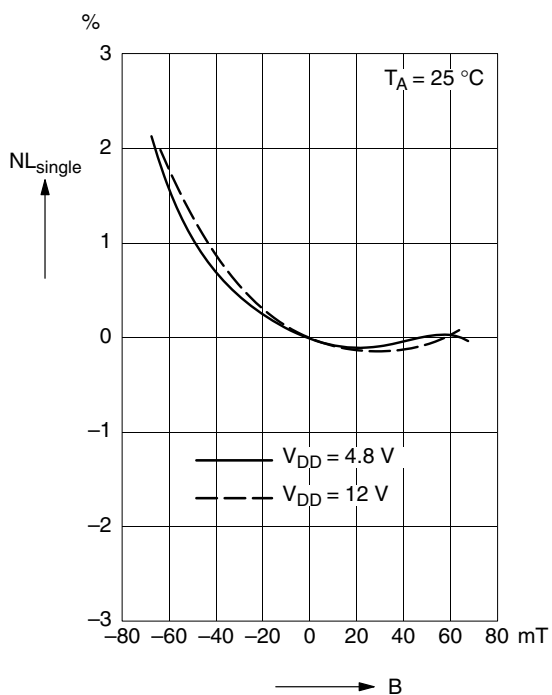
**Fig. 3-10:** Typical differential magnetic sensitivity versus ambient temperature



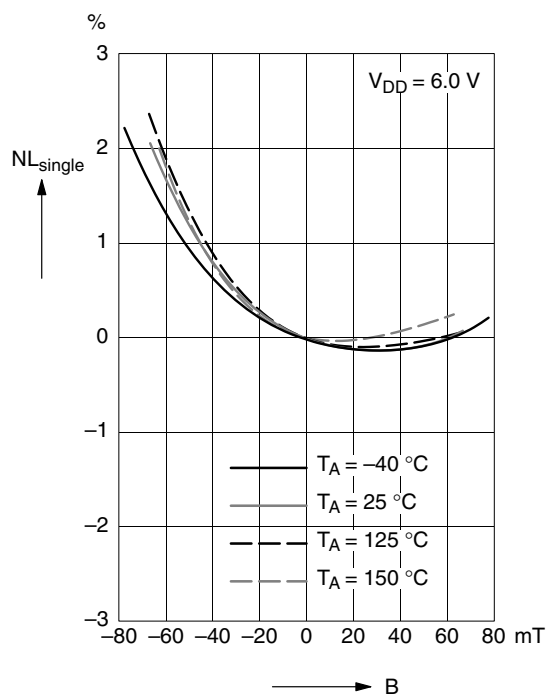
**Fig. 3-9:** Typical non-linearity of differential output versus magnetic flux density



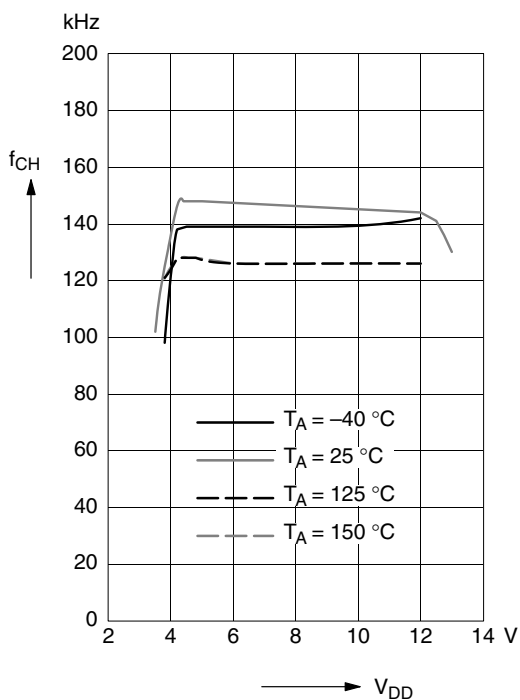
**Fig. 3-11:** Typical non-linearity of differential output versus magnetic flux density



