# BMA250 Digital, triaxial acceleration sensor

# Data sheet

# **Bosch Sensortec**





# **BMA250 Data sheet**

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# **BMA250**

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# **BMA250**

# Digital, triaxial ±2g to ±16g acceleration sensor with intelligent on-chip motion-triggered interrupt controller

#### **Key features**

Ultra-Small package LGA package (12 pins), footprint 2mm x 2mm,

height 0.95mm

Digital interface
 SPI (4-wire, 3-wire), I<sup>2</sup>C, 2 interrupt pins

V<sub>DDIO</sub> voltage range: 1.2V to 3.6V

Programmable functionality Acceleration ranges ±2g/±4g/±8g/±16g

Low-pass filter bandwidths 1kHz - <8Hz

On-chip interrupt controller Motion-triggered interrupt-signal generation for

- new data

- any-motion (slope) detection

- tap sensing (single tap / double tap)

- orientation recognition

- flat detection

- low-g/high-g detection

Stand-alone capability (no microcontroller needed)

• Ultra-low power ASIC Low current consumption, short wake-up time,

Advanced features for system power management

RoHS compliant, halogen-free

#### Typical applications

- Display profile switching
- Menu scrolling, tap / double tap sensing
- Gaming
- Pedometer / step counting
- Free-fall detection
- E-compass tilt compensation
- Drop detection for warranty logging
- Advanced system power management for mobile applications

#### **General description**

The BMA250 is a triaxial, low-g acceleration sensor with digital output for consumer market applications. It allows measurements of acceleration in three perpendicular axes. An evaluation circuitry (ASIC) converts the output of a micromechanical acceleration-sensing structure (MEMS) that works according to the differential capacitance principle.



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Package and interfaces of the BMA250 have been defined to match a multitude of hardware requirements. Since the sensor features an ultra-small footprint and a flat package it is ingeniously suited for mobile applications.

The BMA250 offers a variable  $V_{\text{DDIO}}$  voltage range from 1.2V to 3.6V and can be programmed to optimize functionality, performance and power consumption in customer specific applications. In addition it features an on-chip interrupt controller enabling motion-based applications without use of a microcontroller.

The BMA250 senses tilt, motion and shock vibration in cell phones, handhelds, computer peripherals, man-machine interfaces, virtual reality features and game controllers.

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# 1. Specification

If not stated otherwise, the given values are over lifetime and full performance temperature and voltage ranges, minimum/maximum values are  $\pm 3 \, \sigma$ .

**Table 1: Parameter Specification** 

Parameter	Symbol	Condition	Min	Тур	Max	Units
	g <sub>FS2g</sub>			±2		g
Assalsastina Dansas	<b>g</b> FS4g	Selectable		±4		g
Acceleration Range	g <sub>FS8g</sub>	via serial digital interface		±8		g
	<b>g</b> FS16g			±16		g
Supply Voltage Internal Domains	V <sub>DD</sub>		1.62	2.4	3.6	V
Supply Voltage I/O Domain	$V_{DDIO}$		1.2	2.4	3.6	V
Voltage Input Low Level	V <sub>IL</sub>	SPI & I <sup>2</sup> C			0.3V <sub>DDIO</sub>	-
Voltage Input High Level	V <sub>IH</sub>	SPI & I <sup>2</sup> C	0.7V <sub>DDIO</sub>			-
Voltage Output V <sub>OL</sub> Low Level		$V_{DDIO} = 1.62V$ $I_{OL} = 3mA, SPI \& I^2C$			0.2V <sub>DDIO</sub>	-
		$V_{DDIO} = 1.2V$ $I_{OL} = 3mA, SPI \& I^2C$			0.23 V <sub>DDIO</sub>	-
Voltage Output V <sub>OH</sub> High Level		V <sub>DDIO</sub> = 1.62V I <sub>OL</sub> = 2mA, SPI & I <sup>2</sup> C	0.8V <sub>DDIO</sub>			-
		$V_{DDIO}$ = 1.2V $I_{OL}$ = 2mA, SPI & I <sup>2</sup> C	0.62 V <sub>DDIO</sub>			-
Supply Current in Normal Mode	I <sub>DD</sub>	Nominal $V_{DD}$ supplies $T_A=25^{\circ}C$ , bw = 1kHz		139		μΑ
Supply Current in Low-Power Mode	I <sub>DDlp</sub>	Nominal $V_{DD}$ supplies $T_A$ =25°C, bw = 1kHz sleep duration $\geq$ 25ms		7		μΑ
Supply Current in Suspend Mode	I <sub>DDsm</sub>	Nominal V <sub>DD</sub> supplies T <sub>A</sub> =25°C		0.5		μΑ
Wake-Up Time	t <sub>w_up</sub>	from Low-Power Mode or Suspend Mode, bw = 1kHz		0.8		ms
Start-Up Time	t <sub>s_up</sub>	POR, bw = 1kHz		2		ms
Operating Temperature	T <sub>A</sub>		-40		+85	°C



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Parameter	Symbol	Condition	Min	Тур	Max	Units
Device Resolution	D <sub>res</sub>	g <sub>FS2g</sub>		3.91		mg
	S <sub>2g</sub>	g <sub>FS2g</sub> , T <sub>A</sub> =25°C		256		LSB/g
0 11 1	S <sub>4g</sub>	g <sub>FS4g</sub> , T <sub>A</sub> =25°C		128		LSB/g
Sensitivity	S <sub>8g</sub>	g <sub>FS8g</sub> , T <sub>A</sub> =25°C		64		LSB/g
	S <sub>16g</sub>	g <sub>FS16g</sub> , T <sub>A</sub> =25°C		32		LSB/g
Sensitivity Temperature Drift	TCS	$g_{FS2g}$ , -40°C $\leq T_A \leq +85$ °C Nominal $V_{DD}$ supplies		±0.02		%/K
Zero-g Offset	Off	g <sub>FS2g</sub> , T <sub>A</sub> =25°C Nominal V <sub>DD</sub> supplies		±80		mg
Zero-g Offset Temperature Drift	TCO	$g_{FS2g}$ , $-40^{\circ}C \le T_A \le +85^{\circ}C$ Nominal $V_{DD}$ supplies		±1		mg/K
	bw <sub>8</sub>			8		Hz
	bw <sub>16</sub>			16		Hz
	bw <sub>31</sub>			31		Hz
ما المان	bw <sub>63</sub>	1 <sup>st</sup> order filter, selectable		63		Hz
Bandwidth	bw <sub>125</sub>	via serial digital interface		125		Hz
	bw <sub>250</sub>	]		250		Hz
	bw <sub>500</sub>	]		500		Hz
	bw <sub>1000</sub>	]		1000		Hz
Nonlinearity	NL	best fit straight line		±0.5		%FS
Output Noise	n <sub>rms</sub>	g <sub>FS2g</sub> , T <sub>A</sub> =25°C Nominal V <sub>DD</sub> supplies Normal mode		0.8		mg/√Hz
Power Supply Rejection Rate	PSRR	T <sub>A</sub> =25°C Nominal V <sub>DD</sub> supplies			20	mg/V
Temperature Sensor Measurement Range	T <sub>S</sub>	T <sub>A</sub> =25°C Nominal V <sub>DD</sub> supplies	-40		+87.5	°C
Temperature Sensor Slope	dT <sub>S</sub>	T <sub>A</sub> =25°C Nominal V <sub>DD</sub> supplies		0.5		LSB/K
Temperature Sensor OT <sub>S</sub> T <sub>A</sub> =25°C Nominal V <sub>DD</sub> supplies				±5		K
MECHANICAL CHARAC				T = '		1
Parameter	Symbol	Condition	Min	Тур	Max	Units
Cross Axis Sensitivity	S	relative contribution between any two of the three axes		1		%
Alignment Error	E <sub>A</sub>	relative to package outline		±0.5		0

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# 2. Absolute maximum ratings

**Table 2: Absolute maximum ratings** 

Parameter	Condition	Min	Max	Units
Voltage at Supply Pin	V <sub>DD</sub> Pin	-0.3	4.25	V
Voltage at Supply Fill	V <sub>DDIO</sub> Pin	-0.3	4.25	V
Voltage at any Logic Pad	Non-Supply Pin	-0.3	V <sub>DDIO</sub> +0.3	V
Passive Storage Temp. Range	≤ 65% rel. H.	-50	+150	°C
	Duration ≤ 200µs		10,000	g
Mechanical Shock	Duration ≤ 1.0ms		2,000	g
	Free fall onto hard surfaces		1.8	m
ESD	HBM, at any Pin		2	kV
	CDM		500	V

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#### 3. Block diagram

Figure 1 shows the basic building blocks of the BMA250:

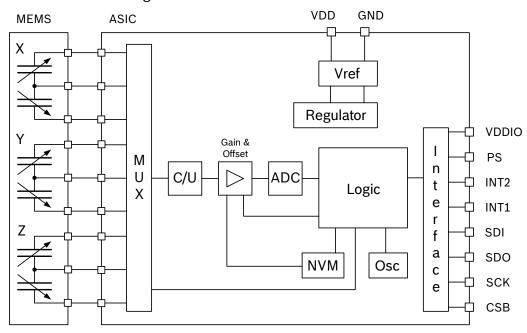


Figure 1: Block diagram of BMA250

#### 4. Functional description

Note: Default values for registers can be found in chapter 5.

#### 4.1 Power management

The BMA250 has two distinct power supply pins:

- V<sub>DD</sub> is the main power supply for all internal analog and digital functional blocks;
- V<sub>DDIO</sub> is a separate power supply pin, exclusively used for the supply of the digital interface.

There are no limitations on the voltage levels of both pins relative to each other, as long as each of them lies within its operating range. Furthermore, the device can be completely switched off ( $V_{DD} = 0V$ ) while keeping the  $V_{DDIO}$  supply on ( $V_{DDIO} > 0V$ ). To switch off the interface supply ( $V_{DDIO} = 0V$ ) and keep the internal supply on ( $V_{DD} > 0V$ ) is safe only in normal mode. If the device is in low-power mode or suspend mode while  $V_{DDIO} = 0V$ , there is a risk of excess current consumption on the  $V_{DD}$  supply (non-destructive).

It is absolutely prohibited to keep any interface at a logical high level when  $V_{DDIO}$  is switched off. Such a configuration will permanently damage the device (i.e. if  $V_{DDIO} = 0 \rightarrow [SDI \& SDO \& SCK \& CSB] \neq high)$ .

The device contains a power-on reset (POR) generator. It resets the logic part and the register values after powering-on  $V_{DDIO}$ . There is no limitation on the sequence of switching on

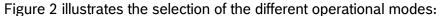
both supply voltages. In case the  $I^2C$  interface shall be used, a direct electrical connection between  $V_{DDIO}$  supply and the PS pin is needed in order to ensure reliable protocol selection (see section 4.2 Operational modes).

#### 4.2 Operational modes

Depending on the configuration the BMA250 is able to operate in two different operational modes:

- <u>General mode:</u> The device is acting as a slave on a digital interface (SPI or I²C) and is controlled by the external bus master (e.g. μC). The master gets measurement data and status information from the device through the digital interface. In particular, the master can configure the interrupt controller and read out the interrupt status registers. Moreover, it can freely configure and use the two interrupt pins (INT1, INT2). Several interrupts may be enabled in parallel.
- <u>Dedicated mode:</u> The dedicated mode allows the sensor to be operated as a standalone device in a simple μC-less system without abandon of the interrupt functionality. No digital interface is needed and, as a consequence, no measurement data can be read from the device. Instead of the digital interface the internal interrupt engine with its default setting is used. The interrupt status is mapped onto dedicated output pins. One out of three different sub-modes can be chosen: A) orientation recognition, B) tap sensing or C) slope (any-motion) detection. Only one interrupt at a time can be assigned.

The selection of the operational mode is done during start-up or reset by the state of the PS pin. If PS is floating, the dedicated mode is selected. A defined digital state selects the general mode. All pads are in input mode (no output driver active) during the start-up sequence until the operational mode and, in case of the general mode, the interface type is selected. The start-up sequence is run after power-up and after reset.



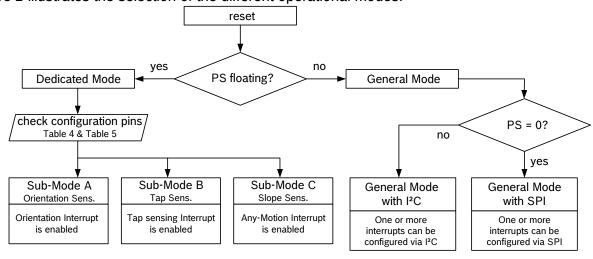


Figure 2: Operational mode selection



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#### 4.2.1 General mode

A defined digital state at the PS pin selects the general mode. Its polarity determines the kind of interface to be used:

PS = GND enables the digital SPI interface
 PS = V<sub>DDIO</sub> enables the digital I<sup>2</sup>C interface
 PS = float enables the dedicated mode

#### 4.2.2 Dedicated mode (µC-less or stand-alone mode)

The dedicated mode operates with pre-defined settings of the interrupt engine in order to generate the motion-triggered interrupt-signals, i.e. bandwidth, sleep time, low-power mode, threshold, and hysteresis are use case optimized. Nevertheless some minor configurations can be selected by the user. The dedicated mode is entered if the device is connected according to table 3. During the start-up / power on sequence the PS pin (#11) must float.

Table 3: Entering and operating dedicated mode

VDDIO	NC	VDD	GNDIO	GND	PS
Pin#3	Pin#4	Pin#7	Pin#8	Pin#9	Pin#11
$V_{DDIO}$	NC	$V_{DD}$	GND	GND	float

Depending on the configuration of the other device pins according to table 4 the corresponding sub-mode of the dedicated mode is entered. In table 4 and table 5 the unshaded entries represent necessary input values for the corresponding sub-mode selection while the shaded entries represent corresponding output parameters of the events to be detected.

Table 4: Sub-mode selection and specific outputs of the dedicated mode

Sub-Mode	SDO	SDx	INT1	INT2	CSB	SCx
	Pin#1	Pin#2	Pin#5	Pin#6	Pin#10	Pin#12
Orientation	output orient1-detect	output orient0-detect	output orient2-detect	output flat-detect	select orient sleep	GND
Тар	output double-detect	output single-detect	GND	select tap type	select tap sleep	$V_{DD}$
Slope	GND	output motion-detect	$V_{DD}$	GND	select slope sleep	$V_{DD}$

Table 5 contains state and description details of the parameters introduced in table 4. Unshaded entries represent input values to be set, shaded entries represent output parameters to be detected.



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Table 5: Description of the parameters of table 4

Sub-Mode	Parameter see Table 4	State	Description	
	output	low	"upright" for portrait / "left" for landscape	
	orient0-detect	high	"upside-down" for portrait / "right" for landscape	
	output	low	portrait	
	orient1-detect	high	landscape	
Orientation	output	low	z-axis upward looking i.e. $ \theta $ < 90° (Fig. 8)	
BW = 62.5 Hz	orient2-detect	high	z-axis downward looking i.e. $ \theta  > 90^{\circ}$ (Fig. 8)	
577 0210 112	output	low	non flat i.e. $ \theta  > 19,5^{\circ}$ (Fig. 8)	
	flat-detect	high	flat i.e.  θ  < 19,5° (Fig. 8)	
	select orient sleep	GND	Low-Power mode enabled, sleep time = 100ms	
		$V_{DD}$	Low-Power mode enabled, sleep time = 1s	
	output	low	currently no Double-Tap event	
	double-detect	high	Double-Tap event detected	
_	output	low	currently no single-tap event	
Тар	single-detect	high	Single-Tap event detected	
BW = 1k Hz	select	GND	Single-Tap detection enabled	
	tap type	$V_{DD}$	Double-Tap detection enabled	
	select	GND	Low-Power Mode disabled	
	tap sleep	$V_{DD}$	Low-Power Mode enabled, sleep time = 10ms	
Clara a	output	low	currently no Any-Motion event	
Slope	motion-detect	high	Any-Motion event detected	
BW = 125 Hz	select	GND	Low-Power mode enabled, sleep time = 50ms	
	slope sleep	$V_{DD}$	Low-Power mode enabled, sleep time = 1s	

low = GND, high =  $V_{DDIO}$ 

For more details, refer to chapter 4.3 Power modes and 4.8 Interrupt Controller

Orientation recognition sub mode

→ refer to chapter 4.8.7

• Tap sensing sub mode

→ refer to chapter 4.8.6

• Any-motion (slope) detection) sub mode

→ refer to chapter 4.8.5

#### 4.3 Power modes

The BMA250 has three different power modes. Besides normal mode, which represents the fully operational state of the device, there are two special energy saving modes: low-power mode and suspend mode.

The possible transitions between the power modes are illustrated in figure 3:

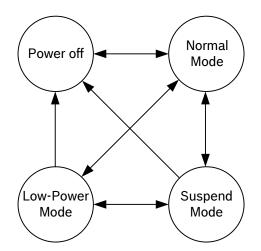


Figure 3: Power mode transition diagram

In normal mode, all parts of the electronic circuit are held powered-up and data acquisition is performed continuously.

In contrast to this, in suspend mode the whole analog part, oscillators included, is powered down. No data acquisition is performed, the only supported operations are reading registers (latest acceleration data are kept) and writing to the (0x11) suspend bit or (0x14) softreset register. Suspend mode is entered (left) by writing '1' ('0') to the (0x11) suspend bit.

In low-power mode, the device is periodically switching between a sleep phase and a wake-up phase. The wake-up phase essentially corresponds to operation in normal mode with complete power-up of the circuitry. During the sleep phase the analog part except the oscillators is powered down. Low-power mode is entered (left) by writing '1' ('0') to the (0x11) lowpower\_en bit.

During the wake-up phase the number of samples required by any enabled interrupt is processed. If an interrupt is detected, the device stays in the wake-up phase as long as the interrupt condition endures (non-latched interrupt), or until the latch time expires (temporary interrupt), or until the interrupt is reset (latched interrupt). If no interrupt is detected, the device enters the sleep phase.

The duration of the sleep phase is set by the (0x11) sleep\_dur bits as shown in the following table:



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Table 6: Sleep phase duration settings

(0x11)	Sleep Phase
sleep_dur	Duration
	<b>t</b> sleep
0000b	0.5ms
0001b	0.5ms
0010b	0.5ms
0011b	0.5ms
0100b	0.5ms
0101b	0.5ms
0110b	1ms
0111b	2ms
1000b	4ms
1001b	6ms
1010b	10ms
1011b	25ms
1100b	50ms
1101b	100ms
1110b	500ms
1111b	1s

The current consumption of the BMA250 can be calculated according to this formula:

$$I_{DDlp} pprox rac{t_{sleep} \cdot I_{DDsm} + t_{active} \cdot I_{DD}}{t_{sleep} + t_{active}}.$$

When making an estimation about the length of the wake-up phase  $t_{active}$ , the wake-up time,  $t_{w\_up}$ , has to be considered. Therefore,  $t_{active} = t_{ut} + t_{w\_up}$ , where  $t_{ut}$  is given in table 8. During the wake-up phase all analog modules are held powered-up, while during the sleep phase most analog modules are powered down. As a consequence, a wake-up time of less than 1ms (typ. value 0.8ms) is needed to settle the analog modules in order to get reliable acceleration data.

Table 7 gives an overview of the resulting average supply currents  $I_{DDlpe}$  for the different sleep phase durations and a selected bandwidth of 1000Hz, assuming no interrupt is active and thus only one sample per wake-up phase is taken:

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Table 7: Average current consumption in low-power mode

Sleep phase duration	Average current
duration	
	consumption
0.5ms	100.5 μΑ
1ms	78.8 µA
2ms	55.0 μA
4ms	34.5 µA
6ms	25.2 μΑ
10ms	16.4 µA
25ms	7.4 µA
50ms	4.0 μΑ
100ms	2.3 μΑ
500ms	0.9 μΑ
1s	0.7 μΑ

#### 4.4 Sensor data

#### 4.4.1 Acceleration data

The width of acceleration data is 10 bits given in two's complement representation. The 10 bits for each axis are split into an MSB upper part (one byte containing bits 9 to 2) and an LSB lower part (one byte containing bits 1 and 0 of acceleration and a (0x02, 0x04, 0x06) new\_data flag). Reading the acceleration data registers shall always start with the LSB part. The content of an MSB register is updated by reading the corresponding LSB register (shadowing procedure). The shadowing procedure can be disabled (enabled) by writing '1' ('0') to the bit shadow\_dis. With disabled shadowing, the content of both MSB and LSB registers is updated by a new value immediately. Unused bits of the LSB registers are fixed to 0. The (0x02, 0x04, 0x06) new\_data flag of each LSB register is set if the data registers are updated, it is reset if either the corresponding MSB or LSB part is read.

Two different streams of acceleration data are available, unfiltered and filtered. The unfiltered data is sampled with 2kHz. The sampling rate of the filtered data depends on the selected filter bandwidth; it is twice the bandwidth. Which kind of data is stored in the acceleration data registers depends on bit (0x13)  $data_high_bw$ . If (0x13)  $data_high_bw$  is '0' ('1'), then filtered (unfiltered) data is stored in the registers. Both data streams are separately offset-compensated. Both kinds of data can be processed by the interrupt controller.



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The bandwidth of filtered acceleration data is determined by setting the (0x10) bw bit as followed:

Table 8: Bandwidth configuration

bw	Bandwidth	Update Time
		<b>t</b> ut
00xxx	*)	-
01000	7.81Hz	64ms
01001	15.63Hz	32ms
01010	31.25Hz	16ms
01011	62.5Hz	8ms
01100	125Hz	4ms
01101	250Hz	2ms
01110	500Hz	1ms
01111	1000Hz	0.5ms
1xxxx	*)	-

<sup>\*)</sup> Note: Settings 00xxx result in a bandwidth of 7.81 Hz; settings 1xxxx result in a bandwidth of 1000 Hz. It is recommended to actively use the range from '01000b' to '01111b' only in order to be compatible with future products.

The BMA250 supports four different acceleration measurement ranges. A measurement range is selected by setting the (0x0F) range bits as follows:

**Table 9: Range selection** 

Range	Acceleration measurement range	Resolution
0011	±2g	3.91mg/LSB
0101	±4g	7.81mg/LSB
1000	±8g	15.62mg/LSB
1100	±16g	31.25mg/LSB
others	reserved	-

#### 4.4.2 Temperature data

The width of temperature data is 8 bits given in two's complement representation. Temperature values are available in the (0x08) temp register.

The slope of the temperature sensor is 0.5 K/LSB, its center temperature is  $24^{\circ}\text{C}$  [(0x08) temp = 0x00]. Therefore, the typical temperature measurement range is  $-40^{\circ}\text{C}$  up to  $87.5^{\circ}\text{C}$ .



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#### 4.5 Self-test

This feature permits to check the sensor functionality by applying electrostatic forces to the sensor core instead of external accelerations. By actually deflecting the seismic mass, the entire signal path of the sensor can be tested. Activating the self-test results in a static offset of the acceleration data; any external acceleration or gravitational force applied to the sensor during active self-test will be observed in the output as a superposition of both acceleration and self-test signal.

The self-test is activated individually for each axis by writing the proper value to the (0x32) self\_test\_axis bits ('01b' for x-axis, '10b' for y-axis, '11b' for z-axis, '00b' to deactivate self-test). It is possible to control the direction of the deflection through bit (0x32) self\_test\_sign. The excitation occurs in positive (negative) direction if (0x32) self\_test\_sign = '0b' ('1b').

In order to ensure a proper interpretation of the self-test signal it is recommended to perform the self-test for both (positive and negative) directions and then to calculate the difference of the resulting acceleration values. Table 10 shows the minimum differences for each axis. The actually measured signal differences can be significantly larger.

x-axis signal y-axis signal z-axis signal resulting minimum +0.8 g +0.8 g +0.4 g difference signal

Table 10: Self-test difference values

It is recommended to perform a reset of the device after self-test. If the reset cannot be performed, the following sequence must be kept to prevent unwanted interrupt generation: disable interrupts, change parameters of interrupts, wait for at least 600  $\mu$ s, enable desired interrupts.

#### 4.6 Offset compensation

Offsets in measured signals can have several causes but they are always unwanted and disturbing in many cases. Therefore, the BMA250 offers an advanced set of four digital offset compensation methods which are closely matched to each other. These are slow, fast, and manual compensation, and inline calibration.

The compensation is performed for unfiltered and filtered data independently. It is done by adding a compensation value to the acceleration data coming from the ADC. The result of this computation is saturated if necessary to prevent any overflow errors (the smallest or biggest possible value is set, depending on the sign). However, the public registers used to read and write compensation values have only a width of 8 bits.