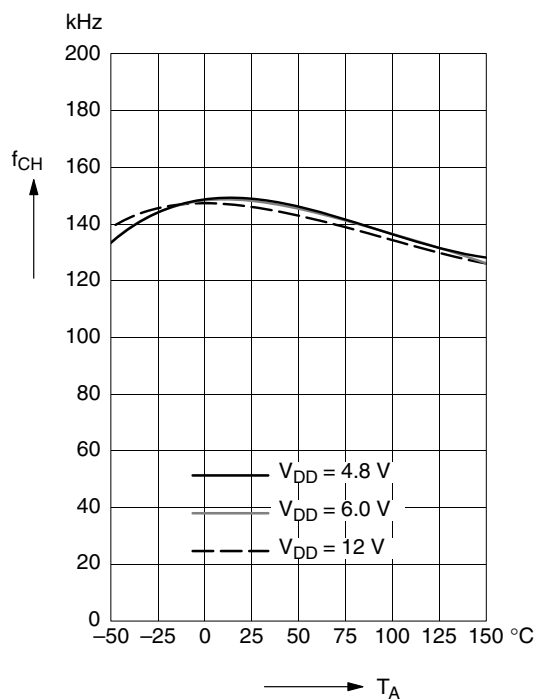
**Fig. 3-12:** Typical single-ended non-linearity versus magnetic flux density



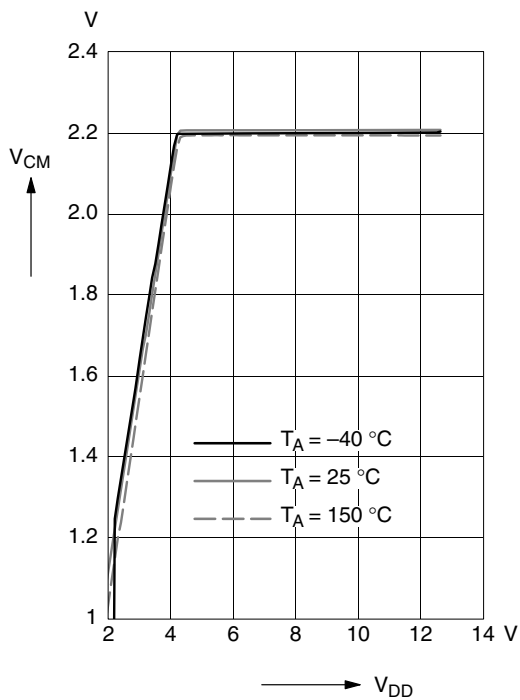
**Fig. 3-14:** Typical non-linearity of single-ended output versus magnetic flux density



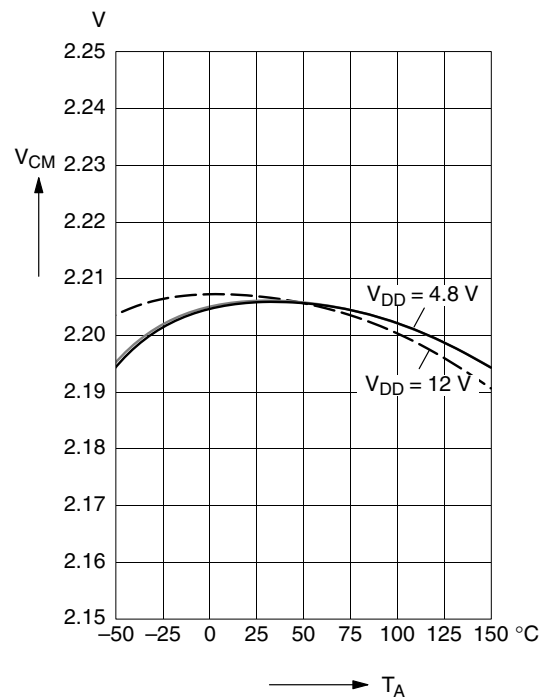
**Fig. 3-13:** Typical chopper frequency versus supply voltage



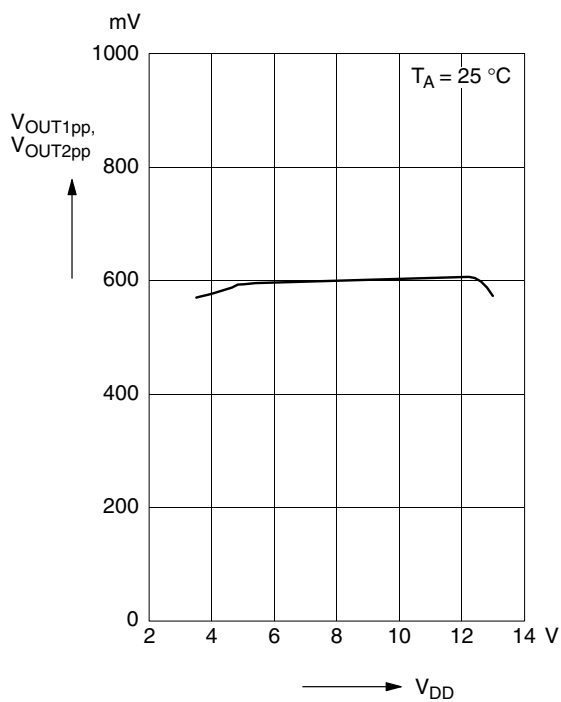
**Fig. 3-15:** Typical chopper frequency versus ambient temperature



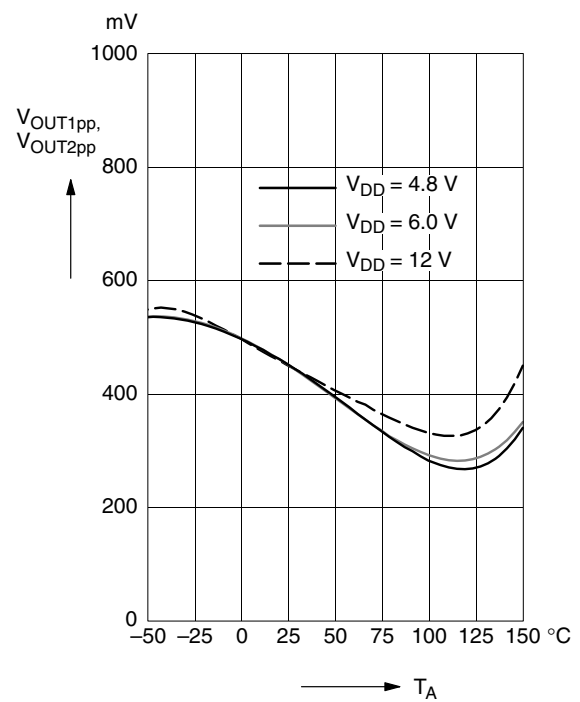
**Fig. 3-16:** Typical common mode output voltage versus supply voltage



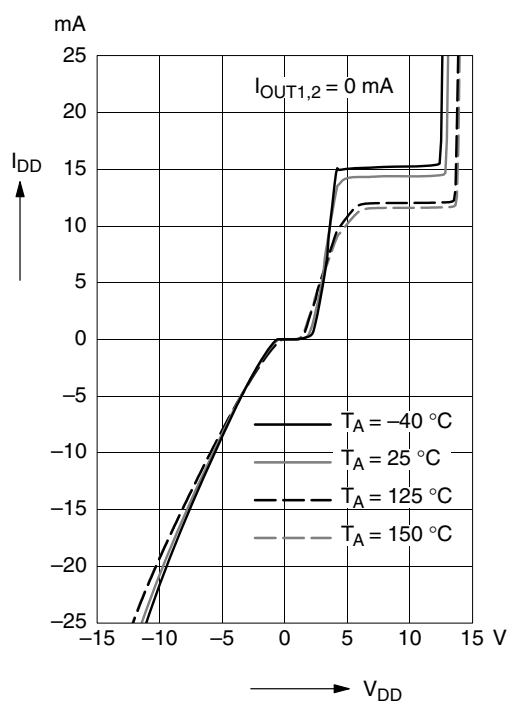
**Fig. 3-18:** Typical common mode output voltage versus ambient temperature