An overview of the offset compensation principle is given in figure 4:

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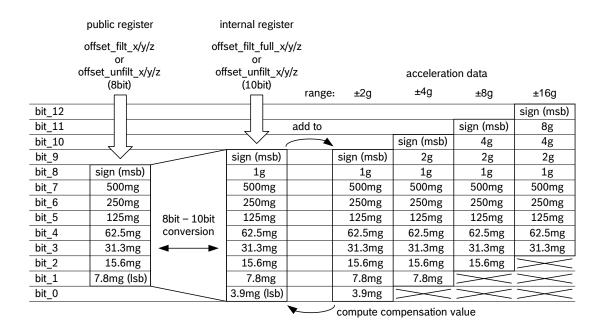


Figure 4: Principle of offset compensation

The meaning of both public and internal registers is the same for all acceleration measurement ranges. Therefore, with measurement ranges other than ±2g, one or more lower significant bits of the internal registers are lost when added to an acceleration value, or are set to zero when the internal compensation value is computed. If a compensation value is too small or too big to fit into the corresponding internal register, it is saturated to prevent an overflow error.

In a similar way the conversion of the internal register value to the public register value (10bit to 8bit) uses saturation.

Summarized, in dependence to the measurement range which has been set, the compensation value, which has been written into the public register will correct the data output according to figure 4.

```
e.g. \pm 2g range:

public register = 00000001b \rightarrow add to acceleration data = \pm 7.8mg = \pm 2LSB

public register = 00000010b \rightarrow add to acceleration data = \pm 15.6mg = \pm 4LSB

public register = 00000101b \rightarrow add to acceleration data = \pm 39.1mg = \pm 10LSB
```

The public registers are image registers of EEPROM registers. With each image update (see section 4.7 Non-volatile memory for details) the contents of the non-volatile EEPROM registers are written to the public registers. At any time the public register can be over-written by the user. After changing the contents of the public registers by either an image update or manually, all 8bit values are widened to 10bit values and stored in the corresponding internal registers. In the opposite direction, if the value of an internal register changes due to the computation performed by a compensation algorithm, it is converted to an 8bit value and stored in the public register.



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For slow and fast offset compensation, the compensation target can be chosen by setting the bits (0x37) offset\_target\_x, (0x37) offset\_target\_y, and (0x37) offset\_target\_z according to table 11:

 (0x37)
 Target value

 offset\_target\_x/y/z
 00b
 0g

 01b
 +1g

 10b
 -1g

 11b
 0g

**Table 11: Offset target settings** 

By writing '1' to the (0x36) offset\_reset bit, all offset compensation registers are reset to zero.

#### 4.6.1 Slow compensation

Slow compensation is a quasi-continuous process which regulates the acceleration value of each axis towards the target value by comparing the current value with the target and adding or subtracting a fixed value depending on the comparison.

The algorithm in detail: If an acceleration value is larger (smaller) than the target value (0x37) offset\_target\_x/y/z for a number of samples (given by the parameter Offset Period see table 12), the internal offset compensation value (0x38, 0x039, 0x3A) offset\_filt\_x/y/z or (0x3B, 0x03C, 0x3D) offset\_unfilt\_x/y/z is decremented (incremented) by 4 LSB.

The public registers (0x38, 0x039, 0x3A) offset\_filt\_x/y/z and (0x3B, 0x03C, 0x3D) offset\_unfilt\_x/y/z are not used for the computations but they are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

The compensation period offset\_period is set by the (0x37)  $cut_off$  bit as represented in table 12:

Table 12: Compensation period settings

(0x37) cut_off	Offset Period
0b	8
1b	16

The slow compensation can be enabled (disabled) for each axis independently by setting the bits  $(0x36) hp_xen, hp_yen, hp_zen$  to '1' ('0'), respectively.

Slow compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of slow compensation are not fulfilled.



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#### 4.6.2 Fast compensation

Fast compensation is a one-shot process by which the compensation value is set in such a way that when added to the raw acceleration, the resulting acceleration value of each axis equals the target value.

The algorithm in detail: An average of 16 consecutive acceleration values is computed and the difference between target value and computed value is written to (0x38, 0x39, 0x3A) offset\_filt\_x/y/z or (0x3B, 0x3C, 0x3D) offset\_unfilt\_x/y/z The public registers (0x38, 0x39, 0x3A) offset\_filt\_x/y/z and (0x3B, 0x3C, 0x3D) offset\_unfilt\_x/y/z are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

Fast compensation is triggered for each axis individually by setting the (0x36) cal\_trigger bits as shown in table 13:

(0x36) cal_trigger	Selected Axis
00b	none
01b	Х
10b	у
11b	Z

Table 13: Fast compensation axis selection

The register (0x36) cal\_trigger keeps its non-zero value while the fast compensation procedure is running. Slow compensation is blocked as long as fast compensation endures. Bit (0x36) cal\_rdy is '0' when (0x36) cal\_trigger is not '00'.

Fast compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of fast compensation are not fulfilled.

#### 4.6.3 Manual compensation

As explained above, the contents of the public compensation registers (0x38, 0x39, 0x3A) offset\_filt\_x/y/z and (0x3B, 0x3C, 0x3D) offset\_unfilt\_x/y/z can be set manually via the digital interface. It is recommended to write into these registers immediately after a new data interrupt in order not to disturb running offset computations.

Writing to the offset compensation registers is not allowed if slow compensation is enabled or if the fast compensation procedure is running.

#### 4.6.4 Inline calibration

For a given application, it is often desirable to calibrate the offset once and to store the compensation values permanently. This can be achieved by using one of the aforementioned offset compensation methods to determine the proper compensation values and then storing



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these values permanently in the non-volatile memory (EEPROM). See section 4.7 Non-volatile memory for details of the storing procedure.

Each time the device is reset, the compensation values are loaded from the non-volatile memory into the image registers and used for offset compensation until they are possibly overwritten using one of the other compensation methods.

#### 4.7 Non-volatile memory

The entire memory of the BMA250 consists of three different kinds of registers: hard-wired, volatile, and non-volatile. Non-volatile memory is implemented as EEPROM. Part of it can be both read and written by the user. Access to non-volatile memory is only possible through (volatile) image registers.

Altogether, there are eight registers (bytes) of EEPROM which are accessible by the customer. The addresses of the image registers range from 0x38 to 0x3F. While the addresses up to 0x3D are used for offset compensation (see 4.6 Offset Compensation), addresses 0x3E and 0x3F are general purpose registers not linked to any sensor-specific functionality.

The content of the EEPROM is loaded to the image registers after a reset (either POR or softreset) or after a user request which is performed by writing '1' to bit (0x33) nvm\_load. As long as the image update is not yet complete, bit (0x33) nvm\_load is '1', otherwise it is '0'.

The image registers can be read and written like any other register.

Writing to the EEPROM is a three-step procedure:

- 1. Write the new contents to the image registers.
- 2. Write '1' to bit (0x33) nvm\_prog\_mode in order to unlock the EEPROM.
- 3. Write '1' to bit (0x33) nvm\_prog\_trig and keep '1' in bit (0x33) nvm\_prog\_mode in order to trigger the write process.

Writing to the EEPROM always renews the entire EEPROM contents. It is possible to check the write status by reading bit  $(0x33) nvm_rdy$ . While  $(0x33) nvm_rdy = '0'$ , the write process is still enduring; if  $(0x33) nvm_rdy = '1'$ , then writing is completed. As long as the write process is ongoing, no power mode change and no image update is allowed. It is forbidden to write to the EEPROM while the image update is running, in low-power mode, and in suspend mode.

#### 4.8 Interrupt controller

Seven interrupt engines are integrated in the BMA250. Each interrupt can be independently enabled and configured. If the condition of an enabled interrupt is fulfilled, the corresponding status bit is set to '1' and the selected interrupt pin is activated. There are two interrupt pins, INT1 and INT2; interrupts can be freely mapped to any of these pins. The pin state is a logic 'or' combination of all mapped interrupts.

The interrupt status registers are updated together with writing new data into the acceleration data registers. If an interrupt is disabled, all active status bits and pins are immediately reset.



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All time constants are based upon the typical frequency of the internal oscillator. This is reflected by the bandwidths (bw) as specified in table 1.

#### 4.8.1 General features

An interrupt is cleared depending on the selected interrupt mode, which is common to all interrupts. There are three different interrupt modes: non-latched, latched, and temporary. The mode is selected by the (0x21) latch\_int bits according to table 14

(0x21)Interrupt mode latch\_int 0000b non-latched 0001b temporary, 250ms 0010b temporary, 500ms 0011b temporary, 1s 0100b temporary, 2s 0101b temporary, 4s 0110b temporary, 8s 0111b latched 1000b non-latched 1001b temporary, 500µs 1010b temporary, 500µs 1011b temporary, 1ms 1100b temporary, 12.5ms 1101b temporary, 25ms 1110b temporary, 50ms 1111b latched

Table 14: Interrupt mode selection

An interrupt is generated if its activation condition is met. It can not be cleared as long as the activation condition is fulfilled. In the non-latched mode the interrupt status bit and the selected pin (the contribution to the 'or' condition for INT1 and/or INT2) are cleared as soon as the activation condition is no more valid. Exceptions to this behaviour are the new data, orientation, and flat interrupts, which are automatically reset after a fixed time.

In the latched mode an asserted interrupt status and the selected pin are cleared by writing '1' to bit (0x21) reset\_int. If the activation condition still holds when it is cleared, the interrupt status is asserted again with the next change of the acceleration registers.

In the temporary mode an asserted interrupt and selected pin are cleared after a defined period of time. The behaviour of the different interrupt modes is shown graphically in figure 5:



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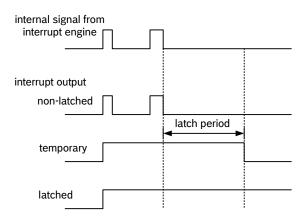


Figure 5: Interrupt modes

Several interrupt engines can use either unfiltered or filtered acceleration data as their input. For these interrupts, the source can be selected with the respective (0x1E)  $int\_src\_...$  bits, in details these are (0x1E)  $int\_src\_data$ , (0x1E)  $int\_src\_tap$ , (0x1E)  $int\_src\_slope$ , (0x1E)  $int\_src\_lope$ , (0

It is strongly recommended to set interrupt parameters prior to enabling the interrupt. Changing parameters of an already enabled interrupt may cause unwanted interrupt generation and generation of a false interrupt history. A safe way to change parameters of an enabled interrupt is to keep the following sequence: disable the desired interrupt, change parameters, wait for at least  $600~\mu s$ , enable the desired interrupt.

#### 4.8.2 Mapping (inttype to INT Pin#)

The mapping of interrupts to the interrupt pins #05 or #06 is done by registers (0x19) to (0x1B). Setting (0x19) int1\_"inttyp" to '1' ('0') maps (unmaps) "inttyp" to pin #5 (INT1), correspondingly setting (0x1B) int2\_"inttyp" to '1' ('0') maps (unmaps) "inttyp" to pin #6 (INT2).

Note: "inttyp" to be replaced with the precise notation, given in the memory map in chapter 5.

Example: For flat interrupt (int1\_flat): Setting (0x19) int1\_flat to '1' maps int1\_flat to pin #5 (INT1).

# 4.8.3 Electrical behaviour (INT pin# to open-drive or push-pull)

Both interrupt pins can be configured to show desired electrical behaviour. The 'active' level of each pin is determined by the (0x20) int1 |v| and (0x20) int2 |v| bits.

If (0x20) int1\_lvl = '1' ('0') / (0x20) int2\_lvl = '1' ('0'), then pin #05 (INT1) / pin #06 (INT2) is active '1' ('0'). In addition to that, also the electric type of the interrupt pins can be selected. By

setting bits (0x20) int1\_od / (0x20) int2\_od to '0', the interrupt pin output type gets push-pull, by setting the configuration bits to '1', the output type gets open-drive.

Remark: Due to their use for sub-mode selection in dedicated mode, the states of both INT pins are not defined during the first 2 ms after power-up.

#### 4.8.4 New data interrupt

This interrupt serves for synchronous reading of acceleration data. It is generated after storing a new value of z-axis acceleration data in the data register. The interrupt is cleared automatically when the next cycle of data acquisition starts. The interrupt status is '0' for at least 50µs.

The interrupt mode of the new data interrupt is fixed to non-latched.

It is enabled (disabled) by writing '1' ('0') to bit (0x17) data\_en. The interrupt status is stored in bit (0x0A) data\_int.

#### 4.8.5 Any-motion (slope) detection

Any-motion detection uses the slope between successive acceleration signals to detect changes in motion. An interrupt is generated when the slope (absolute value of acceleration difference) exceeds a preset threshold. It is cleared as soon as the slope falls below the threshold. The principle is made clear in figure 6.

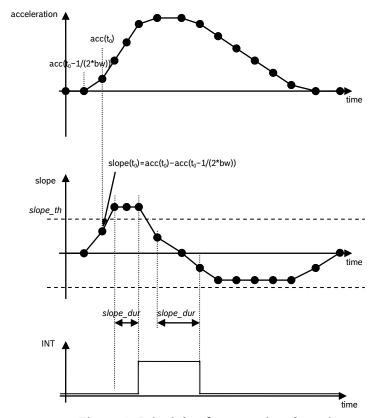


Figure 6: Principle of any-motion detection



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The threshold is set with the value of register (0x28) slope\_th. 1 LSB of (0x28) slope\_th corresponds to 1 LSB of acceleration data. Therefore, an increment of (0x28) slope\_th is 3.91 mg in 2g-range (7.81 mg in 4g-range, 15.6 mg in 8g-range and 31.3 mg in 16g-range). And the maximum value is 996 mg in 2g-range (1.99g in 4g-range, 3.98g in 8g-range and 7.97g in 16g-range).

The time difference between the successive acceleration signals depends on the selected bandwidth and equates to 1/(2\*bandwidth) ( $\triangle t=1/(2*bw)$ ). In order to suppress failure signals, the interrupt is only generated (cleared) if a certain number N of consecutive slope data points is larger (smaller) than the slope threshold given by (0x28) slope\_th. This number is set by the (0x27) slope\_dur bits. It is N = (0x27) slope\_dur + 1 for (0x27).

Example: (0x27) slope\_dur = 00b, ..., 11b = 1decimal, ..., 4decimal

#### 4.8.5.1 Enabling (disabling) for each axis

Any-motion detection can be enabled (disabled) for each axis separately by writing '1' ('0') to bits (0x16)  $slope\_en\_x$ , (0x16)  $slope\_en\_y$ , (0x16)  $slope\_en\_z$ . The criteria for any-motion detection are fulfilled and the slope interrupt is generated if the slope of any of the enabled axes exceeds the threshold (0x28)  $slope\_th$  for [(0x27)  $slope\_dur +1]$  consecutive times. As soon as the slopes of all enabled axes fall or stay below this threshold for [(0x27)  $slope\_dur +1]$  consecutive times the interrupt is cleared unless interrupt signal is latched.

#### 4.8.5.2 Axis and sign information of any motion interrupt

The interrupt status is stored in bit (0x09)  $slope_int$ . The any-motion interrupt supplies additional information about the detected slope. The axis which triggered the interrupt is given by that one of bits (0x0B)  $slope_first_x$ , (0x0B)  $slope_first_y$ , (0x0B)  $slope_first_z$  that contains a '1'. The sign of the triggering slope is held in bit (0x0B)  $slope_sign$ . If (0x0B)  $slope_sign = '0'$  ('1'), the sign is positive (negative).

#### 4.8.5.3 Serial interface and dedicated wake-up mode

When serial interface is active, any-motion detection logic is enabled if any of the axis specific (0x16) slope\_en\_... register bits are set. To disable the any-motion interrupt, clear all the axis specific (0x16) slope en ... bits.

In the dedicated wake-up mode (see chapter 4.2.2), all three axes are enabled for any-motion detection whether the individual axis enable bits are set or not.

#### 4.8.6 Tap sensing

Tap sensing has a functional similarity with a common laptop touch-pad or clicking keys of a computer mouse. A tap event is detected if a pre-defined slope of the acceleration of at least one axis is exceeded. Two different tap events are distinguished: A 'single tap' is a single event within a certain time, followed by a certain quiet time. A 'double tap' consists of a first such event followed by a second event within a defined time frame.

Only one of the tap interrupts can be enabled at the same time. Single tap interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) s\_tap\_en. Double tap interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) d\_tap\_en. If one tries to enable both interrupts by writing '1' to

(0x16) s\_tap\_en and (0x16) d\_tap\_en, then only (0x16) d\_tap\_en keeps the value '1' and the double tap interrupt is enabled.

The status of the single tap interrupt is stored in bit (0x09) s\_tap\_int, the status of the double tap interrupt is stored in bit (0x09) d tap int.

The slope threshold for detecting a tap event is set by bits (0x2B)  $tap_th$ . The meaning of (0x2B)  $tap_th$  depends on the range setting. 1 LSB of (0x2B)  $tap_th$  corresponds to a slope of 62.5mg in 2g-range, 125mg in 4g-range, 250mg in 8g-range, and 500mg in 16g-range.

In figure 7 the meaning of the different timing parameters is visualized:

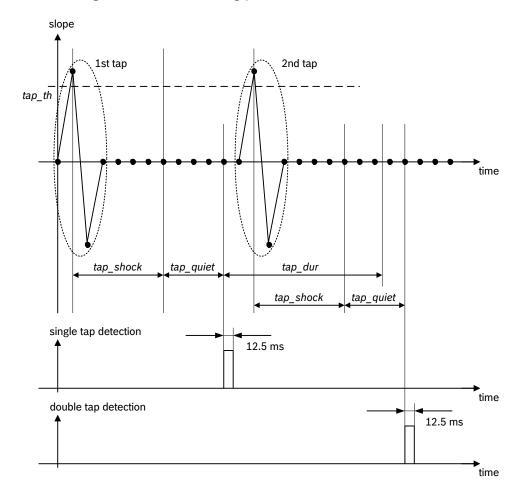


Figure 7: Timing of tap detection

The parameters (0x2A)  $tap\_shock$  and (0x2A)  $tap\_quiet$  apply to both single tap and double tap detection, while (0x2A)  $tap\_dur$  applies to double tap detection only. Within the duration of (0x2A)  $tap\_shock$  any slope exceeding (0x2B)  $tap\_th$  after the first event is ignored. Contrary to this, within the duration of (0x2A)  $tap\_quiet$  no slope exceeding (0x2B)  $tap\_th$  must occur, otherwise the first event will be cancelled.



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#### 4.8.6.1 Single tap detection

A single tap is detected and the single tap interrupt is generated after the combined durations of (0x2A)  $tap\_shock$  and (0x2A)  $tap\_quiet$ , if the corresponding slope conditions are fulfilled. The interrupt is cleared after a delay of 12.5 ms.

#### 4.8.6.2 Double tap detection

A double tap is detected and the double tap interrupt is generated if an event fulfilling the conditions for a single tap occurs within the set duration in (0x2A) tap\_dur after the completion of the first tap event. The interrupt is cleared after a delay of 12.5 ms.

#### 4.8.6.3 Selecting the timing of tap detection

For each of parameters (0x2A)  $tap\_shock$  and (0x2A)  $tap\_quiet$  two values are selectable. By writing '0' ('1') to bit (0x2A)  $tap\_shock$  the duration of (0x2A)  $tap\_shock$  is set to 50 ms (75 ms). By writing '0' ('1') to bit (0x2A)  $tap\_quiet$  the duration of (0x2A)  $tap\_quiet$  is set to 30 ms (20 ms).

The length of (0x2A)  $tap\_dur$  can be selected by setting the (0x2A)  $tap\_dur$  bits according to table 15:

(0x2A) tap_dur	length of tap_dur
000b	50 ms
001b	100 ms
010b	150 ms
011b	200 ms
100b	250 ms
101b	375 ms
110b	500 ms
111b	700 ms

Table 15: Selection of tap\_dur

#### 4.8.6.4 Axis and sign information of tap sensing

The sign of the slope of the first tap which triggered the interrupt is stored in bit (0x0B) tap\_sign ('0' means positive sign, '1' means negative sign). The value of this bit persists after clearing the interrupt.

The axis which triggered the interrupt is indicated by bits (0x0B) tap\_first\_x, (0x0B) tap\_first\_y, and (0x0B) tap\_first\_z.

The bit corresponding to the triggering axis contains a '1' while the other bits hold a '0'. These bits are cleared together with clearing the interrupt status.

#### 4.8.6.5 Tap sensing in low power mode

In low-power mode, a limited number of samples is processed after wake-up to decide whether an interrupt condition is fulfilled. The number of samples is selected by bits (0x2B) tap\_samp according to table 16.

(0x2B) tap_samp	Number of Samples
00b	2
01b	4
10b	8
11b	16

Table 16: Meaning of (0x2B) tap\_samp

# 4.8.7 Orientation recognition

The orientation recognition feature informs on an orientation change of the sensor with respect to the gravitational field vector 'g'. The measured acceleration vector components with respect to the gravitational field are defined as shown in figure 8.

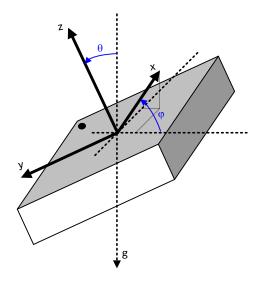


Figure 8: Definition of vector components

Therefore, the magnitudes of the acceleration vectors are calculated as follows:

$$acc_x = 1g \cdot sin\theta \cdot cos\varphi$$
  
 $acc_y = -1g \cdot sin\theta \cdot sin\varphi$   
 $acc_z = 1g \cdot cos\theta$   
 $\rightarrow acc_y/acc_x = -tan\varphi$ 



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Depending on the magnitudes of the acceleration vectors the orientation of the device in the space is determined and stored in the three (0x0C) orient bits. These bits may not be reset in the sleep phase of low-power mode. There are three orientation calculation modes with different thresholds for switching between different orientations: symmetrical, high-asymmetrical, and low-asymmetrical. The mode is selected by setting the (0x2C) orient\_mode bits as given in table 17.

(0x2C)<br/>orient\_modeOrientation Mode00bsymmetrical01bhigh-asymmetrical10blow-asymmetrical11bsymmetrical

Table 17: Orientation mode settings

For each orientation mode the (0x0C) orient bits have a different meaning as shown in table 18 to table 20:

Table 18: Meaning of the (0x0C) orient bits in symmetrical mode

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	315° < φ < 45°	acc_y  <  acc_x  - 'hyst' and acc_x - 'hyst' ≥ 0
x01	portrait upside down	135° < φ < 225°	acc_y  <  acc_x  - 'hyst' and acc_x + 'hyst' < 0
x10	landscape left	45° < φ < 135°	acc_y  ≥  acc_x  + 'hyst' and acc_y < 0
x11	landscape right	225° < φ < 315°	acc_y  ≥  acc_x  + 'hyst' and acc_y ≥ 0

Table 19: Meaning of the (0x0C) orient bits in high-asymmetrical mode

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	297° < φ < 63°	$ acc_y  < 2 \cdot  acc_x  - 'hyst'$
			and acc_x - 'hyst' ≥ 0
x01	portrait upside down	117° < φ < 243°	acc_y  < 2 ·  acc_x  - 'hyst'
			and acc_x + 'hyst' < 0
x10	landscape left	63° < φ < 117°	$ acc_y  \ge 2 \cdot  acc_x  + 'hyst'$
			and acc_y < 0
x11	landscape right	243° < φ < 297°	$ acc_y  \ge 2 \cdot  acc_x  + 'hyst'$
			and acc_y ≥ 0

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Table 20: Meaning of the (0x0C) orient bits in low-asymmetrical mode

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	333° < φ < 27°	$ acc_y  < 0.5 \cdot  acc_x  - 'hyst'$
			and acc_x - 'hyst' ≥ 0
x01	portrait upside down	153° < φ < 207°	$ acc_y  < 0.5 \cdot  acc_x  - 'hyst'$
			and acc_x + 'hyst' < 0
x10	landscape left	27° < φ < 153°	$ acc_y  \ge 0.5 \cdot  acc_x  + 'hyst'$
			and acc_y < 0
x11	landscape right	207° < φ < 333°	$ acc_y  \ge 0.5 \cdot  acc_x  + 'hyst'$
			and acc_y ≥ 0

In the preceding tables, the parameter 'hyst' stands for a hysteresis, which can be selected by setting the (0x0C) orient\_hyst bits. 1 LSB of (0x0C) orient\_hyst always corresponds to 62.5 mg, in 2g-range, 125 mg in 4g-range, 250 mg in 8g-range and 500 mg in 16g-range. It is important to note that by using a hysteresis  $\neq$  0 the actual switching angles become different from the angles given in the tables since there is an overlap between the different orientations.