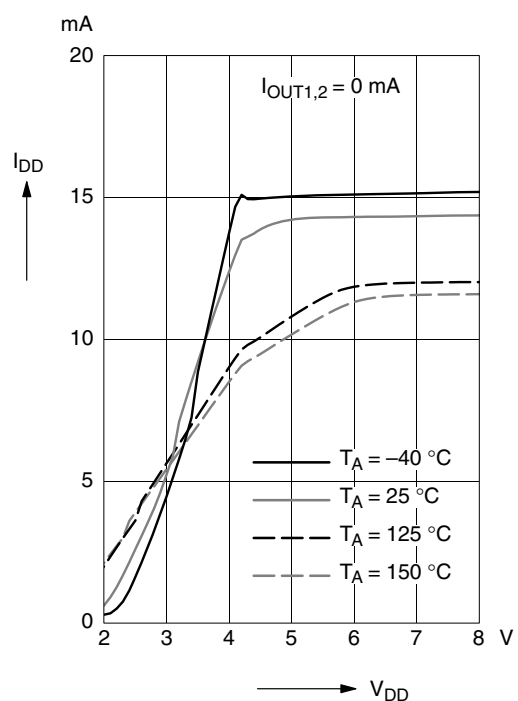
**Fig. 3-17:** Typical output AC voltage versus supply voltage



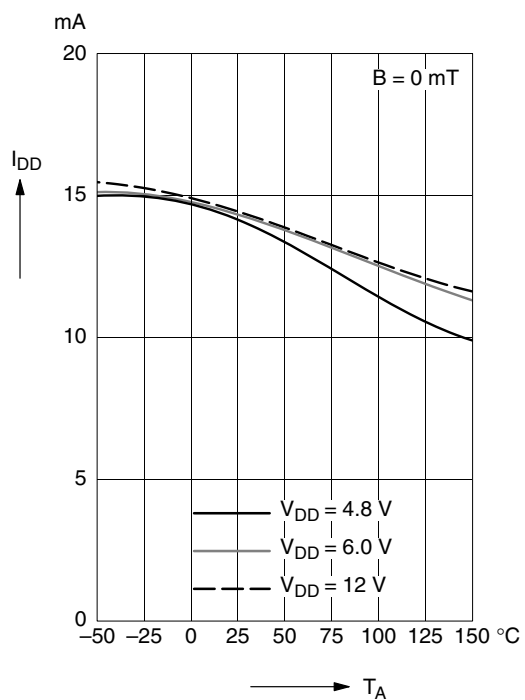
**Fig. 3-19:** Typical output AC voltage versus ambient temperature



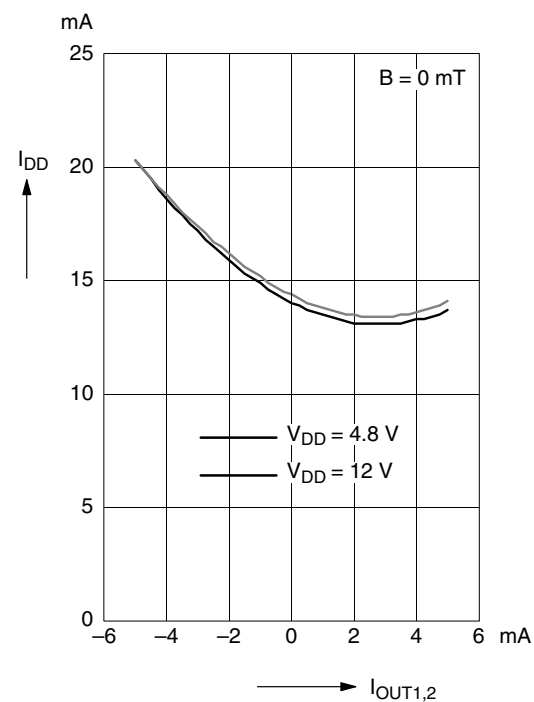
**Fig. 3-20:** Typical supply current versus supply voltage