The most significant bit of the (0x0C) orient bits (which is displayed as an 'x' in the above given tables) contains information about the direction of the z-axis. It is set to '0' ('1') if  $acc_z \ge 0$  ( $acc_z < 0$ ).

Figure 9 shows the typical switching conditions between the four different orientations for the symmetrical mode (i.e. without hysteresis):

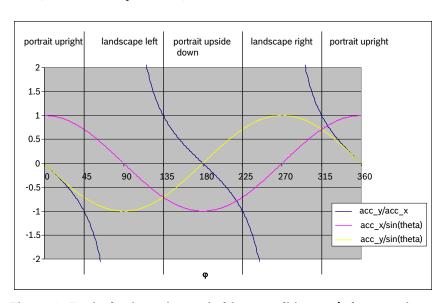


Figure 9: Typical orientation switching conditions w/o hysteresis

The orientation interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) orient\_en. The interrupt is generated if the value of (0x0C) orient has changed. It is automatically cleared after



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one stable period of the (0x0C) orient value. The interrupt status is stored in the (0x09) orient\_int bit.

If temporary or latched interrupt mode is used, after the generation of the interrupt the changed (0x0C) orient value is kept fixed as long as the interrupt persists (e. g. until the latch time expires or the interrupt is reset). After clearing the interrupt, the (0x0C) orient is only updated with the next following value change (i.e. with the next occurring interrupt). In order to ensure the continuous availability of up-to-date orientation data it is therefore optimal to use the non-latched interrupt. It is strongly advised against using latched interrupt mode or temporary interrupt mode with latch times above 50 ms for orient recognition.

#### 4.8.7.1 Orientation blocking

The change of the (0x0C) orient value and – as a consequence – the generation of the interrupt can be blocked according to conditions selected by setting the value of the (0x2C) orient\_blocking bits as described by table 21.

_	_
(0x2C) orient_blocking	Conditions
00b	no blocking
01b	theta blocking
10b	theta blocking
	or
	acceleration slope in any axis > 0.2 g
11b	value of orient is not stable for at least 100 ms
	or
	theta blocking
	or
	acceleration slope in any axis > 0.4 g

Table 21: Blocking conditions for orientation recognition

The theta blocking is defined by the following inequality:

$$|\tan \theta| < \frac{\sqrt{blocking\_theta}}{8}.$$

The parameter *blocking\_theta* of the above given equation stands for the contents of the *(0x2D)* orient\_theta bits. Hereby it is possible to define a blocking angle between 0° and 44.8°. The internal blocking algorithm saturates the acceleration values before further processing. As a consequence, the blocking angles are strictly valid only for a device at rest; they can be different if the device is moved.

#### Example:

To get a maximum blocking angle of 19° the parameter *blocking\_theta* is determined in the following way:  $(8 * tan(19°))^2 = 7.588$ , therefore, *blocking\_value* = 8dec = 001000b has to be chosen.



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In order to avoid unwanted generation of the orientation interrupt in a nearly flat position ( $z \sim 0$ , sign change due to small movements or noise), a hysteresis of 0.2 g is implemented for the z-axis, i. e. a after a sign change the interrupt is only generated after |z| > 0.2 g.

#### 4.8.8 Flat detection

The flat detection feature gives information about the orientation of the devices' z-axis relative to the g-vector, i. e. it recognizes whether the device is in a flat position or not.

The condition for the device to be in the flat position is

$$|\tan \theta| < \frac{\sqrt{parameter\_theta}}{8}.$$

Like  $blocking\_theta$ , used with orientation recognition, the  $parameter\_theta$  stands for a user-defined setting. In this case the content of the (0x2E)  $flat\_theta$  bits. The possible flat angles also range from 0° to 44.8°. To ensure proper operation,  $parameter\_theta$  has to be less than or equal to  $blocking\_theta$ .

The flat interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) flat\_en. The flat interrupt is generated if the flat value has changed and the new value is stable for at least the time given by the (0x2F) flat\_hold\_time bits. The flat value is stored in the (0x0C) flat bit if the interrupt is enabled. This value is '1' if the device is in the flat position, it is '0' otherwise. The content of the (0x0C) flat bit is changed only if the interrupt is generated. The interrupt is automatically cleared after one sample period. Its status is stored in the (0x09) flat int bit.

If temporary or latched interrupt mode is used, after the generation of the interrupt the changed (0x0C) flat value is kept fixed as long as the interrupt persists (e. g. until the latch time expires or the interrupt is reset). After clearing the interrupt, the (0x0C) flat value is only updated with the next following value change (i.e. with the next occurring interrupt).

The meaning of the (0x2F) flat\_hold\_time bits can be seen from table 22.

Table 22: Meaning of flat\_hold\_time

(0x2F) flat_hold_time	Time
00b	0
01b	512 ms
10b	1024 ms
11b	2048 ms



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#### 4.8.9 Low-g interrupt

This interrupt is based on the comparison of acceleration data against a low-g threshold, which is most useful for free-fall detection.

The interrupt is enabled (disabled) by writing '1' ('0') to the (0x17) low\_en bit. There are two modes available, 'single' mode and 'sum' mode. In 'single' mode, the acceleration of each axis is compared with the threshold; in 'sum' mode, the sum of absolute values of all accelerations  $|acc_x| + |acc_y| + |acc_z|$  is compared with the threshold. The mode is selected by the contents of the (0x24) low mode bit: '0' means 'single' mode, '1' means 'sum' mode.

The low-g threshold is set through the (0x23) low\_th register. 1 LSB of (0x23) low\_th always corresponds to an acceleration of 7.81 mg (i.e. increment is independent from g-range setting).

A hysteresis can be selected by setting the (0x24) low\_hy bits. 1 LSB of (0x24) low\_hy always corresponds to an acceleration difference of 125 mg in any g-range (as well, increment is independent from g-range setting).

The low-g interrupt is generated if the absolute values of the acceleration of all axes ('and' relation, in case of single mode) or their sum (in case of sum mode) are lower than the threshold for at least the time defined by the  $(0x22) low_dur$  register. The interrupt is reset if the absolute value of the acceleration of at least one axis ('or' relation, in case of single mode) or the sum of absolute values (in case of sum mode) is higher than the threshold plus the hysteresis for at least one data acquisition. In bit  $(0x09) low_int$  the interrupt status is stored.

The relation between the content of (0x22)  $low_dur$  and the actual delay of the interrupt generation is: delay [ms] = [(0x22)  $low_dur + 1] \cdot 2$  ms. Therefore, possible delay times range from 2 ms to 512 ms.

#### 4.8.10 High-g interrupt

This interrupt is based on the comparison of acceleration data against a high-g threshold for the detection of shock or other high-acceleration events.

The high-g interrupt is enabled (disabled) per axis by writing '1' ('0') to bits (0x17) high\_en\_x, (0x17) high\_en\_y, and (0x17) high\_en\_z, respectively. The high-g threshold is set through the (0x26) high\_th register. The meaning of an LSB of (0x26) high\_th depends on the selected grange: it corresponds to 7.81 mg in 2g-range, 15.63 mg in 4g-range, 31.25 mg in 8g-range, and 62.5 mg in 16g-range (i.e. increment depends from g-range setting).

A hysteresis can be selected by setting the (0x24) high\_hy bits. Analogously to (0x26) high\_th, the meaning of an LSB of (0x24) high\_hy is g-range dependent: it corresponds to an acceleration difference of 125 mg in 2g-range, 250 mg in 4g-range, 500 mg in 8g-range, and 1000mg in 16g-range (as well, increment depends from g-range setting).

The high-g interrupt is generated if the absolute value of the acceleration of at least one of the enabled axes ('or' relation) is higher than the threshold for at least the time defined by the (0x25) high\_dur register. The interrupt is reset if the absolute value of the acceleration of all enabled axes ('and' relation) is lower than the threshold minus the hysteresis for at least the



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time defined by the (0x25) high\_dur register. In bit (0x09) high\_int the interrupt status is stored. The relation between the content of (0x25) high\_dur and the actual delay of the interrupt generation is delay [ms] = [(0x22) low\_dur + 1] • 2 ms. Therefore, possible delay times range from 2 ms to 512 ms.

# 4.8.10.1 Axis and sign information of high-g interrupt

The axis which triggered the interrupt is indicated by bits (0x0C) high\_first\_x, (0x0C) high\_first\_y, and (0x0C) high\_first\_z. The bit corresponding to the triggering axis contains a '1' while the other bits hold a '0'. These bits are cleared together with clearing the interrupt status. The sign of the triggering acceleration is stored in bit (0x0C) high\_sign. If (0x0C) high\_sign = '0' ('1'), the sign is positive (negative).

#### 5. Register description

#### 5.1 General remarks

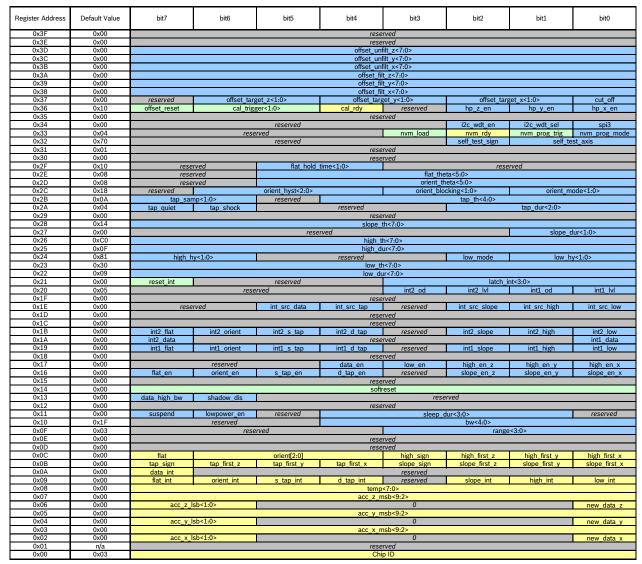
The entire communication with the device is performed by reading from and writing to registers (exception: dedicated mode, see chapter 4.2.2). Registers have a width of 8 bits; they are mapped to a common space of 64 addresses from (0x00) up to (0x3F). Within the used range there are several registers which are either completely or partially marked as 'reserved'. Any reserved bit is ignored when it is written and no specific value is guaranteed when read. It is recommended not to use registers at all which are completely marked as 'reserved'. Furthermore it is recommended to mask out (logical and with zero) reserved bits of registers which are partially marked as reserved.

Registers with addresses from (0x00) up to (0x0E) are read-only. Any attempt to write to these registers is ignored. There are bits within some registers that connected with an action to be done and, therefore, are intended for write-only access, e. g. (0x21) reset\_int or the entire (0x14) softreset register. Such bits always give '0' when read.



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## 5.2 Register map



w/r
write only
read only
reserved



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## 5.3 Chip ID

**Register** (0x00) Chip ID contains the chip identification number.

Table 23: Chip identification number, register (0x00)

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ĺ	0	0	0	0	0	0	1	1

Register (0x01) is reserved

#### 5.4 Acceleration data

**Register (0x02)** contains the LSB part of x-axis acceleration data and the new data flag for the x-axis.

Table 24: LSB part of x-axis acceleration, register (0x02)

(0x02) Bit	Name	Description
Bit 7	acc_x_lsb <1>	Bit 1 of x-axis acceleration data
Bit 6	acc_x_lsb <0>	Bit 0 of x-axis acceleration data = x LSB
Bit 5	-	(fixed to 0)
Bit 4	-	(fixed to 0)
Bit 3	-	(fixed to 0)
Bit 2	-	(fixed to 0)
Bit 1	-	(fixed to 0)
Bit 0	new_data_x	New data flag of x-axis

**Register (0x03)** contains the MSB part of x-axis acceleration data.

Table 25: MSB part of x-axis acceleration, register (0x03)

(0x03) Bit	Name	Description
Bit 7	acc_x_msb <9>	Bit 9 of x-axis acceleration data = x MSB
Bit 6	acc_x_msb <8>	Bit 8 of x-axis acceleration data
Bit 5	acc_x_msb <7>	Bit 7 of x-axis acceleration data
Bit 4	acc_x_msb <6>	Bit 6 of x-axis acceleration data
Bit 3	acc_x_msb <5>	Bit 5 of x-axis acceleration data
Bit 2	acc_x_msb <4>	Bit 4 of x-axis acceleration data
Bit 1	acc_x_msb <3>	Bit 3 of x-axis acceleration data
Bit 0	acc_x_msb <2>	Bit 2 of x-axis acceleration data



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**Register** (0x04) contains the LSB part of y-axis acceleration data and the new data flag for the y-axis.

Table 26: LSB part of y-axis acceleration, register (0x04)

(0x04) Bit	Name	Description
Bit 7	acc_y_lsb <1>	Bit 1 of y-axis acceleration data
Bit 6	acc_y_lsb <0>	Bit 0 of y-axis acceleration data = y LSB
Bit 5	-	(fixed to 0)
Bit 4	-	(fixed to 0)
Bit 3	-	(fixed to 0)
Bit 2	-	(fixed to 0)
Bit 1	-	(fixed to 0)
Bit 0	new_data_y	New data flag of y-axis

**Register (0x05)** contains the MSB part of acceleration data for the y-axis.

Table 27: MSB part of y-axis acceleration, register (0x05)

(0x05) Bit	Name	Description
Bit 7	acc_y_msb <9>	Bit 9 of y-axis acceleration data = y MSB
Bit 6	acc_y_msb <8>	Bit 8 of y-axis acceleration data
Bit 5	acc_y_msb <7>	Bit 7 of y-axis acceleration data
Bit 4	acc_y_msb <6>	Bit 6 of y-axis acceleration data
Bit 3	acc_y_msb <5>	Bit 5 of y-axis acceleration data
Bit 2	acc_y_msb <4>	Bit 4 of y-axis acceleration data
Bit 1	acc_y_msb <3>	Bit 3 of y-axis acceleration data
Bit 0	acc_y_msb <2>	Bit 2 of y-axis acceleration data

**Register (0x06)** contains the LSB part of acceleration data and the new data flag for the z-axis.

Table 28: LSB part of y-axis acceleration, register (0x06)

(0x06) Bit	Name	Description
Bit 7	acc_z_lsb <1>	Bit 1 of z-axis acceleration data
Bit 6	acc_z_lsb <0>	Bit 0 of z-axis acceleration data = z LSB
Bit 5	-	(fixed to 0)
Bit 4	-	(fixed to 0)
Bit 3	-	(fixed to 0)
Bit 2	-	(fixed to 0)
Bit 1	-	(fixed to 0)
Bit 0	new_data_z	New data flag of z-axis



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**Register (0x07)** contains the MSB part of acceleration data for the z-axis.

Table 29: MSB part of z-axis acceleration, register (0x07)

(0x07) Bit	Name	Description
Bit 7	acc_z_msb <9>	Bit 9 of z-axis acceleration data = z MSB
Bit 6	acc_z_msb <8>	Bit 8 of z-axis acceleration data
Bit 5	acc_z_msb <7>	Bit 7 of z-axis acceleration data
Bit 4	acc_z_msb <6>	Bit 6 of z-axis acceleration data
Bit 3	acc_z_msb <5>	Bit 5 of z-axis acceleration data
Bit 2	acc_z_msb <4>	Bit 4 of z-axis acceleration data
Bit 1	acc_z_msb <3>	Bit 3 of z-axis acceleration data
Bit 0	acc_z_msb <2>	Bit 2 of z-axis acceleration data

## 5.5 Temperature data

**Register** (0x08) temp contains temperature data in two's complement representation. Center temperature =  $24 \, ^{\circ}\text{C} \rightarrow \text{i.e.}$  (0x08) temp = 000000000b 1 LSB increment of temperature sensor is  $0.5 \, ^{\circ}\text{C}$  (0.9  $^{\circ}\text{F}$ ).

Table 30: Temperature data, register (0x08)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Temp <7>	Temp <6>	Temp <5>	Temp <4>	Temp <3>	Temp <2>	Temp <1>	Temp <0>

#### 5.6 Status registers

**Register (0x09)** contains the states of several interrupts.

Table 31: Interrupt status, register (0x09)

(0x09) Bit	Name	Description
Bit 7	flat_int	Flat interrupt status
Bit 6	orient_int	Orientation interrupt status
Bit 5	s_tap_int	Single tap interrupt status
Bit 4	d_tap_int	Double tap interrupt status
Bit 3	- reserved -	reserved
Bit 2	slope_int	Slope interrupt status
Bit 1	high_int	High-g interrupt status
Bit 0	low_int	Low-g interrupt status



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**Register (0x0A)** contains the status of the new data interrupt.

Table 32: New data status, register (0x0A)

(0x0A) Bit	Name	Description
Bit 7	data_int	New data interrupt status
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	- reserved -	reserved
Bit 3	- reserved -	reserved
Bit 2	- reserved -	reserved
Bit 1	- reserved -	reserved
Bit 0	- reserved -	reserved

**Register** (0x0B) contains the sign and triggering axis information for the tap and slope interrupts. Here tap interrupt comprises both single and double tap interrupt.

Table 33: Tap and slope interrupts status, register (0x0B)

(0x0B) Bit	Name	Description
Bit 7	tap_sign	Sign of 1 <sup>st</sup> tap that triggered the interrupt ('0'=positive, '1'=negative)
Bit 6	tap_first_z	'1' indicates that z-axis is triggering axis of tap interrupt
Bit 5	tap_first_y	'1' indicates that y-axis is triggering axis of tap interrupt
Bit 4	tap_first_x	'1' indicates that x-axis is triggering axis of tap interrupt
Bit 3	slope_sign	Sign of slope that triggered the interrupt ('0'=positive, '1'=negative)
Bit 2	slope_first_z	'1' indicates that z-axis is triggering axis of slope interrupt
Bit 1	slope_first_y	'1' indicates that y-axis is triggering axis of slope interrupt
Bit 0	slope_first_x	'1' indicates that x-axis is triggering axis of slope interrupt

**Register** (0x0C) contains the flat and orientation status, and the sign and triggering axis information for the high-g interrupt. Registers (0x0D) and (0x0E) are reserved.

Table 34: Flat and orientation Status, register (0x0C)

(0x0C) Bit	Name	Description
Bit 7	flat	flat detection ('1' if flat condition is fulfilled, '0' otherwise)
Bit 6	orient <2>	orientation value of z-axis ('0' if upward looking, '1' if downward looking)
Bit 5	orient <1>	orientation value of x-y plane ('00'=portrait upright,
Bit 4	orient <0>	'01'=portrait upside-down, '10'=landscape left, '11'=landscape right)
Bit 3	high_sign	Sign of slope that triggered the interrupt ('0'=positive, '1'=negative)
Bit 2	high_first_z	'1' indicates that z-axis is triggering axis of high-g interrupt
Bit 1	high_first_y	'1' indicates that y-axis is triggering axis of high-g interrupt
Bit 0	high_first_x	'1' indicates that x-axis is triggering axis of high-g interrupt

**Registers** (0x0D) and (0x0E) are reserved.



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## 5.7 g-range selection

**Register** (0x0F) contains the selection of the g-range. Proper settings for (0x0F) range are '0011b' (selects ±2g range), '0101b' (selects ±4g range), '1000b' (selects ±8g range), '1100b' (selects ±16g range). All other settings are irregular; if such a setting is used, ±2g range is selected. Default value of (0x0F) range (after reset) is '0011b'.

Table 35: g-range, register (0x0F)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	range	range	range	range
				<3>	<2>	<1>	<0>

#### 5.8 Bandwidths

**Register** (0x10) contains the selection of the bandwidth for filtered acceleration data. Settings for (0x10) bw are '00xxxb' (bandwidth = 7.81 Hz), '01000b' (bandwidth = 7.81 Hz), '01001b' (bandwidth = 15.63 Hz), '01010b' (bandwidth = 31.25 Hz), '01011b' (bandwidth = 62.5 Hz), '01100b' (bandwidth = 125 Hz), '01101b' (bandwidth = 250 Hz), '01111b' (bandwidth = 1000 Hz), '1xxxxb' (bandwidth = 1000 Hz). Default value of (0x10) bw (after reset) is '11111b'. It is recommended to actively use the range from '01000b' to '01111b' only in order to be compatible with future products.

Table 36: Bandwidths, register (0x10)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	bw <4>	bw <3>	bw <2>	bw <1>	bw <0>

#### 5.9 Power modes

**Register** (0x11) contains the configuration of the power modes. (0x11) suspend = '1' ('0') sets (resets) suspend mode; default value of (0x11) suspend is '0'.

(0x11) lower on = '1' ('0') sets (resets) lower mode, default value of (0x11)

(0x11) lowpower\_en = '1' ('0') sets (resets) low-power mode, default value of (0x11) lowpower\_en is '0'.

The settings for (0x11) sleep\_dur are '0000b' to '0101b' (sleep phase duration = 0.5 ms), '0110b' (sleep phase duration = 1 ms), '0111b' (sleep phase duration = 2 ms), '1000b' (sleep phase duration = 4 ms), '1001b' (sleep phase duration = 6 ms), '1010b' (sleep phase duration = 10 ms), '1011b' (sleep phase duration = 25 ms), '1100b' (sleep phase duration = 50 ms), '1101b' (sleep phase duration = 100 ms), '1110b' (sleep phase duration = 500 ms), '1111b' (sleep phase duration = 1 s). Default value of (0x11) sleep\_dur is '0000b'.

Table 37: Power modes, register (0x11)

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
su	ıspend	lowpower	reserved	sleep_	sleep_	sleep_	sleep_	reserved
		_en		dur<3>	dur<2>	dur<1>	dur<0>	



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## 5.10 Special control settings

**Register** (0x12) is reserved.

**Register** (0x13) contains settings for the configuration of the acceleration data acquisition and the data output format.

 $(0x13) \ data\_high\_bw = \ 0\ (\ 1\ )$  selects filtered (unfiltered) acceleration data to be written into the data registers (0x02) to (0x07). Default value of  $(0x13) \ data\_high\_bw$  is  $\ 0\ .$ 

(0x13) shadow\_dis = '0' ('1') enables (disables) the shadowing procedure. Shadowing means that the MSB register is updated by reading the corresponding LSB register. Default value of (0x13) shadow\_dis is '0'.

Table 38: Acceleration data acquisition & data output format, register (0x13)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
data_high	shadow	reserved	reserved	reserved	reserved	reserved	reserved
_bw	_dis						

**Register** (0x14) is the *softreset* register. A user-triggered reset (softreset) of the sensor is performed after writing '0xB6' to the softreset register. After that reset all registers return to their default values. Reading (0x14) softreset returns 0x00.

**Register (0x15)** is reserved.

## **5.11** Interrupt settings

**Registers** (0x16) and (0x17) contain the enable bits for the interrupts. Default value of each enable bit is '0'.

Table 39: Interrupt setting, register (0x16)

(0x16) Bit	Name	Description
Bit 7	flat_en	'1' ('0') enables (disables) flat interrupt
Bit 6	orient_en	'1' ('0') enables (disables) orientation interrupt
Bit 5	s_tap_en	'1' ('0') enables (disables) single tap interrupt
Bit 4	d_tap_en	'1' ('0') enables (disables) double tap interrupt
Bit 3	- reserved -	reserved
Bit 2	slope_en_z	'1' ('0') enables (disables) slope interrupt for z-axis
Bit 1	slope_en_y	'1' ('0') enables (disables) slope interrupt for y-axis
Bit 0	slope_en_x	'1' ('0') enables (disables) slope interrupt for x-axis

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Table 40: Interrupt setting, register (0x17)

(0x17) Bit	Name	Description
Bit 7	- reserved -	reserved
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	data_en	'1' ('0') enables (disables) new data interrupt
Bit 3	low_en	'1' ('0') enables (disables) low-g interrupt
Bit 2	high_en_z	'1' ('0') enables (disables) high-g interrupt for z-axis
Bit 1	high_en_y	'1' ('0') enables (disables) high-g interrupt for y-axis
Bit 0	high_en_x	′1′ (′0′) enables (disables) high-g interrupt for x-axis

Register (0x18) is reserved.

**Registers** (0x19) to (0x1B) contain the mapping of interrupts onto the interrupt pins. Default value of each mapping bit is '0'.