**Fig. 3-22:** Typical supply current versus supply voltage

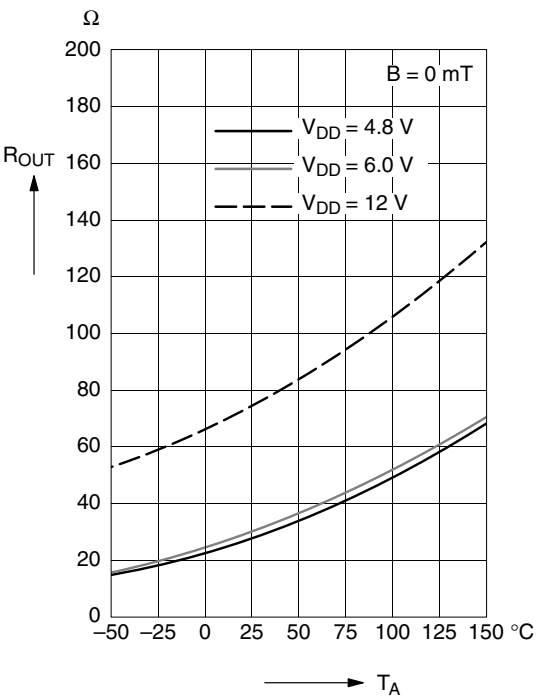


**Fig. 3-21:** Typical supply current versus temperature

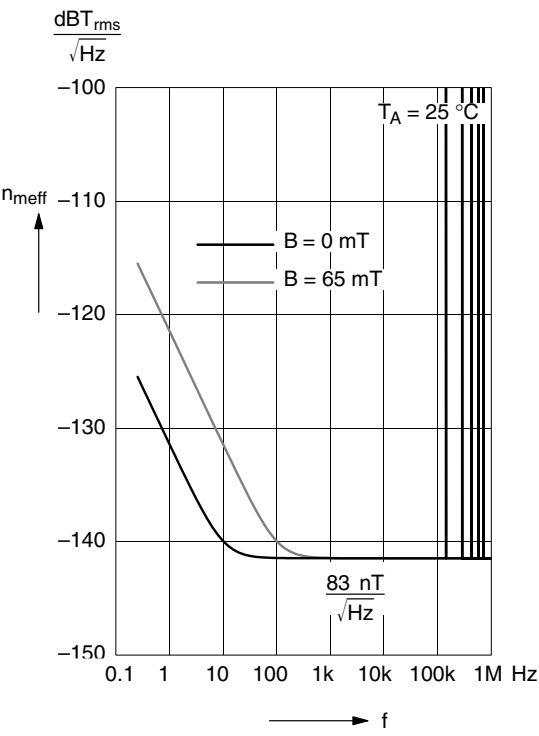


**Fig. 3-23:** Typical supply current versus output current

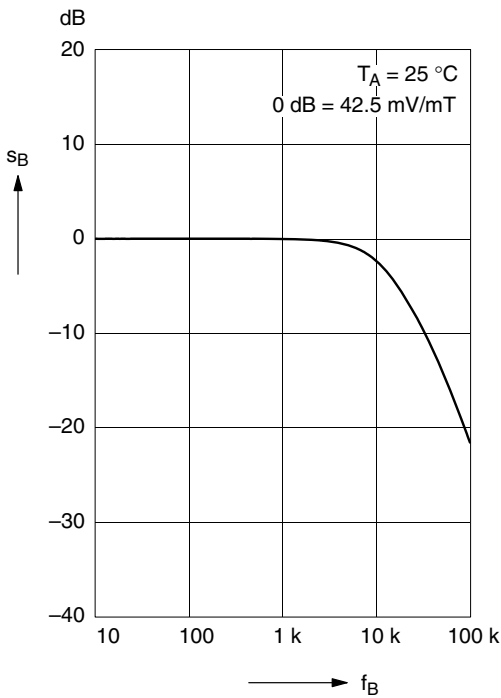




**Fig. 3–24:** Typical dynamic differential output resistance versus temperature



**Fig. 3–26:** Typical magnetic noise spectrum



**Fig. 3–25:** Typical magnetic frequency response

## 4. Application Notes

Mechanical stress on the device surface (caused by the package of the sensor module or overmolding) can influence the sensor performance.

The parameter  $V_{OUTACpp}$  (see Fig. 2–2) increases with external mechanical stress. This can cause linearity errors at the limits of the recommended operation conditions.

### 4.1. Ambient Temperature

Due to internal power dissipation, the temperature on the silicon chip (junction temperature  $T_J$ ) is higher than the temperature outside the package (ambient temperature  $T_A$ ).

$$T_J = T_A + \Delta T$$

At static conditions and continuous operation, the following equation applies:

$$\Delta T = I_{DD} * V_{DD} * R_{thJSB}$$

For all sensors, the junction temperature range  $T_J$  is specified. The maximum ambient temperature  $T_{Amax}$  can be calculated as:

$$T_{Amax} = T_{Jmax} - \Delta T$$

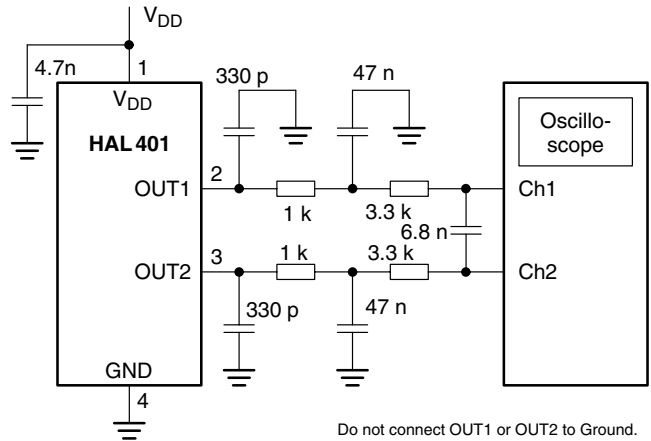
For typical values, use the typical parameters. For worst case calculation, use the max. parameters for  $I_{DD}$  and  $R_{th}$ , and the max. value for  $V_{DD}$  from the application.

### 4.2. EMC and ESD

Please contact Micronas for detailed information on EMC and ESD results.

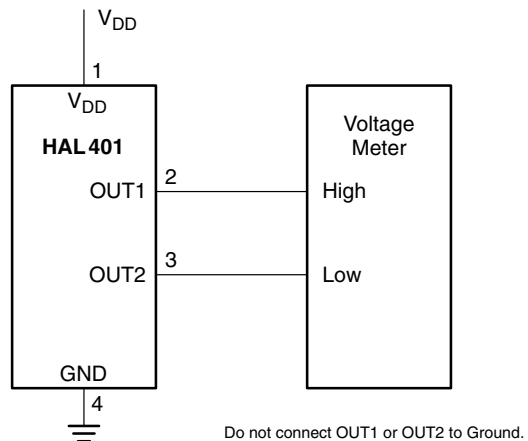
## 4.3. Application Circuit

The normal integrating characteristics of a voltmeter is sufficient for signal filtering.