Table 41: Interrupt mapping, register (0x19)

(0x19) Bit	Name	Description
Bit 7	int1_flat	'1' ('0') maps (unmaps) flat interrupt to INT1 pin
Bit 6	int1_orient	'1' ('0') maps (unmaps) orientation interrupt to INT1 pin
Bit 5	int1_s_tap	'1' ('0') maps (unmaps) single tap interrupt to INT1 pin
Bit 4	int1_d_tap	'1' ('0') maps (unmaps) double tap interrupt to INT1 pin
Bit 3	- reserved -	reserved
Bit 2	int1_slope	'1' ('0') maps (unmaps) slope interrupt to INT1 pin
Bit 1	int1_high	'1' ('0') maps (unmaps) high-g interrupt to INT1 pin
Bit 0	int1_low	'1' ('0') maps (unmaps) low-g interrupt to INT1 pin

Table 42: Interrupt mapping, register (0x1A)

(0x1A) Bit	Name	Description
Bit 7	int2_data	'1' ('0') maps (unmaps) new data interrupt to INT2 pin
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	- reserved -	reserved
Bit 3	- reserved -	reserved
Bit 2	- reserved -	reserved
Bit 1	- reserved -	reserved
Bit 0	int1_data	'1' ('0') maps (unmaps) new data interrupt to INT1 pin

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Table 43: Interrupt mapping, register (0x1B)

(0x1B) Bit	Name	Description
Bit 7	int2_flat	'1' ('0') maps (unmaps) flat interrupt to INT2 pin
Bit 6	int2_orient	'1' ('0') maps (unmaps) orientation interrupt to INT2 pin
Bit 5	int2_s_tap	'1' ('0') maps (unmaps) single tap interrupt to INT2 pin
Bit 4	int2_d_tap	'1' ('0') maps (unmaps) double tap interrupt to INT2 pin
Bit 3	- reserved -	reserved
Bit 2	int2_slope	'1' ('0') maps (unmaps) slope interrupt to INT2 pin
Bit 1	int2_high	'1' ('0') maps (unmaps) high-g interrupt to INT2 pin
Bit 0	int2_low	'1' ('0') maps (unmaps) low-g interrupt to INT2 pin

**Registers (0x1C) and (0x1D)** are reserved.

**Register** (0x1E) contains the data source definition for those interrupts with selectable data source. Default value of each data source selection bit is '0'.

Table 44: Interrupt data source definition, register (0x1E)

(0x1E) Bit	Name	Description
Bit 7	- reserved -	reserved
Bit 6	- reserved -	reserved
Bit 5	int_src_data	'1' ('0') selects unfiltered (filtered) data for the new data
		interrupt
Bit 4	int_src_tap	'1' ('0') selects unfiltered (filtered) data for the single tap and
		double tap interrupts
Bit 3	- reserved -	reserved
Bit 2	int_src_slope	'1' ('0') selects unfiltered (filtered) data for the slope interrupt
Bit 1	int_src_high	'1' ('0') selects unfiltered (filtered) data for the high-g interrupt
Bit 0	int_src_low	'1' ('0') selects unfiltered (filtered) data for the low-g interrupt

**Register (0x1F)** is reserved.

**Register** (0x20) contains the behavioural configuration (electrical behaviour) of the interrupt pins. Default value of (0x20)  $int1_od$  and (0x20)  $int2_od$  is '0'. Default value of (0x20)  $int1_lv$  and (0x20)  $int2_lv$  is '1'.

Table 45: Electrical behaviour of interrupt pin, register (0x20)

(0x20) Bit	Name	Description
Bit 7	- reserved -	reserved
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	- reserved -	reserved
Bit 3	int2_od	'0' selects push-pull, '1' selects open drive for INT2 pin
Bit 2	int2_lvl	'0' ('1') selects active level '0' ('1') for INT2 pin
Bit 1	int1_od	'0' selects push-pull, '1' selects open drive for INT1
Bit 0	int1_lvl	'0' ('1') selects active level '0' ('1') for INT1 pin

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**Register** (0x21) contains the interrupt reset bit and the interrupt mode selection. Writing '1' to (0x21) reset int resets any latched interrupt.

The settings for (0x21) latch\_int are '0000b' (non-latched), '0001b' (temporary, 250 ms), '0010b' (temporary, 500 ms), '0011b' (temporary, 1 s), '0100b' (temporary, 2 s), '0101b' (temporary, 4 s), '0110b' (temporary, 8 s), '0111b' (latched), '1000b' (non-latched), '1001b' (temporary, 500  $\mu$ s), '1010b' (temporary, 1010b' (temporary, 1010b' (temporary, 1010b' (temporary, 1010b' (temporary, 1010b' (temporary, 500  $\mu$ s), '1110b' (temporary, 500  $\mu$ s), '1111b' (latched).

Default value of (0x21) latch\_int is '0000b'.

Table 46: Interrupt reset bit and interrupt mode selection, register (0x21)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reset_int	reserved	reserved	reserved	latch_ int<3>	latch_ int<2>	latch_ int<1>	latch_ int<0>

**Register** (0x22) contains the delay time definition for the low-g interrupt. The physical delay time can be computed from the content of  $(0x22) low_dur$  according to: delay [ms] =  $[(0x22) low_dur + 1] \cdot 2$  ms.

Possible delay times range from 2 ms to 512 ms. Default value of (0x22) low\_dur is 0x09, corresponding to a delay of 20 ms.

Table 47: Delay time definition for the low-g interrupt, register (0x22)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
low_							
dur<7>	dur<6>	dur<5>	dur<4>	dur<3>	dur<2>	dur<1>	dur<0>

**Register** (0x23) contains the threshold definition for the low-g interrupt. An LSB of (0x23)  $low_th$  corresponds to an actual acceleration of 7.81 mg. Therefore, the threshold ranges from 0 g to 1.992 g. Default value of (0x23)  $low_th$  is 0x30, corresponding to an acceleration of 375 mg.

Table 48: Threshold definition for the low-g interrupt, register (0x23)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
low_							
th<7>	th<6>	th<5>	th<4>	th<3>	th<2>	th<1>	th<0>

**Register** (0x24) contains the low-g interrupt mode selection, the low-g interrupt hysteresis setting, and the high-g interrupt hysteresis setting. Setting (0x24) low\_mode to '0' ('1') selects 'single' mode ('sum' mode). Default value is '0' ('single' mode).



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(0x24) low\_hy sets the hysteresis of the low-g interrupt. An LSB of (0x24) low\_hy corresponds to an acceleration difference of 125 mg. Default value of (0x24) low\_hy is '01b'.

(0x24) high\_hy sets the hysteresis of the high-g interrupt. The meaning of an LSB of (0x24) high\_hy depends on the selected g-range. It corresponds to an acceleration difference of 125 mg in 2g-range, 250 mg in 4g-range, 500 mg in 8g-range, and 1000mg in 16g-range. Default value of (0x24) high\_hy is '10b'.

Table 49: Threshold definition for the low-g interrupt, register (0x24)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
high_ hy<1>	high_ hy<0>	reserved	reserved	reserved	low_ mode	low_ hy<1>	low_ hy<0>

**Register** (0x25) contains the delay time definition for the high-g interrupt. The physical delay time can be computed from the content of (0x25)  $high\_dur$  according to delay [ms] = [(0x25)  $high\_dur + 1$ ] • 2 ms. Possible delay times range from 2 ms to 512 ms. Default value of (0x25)  $high\_dur$  is 0x0F, corresponding to a delay of 32 ms.

Table 50: Delay time definition for the high-g interrupt, register (0x25)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
high_							
dur<7>	dur<6>	dur<5>	dur<4>	dur<3>	dur<2>	dur<1>	dur<0>

**Register** (0x26) contains the threshold definition for the high-g interrupt. The meaning of an LSB of (0x26) high\_th depends on the selected g-range. It corresponds to 7.81 mg in 2g-range, 15.63 mg in 4g-range, 31.25 mg in 8g-range, and 62.5 mg in 16g-range. Default value of (0x26) high th is 0xC0.

Table 51: Threshold definition for the high-g interrupt, register (0x26)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
high_							
th<7>	th<6>	th<5>	th<4>	th<3>	th<2>	th<1>	th<0>

**Register** (0x27) contains the definition of the number of samples to be evaluated for the slope interrupt (any-motion detection). The number of samples is N = (0x27) slope\_dur + 1. Default value of (0x27) slope\_dur is '00b'.

Table 52: Samples number definition for the slope interrupt, register (0x27)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	reserved	reserved	slope_ dur<1>	slope_ dur<0>



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**Register** (0x28) contains the threshold definition for the slope interrupt. An LSB of (0x28) slope\_th corresponds to an LSB of acceleration data. Its meaning therefore depends on the selected g-range. Default value of (0x28) slope\_th is 0x14.

Table 53: Slope threshold for the slope interrupt, register (0x28)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
slope_							
th<7>	th<6>	th<5>	th<4>	th<3>	th<2>	th<1>	th<0>

**Register (0x29)** is reserved.

**Register** (0x2A) contains the timing definitions for the single tap and double tap interrupts.

(0x2A) tap\_quiet = '0' ('1') selects a quiet duration of 30 ms (20 ms). The default value of (0x2A) tap\_quiet is '0'.

(0x2A) tap\_shock = '0' ('1') selects a shock duration of 50 ms (75 ms). The default value of (0x2A) tap\_shock is '0'.

(0x2A)  $tap\_dur$  selects the length of the time window for the second shock event (for double tap detection). The settings for (0x2A)  $tap\_dur$  are '000b' (50 ms), '001b' (100 ms), '010b' (150 ms), '011b' (200 ms), '100b' (250 ms), '101b' (375 ms), '110b' (500 ms), '111b' (700 ms). The default value of (0x2A)  $tap\_dur$  is '100b'.

Table 54: Tap Quiet duration and tap shock duration, register (0x2A)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
tap_	tap_	reserved	reserved	reserved	tap_	tap_	tap_
quiet	shock				dur<2>	dur<1>	dur<0>

**Register** (0x2B) contains the definition of the number of samples to be processed after wake-up in low-power mode and the threshold definition for the single and double tap interrupts. (0x2B)  $tap\_samp$  selects the number of samples that are processed after wake-up in the low-power mode. The settings for (0x2B)  $tap\_samp$  are '00b' (2 samples), '01b' (4 samples), '10b' (8 samples), and '11b' (16 samples). Default value of (0x2B)  $tap\_samp$  is '00b'.

The meaning of an LSB of (0x2B)  $tap\_th$  depends on the selected g-range. It corresponds to an acceleration difference of 62.5mg in 2g-range, 125mg in 4g-range, 250mg in 8g-range, and 500mg in 16g-range. Default value of (0x2B)  $tap\_th$  is 0x0A.

Table 55: Samples number after wake-up and threshold tap interrupt, register (0x2B)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
tap_ samp<1>	tap_ samp<0>	reserved	tap_ th<4>	tap_ th<3>	tap_ th<2>	tap_ th<1>	tap_ th<0>



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**Register** (0x2C) contains the definition of hysteresis, blocking, and mode for the orientation interrupt. (0x2C) orient\_hyst sets the hysteresis of the orientation interrupt; 1 LSB always corresponds to 62.5 mg, in any g-range (i.e. increment is independent from g-range setting). Default value of (0x2C) orient\_hyst is '001b'.

(0x2C) orient\_blocking selects the kind of blocking that is used for the generation of the orientation interrupt. The settings for (0x2C) orient\_blocking are '00b' (no blocking), '01b' (theta blocking), '10b' (theta blocking or slope in any axis > 0.2 g), and '11b' (orient value not stable for at least 100 ms or theta blocking or slope in any axis > 0.4 g). Default value of (0x2C) orient blocking is '10b'.

(0x2C) orient\_mode sets the thresholds for switching between the different orientations. The settings for (0x2C) orient\_mode are '00b' (symmetrical), '01b' (high-asymmetrical), '10b' (low-asymmetrical), '11b' (symmetrical). Default value of (0x2C) orient\_mode is '00b'.

Table 56: Hysteresis, Blocking for Orientation Interrupt, Register (0x2C)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	orient_	orient_	orient_	orient_	orient_	orient_	orient_
	hyst<2>	hyst<1>	hyst<0>	blocking<1>	blocking<0>	mode<1>	mode<0>

**Register** (0x2D) contains the definition of the theta blocking angle for the orientation interrupt. (0x2D) orient\_theta defines a blocking angle between 0° and 44.8° as described in section "4.8.1.7 Orientation blocking". Default value of (0x2D) orient\_theta is 0x08.

Table 57: Theta blocking angle, register (0x2D)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	orient_	orient_	orient_	orient_	orient_	orient_
		theta<5>	theta<4>	theta<3>	theta<2>	theta<1>	theta<0>

**Register** (0x2E) contains the definition of the flat threshold angle for the flat interrupt. (0x2E) flat\_theta defines a blocking angle between 0° and 44.8° as described in section"4.8.8 Flat detection". Default value of (0x2E) flat\_theta is 0x08.