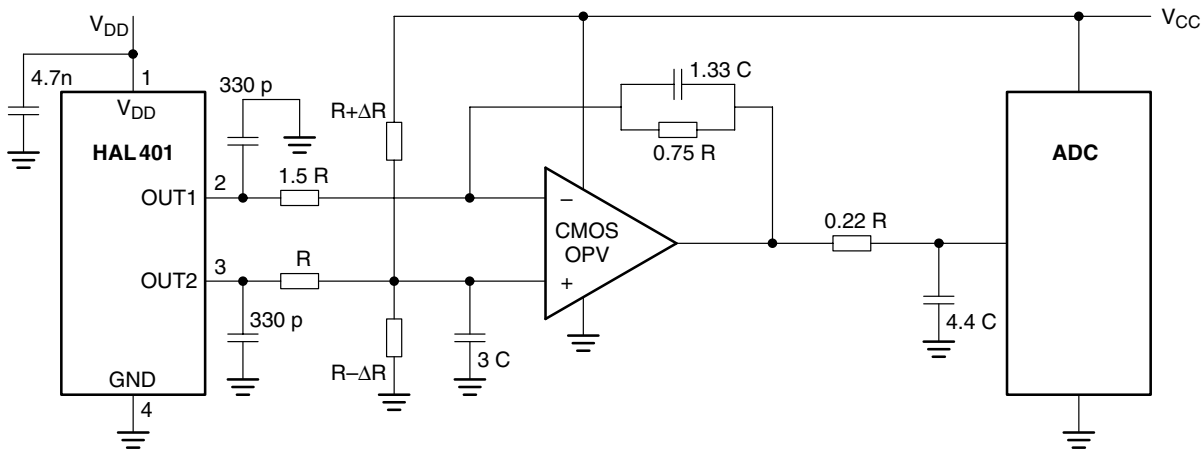


**Fig. 4–1:** Filtering of output signals

Display the difference between channel 1 and channel 2 to show the Hall voltage. Capacitors 4.7 nF and 330 pF for electromagnetic immunity are recommended.

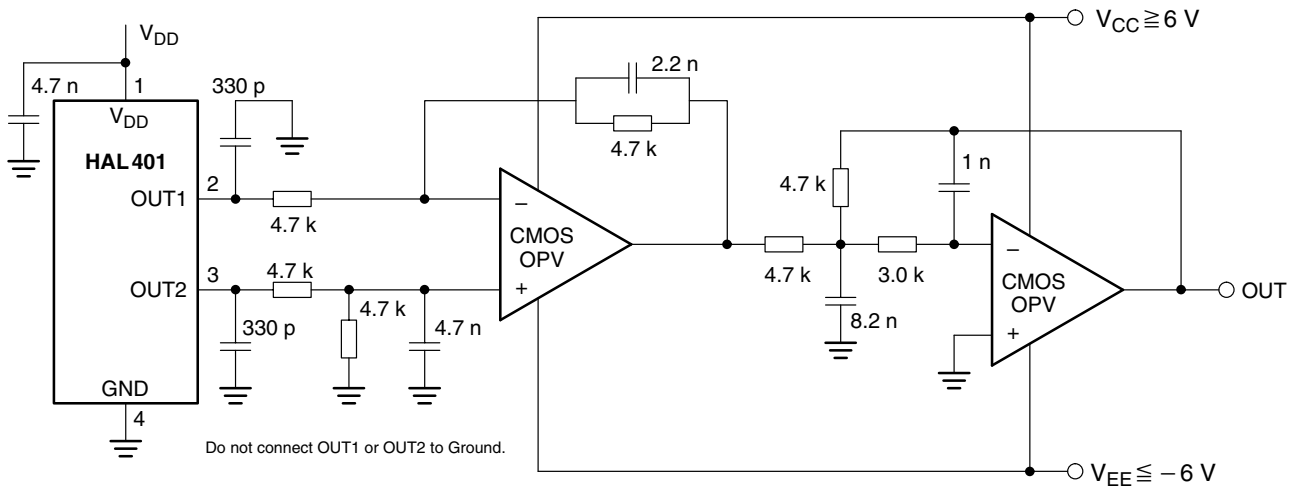


**Fig. 4–2:** Flux density measurement with voltmeter



Do not connect OUT1 or OUT2 to Ground.

**Fig. 4-3:** Differential HAL401 output to single-ended output  
 $R = 10 \text{ k}\Omega$ ,  $C = 7.5 \text{ nF}$ ,  $\Delta R$  for offset adjustment,  $BW_{-3\text{dB}} = 1.3 \text{ kHz}$



**Fig. 4-4:** Differential HAL401 output to single-ended output (referenced to ground), filter –  $BW_{-3\text{dB}} = 14.7 \text{ kHz}$

## 5. Data Sheet History

1. Final Data Sheet: "HAL401 Linear Hall Effect Sensor IC", June 26, 2002, 6251-470-1DS.  
First release of the final data sheet.
2. Final Data Sheet: "HAL401 Linear Hall Effect Sensor IC", Sept. 14, 2004, 6251-470-2DS.  
Second release of the final data sheet.  
Major changes:
  - new package diagram for SOT89-1
3. Final Data Sheet: "HAL401 Linear Hall Effect Sensor IC", Dec. 8, 2008, DSH000018\_002EN  
Third release of the final data sheet.  
Major changes:
  - Section 1.5. "Solderability and Welding" updated
  - package diagrams updated
  - Fig. 3–3: "Recommended footprint SOT89B" added