Table 58: Flat threshold angle, register (0x2E)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	flat_	flat_	flat_	flat_	flat_	flat_
		theta<5>	theta<4>	theta<3>	theta<2>	theta<1>	theta<0>

**Register** (0x2F) contains the definition of the flat hold time. (0x2F) flat\_hold\_time defines the time a new flat value has to be at least stable for before the interrupt is generated. The settings for (0x2F) flat\_hold\_time are '00b' (0), '01b' (512 ms), '10b' (1024 ms), '11b' (2048 ms). Default value of (0x2F) flat\_hold\_time is '01b'.



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Table 59: Flat threshold angle, register (0x2F)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	flat_hold_ time<1>	flat_hold_ time<0>	reserved	reserved	reserved	reserved

Register (0x30) and (0x31) are reserved.

#### 5.12 Self-test

**Register** (0x32) contains the settings for the activation of the sensor self-test.

(0x32) self\_test\_sign sets the sign of the electrostatic excitation. The settings for (0x32) self\_test\_sign are '0' (positive sign) and '1' (negative sign). Default value of (0x32) self test sign is '0'.

(0x32) self\_test\_axis defines the axis which shall be excited. Only one axis can be excited at the same time. The settings for (0x32) self\_test\_axis are '00b' (no self-test), '01' (x-axis), '10' (y-axis), and '11' (z-axis). Default value of (0x32) self\_test\_axis is '00b'.

Table 60: Sensor self-test, register (0x32)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	reserved	self_test _sign	self_test _axis<1>	self_test _axis<0>

#### **5.13 Non-volatile memory control (EEPROM control)**

**Register** (0x33) contains the control settings for the non-volatile memory (EEPROM). (0x33)  $nvm\_load$  is used to perform a user-defined image update. Writing '1' (0x33)  $nvm\_load$  starts the update procedure. The value '1' is kept as long as the update procedure runs, afterwards it is reset to '0'.

(0x33) nvm\_rdy contains the status of writing the EEPROM. (0x33) nvm\_rdy is '0' as long as writing the EEPROM endures, it is '1' if currently no write access is performed and, therefore, a new write access can be initiated.

Writing '1'to (0x33) nvm\_prog\_trig triggers writing the EEPROM. The EEPROM can only be written if it was unlocked before.

Writing '1' to (0x33) nvm\_prog\_mode unlocks the EEPROM.



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Table 61: EEPROM control settings, register (0x33)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	nvm_load	nvm_rdy	nvm_prog _trig	nvm_prog mode

# **5.14** Interface configuration

**Register** (0x34) contains the settings for the digital interfaces. Writing '1'to (0x34)  $i2c\_wdt\_en$  enables the watchdog at the SDI pin (= SDA for I<sup>2</sup>C) if I<sup>2</sup>C is selected. Default value of (0x34)  $i2c\_wdt\_en$  is '0'.

(0x34)  $i2c\_wdt\_sel$  selects the I<sup>2</sup>C data pad watchdog timer period. The settings for (0x34)  $i2c\_wdt\_sel$  are '0' (1 ms) and '1' (50 ms). Default value of (0x34)  $i2c\_wdt\_sel$  is '0'.

(0x34) spi3 selects the SPI mode. The settings for (0x34) spi3 are '0' (4-wire SPI) and '1' (3-wire SPI). Default value of (0x34) spi3 is '0'.

Table 62: EEPROM control settings, register (0x34)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	reserved	i2c_wdt	i2c_wdt	spi3
					_en	_sel	

**Register** (0x35) is reserved.

#### 5.15 Offset compensation

**Register** (0x36) contains settings for the offset compensation in general, for fast offset compensation, and for slow offset compensation. Writing '1' to (0x36) offset\_reset sets all offset compensation registers (0x38 to 0x3D) to zero.

Default value of (0x36) offset reset is '0'.

(0x36) cal\_trigger starts the fast compensation process for the specified axis. The settings for (0x36) cal\_trigger are '00b' (no axis selected), '01b' (x-axis), '10b' (y-axis), '11b' (z-axis). A non-zero value is kept until the fast compensation procedure is finished. Default value of (0x36) cal\_trigger is '00b'.

(0x36) cal\_rdy indicates the state of the fast compensation. (0x36) cal\_rdy is '0' when (0x36) cal\_trigger has a nonzero value, otherwise (0x36) cal\_rdy is '1'.

Writing '1' ('0') to (0x36)  $hp\_z\_en$  enables (disables) slow offset compensation for the z-axis. Writing '1' ('0') to (0x36)  $hp\_y\_en$  enables (disables) slow offset compensation for the y-axis.



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Writing '1' ('0') to (0x36)  $hp_x_en$  enables (disables) slow offset compensation for the x-axis. Default value for each of (0x36)  $hp_x_en$ , (0x36)  $hp_y_en$ , and (0x36)  $hp_x_en$  is '0', respectively.

Table 63: Offset compensation, fast offset compensation, register (0x36)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset	cal_	cal_	cal_rdy	reserved	hp_z_en	hp_y_en	hp_x_en
_reset	trigger<1>	trigger<0>					

**Register** (0x37) contains settings for the offset compensation in general, and for slow offset compensation. (0x37) offset\_target\_z sets the target value for the offset compensation of the z-axis.

(0x37) offset\_target\_y sets the target value for the offset compensation of the y-axis. (0x37) offset\_target\_x sets the target value for the offset compensation of the x-axis. The settings for (0x37) offset\_target\_x, (0x37) offset\_target\_y, and (0x37) offset\_target\_z are '00b' (0 g), '01b' (+1 g), '10b' (-1 g), and '11b' (0 g). Default value of each of (0x37) offset\_target\_x, (0x37) offset\_target\_y, and (0x37) offset\_target\_z is '00b', respectively.

(0x37) cut\_off defines the number of samples for comparison by the slow offset compensation. The settings for (0x37) cut\_off are '0' (8 samples) and '1' (16 samples). The default value of (0x37) cut\_off is '0'.

Table 64: Offset compensation, slow offset compensation, register (0x37)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	offset_tar	offset_tar	offset_tar	offset_tar	offset_tar	offset_tar	cut_off
	get_z<1>	get_z<0>	get_y<1>	get_y<0>	get_x<1>	get_x<0>	

**Register** (0x38) contains the compensation value for filtered data for the x-axis. The contents of each of the registers (0x38) to (0x3D) is added to the corresponding acceleration data; it can be set either automatically by one of the implemented compensation algorithms or manually. These registers are image registers of registers in the EEPROM; the content of the EEPROM is copied to them after every reset.

Table 65: Filtered data compensation for the x-axis, register (0x38)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset_	offset_	offset_	offset_	offset_	offset_	offset_	offset_
filt $x < 7 >$	filt $x < \overline{6} >$	filt $x < \overline{5} >$	filt $x < \overline{4} >$	filt $x < 3$	filt $x < \overline{2} >$	filt $x<\overline{1}>$	filt $x<0>$



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**Register (0x39)** contains the compensation value for filtered data for the y-axis.

Table 66: Filtered data compensation for the y-axis, register (0x39)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset_							
filt_y<7>	filt_y<6>	filt_y<5>	filt_y<4>	filt_y<3>	filt_y<2>	filt_y<1>	filt_y<0>

**Register** (0x3A) contains the compensation value for filtered data for the z-axis.

Table 67: Filtered data compensation for the z-axis, register (0x3A)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset_							
filt_z<7>	filt_z<6>	filt_z<5>	filt_z<4>	filt_z<3>	filt_z<2>	filt_z<1>	filt_z<0>

**Register (0x3B)** contains the compensation value for unfiltered data for the x-axis.

Table 68: Unfiltered data compensation for the x-axis, register (0x3B)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset_							
unfilt_x							
<7>	<6>	<5>	<4>	<3>	<2>	<1>	<0>

Register (0x3C) contains the compensation value for unfiltered data for the y-axis.

Table 69: Unfiltered data compensation for the x-axis, register (0x3C)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset_	offset_	offset_	offset_	offset_	offset_	offset_	offset_
unfilt_y	unfilt_y	unfilt_y	unfilt_y	unfilt_y	unfilt_y	unfilt_y	unfilt_y
<7>	<6>	<i>&lt;5&gt;</i>	<4>	<3>	<2>	<1>	<0>

**Register** (0x3D) contains the compensation value for unfiltered data for the z-axis.

Table 70: Unfiltered data compensation for the y-axis, register (0x3D)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
offset_							
unfilt_z							
<7>	<6>	<5>	<4>	<3>	<2>	<1>	<0>



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**Registers** (0x3E) and (0x3F) are image registers of registers in the EEPROM. They are not linked to any sensor-specific functionality.

# 6. Digital interfaces

The BMA250 supports two serial digital interface protocols for communication as a slave with a host device (when operating in general mode): SPI and I<sup>2</sup>C. The active interface is selected by the state of the Pin#11 (PS) 'protocol select' pin: '0' ('1') selects SPI (I<sup>2</sup>C). For details see section 4.2 Operational modes.

By default, SPI operates in the standard 4-wire configuration. It can be re-configured by software to work in 3-wire mode instead of standard 4-wire mode.

Both interfaces share the same pins. The mapping for each interface is given in the following table:

**Table 71: Mapping of the interface pins** 

Pin#	Name	use w/ SPI	use w/ I <sup>2</sup> C	Description
1	SDO	SDO	address	SPI: Data Output (4-wire mode) I <sup>2</sup> C: Used to set LSB of I <sup>2</sup> C address
2	SDx	SDI	SDA	SPI: Data Input (4-wire mode) Data Input / Output (3-wire mode) I <sup>2</sup> C: Serial Data
10	CSB	CSB	unused	Chip Select (enable)
12	SCx	SCK	SCL	SPI: Serial Clock I <sup>2</sup> C: Serial Clock

The following table shows the electrical specifications of the interface pins:

Table 72: Electrical specification of the interface pins

Parameter	Symbol	Condition	Min	Тур	Max	Units
PS Impedance	R <sub>TS</sub>		1			МΩ
for Tri-state Detection	C <sub>TS</sub>				10	pF
PS Impedance for Non-Tri-state	R <sub>NTS</sub>				5	kΩ
Pull-up Resistance	R <sub>up</sub>	Internal Pull-up Resistance to VDDIO	70	120	190	kΩ
Pull-down Resistance	R <sub>down</sub>	Internal Pull-down Resistance to GND	12	20	32	kΩ
Input Capacitance	C <sub>in</sub>			5	10	pF
I <sup>2</sup> C Bus Load Capacitance (max. drive capability)	C <sub>I2C_Load</sub>				400	pF

# 6.1 Serial peripheral interface (SPI)

The timing specification for SPI of the BMA250 is given in the following table:

Table 73: SPI timing

Parameter	Symbol	Condition	Min	Max	Units
Clock Frequency	f <sub>SPI</sub>	Max. Load on SDI or SDO = 25pF		10	MHz
SCK Low Pulse	t <sub>SCKL</sub>		20		ns
SCK High Pulse	t <sub>SCKH</sub>		20		ns
SDI Setup Time	t <sub>SDI setup</sub>		20		ns
SDI Hold Time	t <sub>SDI_hold</sub>		20		ns
SDO Output Delay	t <sub>SDO_OD</sub>	Load = 25pF		30	ns
	_	Load = 250pF,		40	ns
		$V_{DDIO} = 2.4V$			
CSB Setup Time	t <sub>CSB setup</sub>		20		ns
CSB Hold Time	t <sub>CSB_hold</sub>		40		ns

The following figure shows the definition of the SPI timings given in table 73:

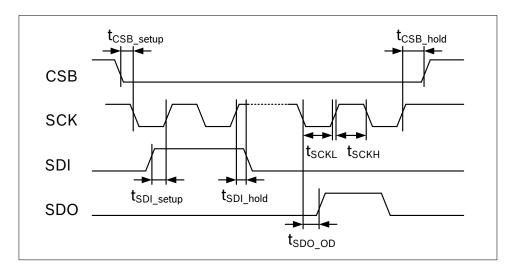


Figure 10: SPI timing diagram

The SPI interface of the BMA250 is compatible with two modes, '00' and '11'. The automatic selection between [CPOL = '0' and CPHA = '0'] and [CPOL = '1' and CPHA = '1'] is done based on the value of SCK after a falling edge of CSB.

Two configurations of the SPI interface are supported by the BMA250: 4-wire and 3-wire. The same protocol is used by both configurations. The device operates in 4-wire configuration by default. It can be switched to 3-wire configuration by writing '1' to (0x34) spi3. Pin SDI is used as the common data pin in 3-wire configuration.



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For single byte read as well as write operations, 16-bit protocols are used. The BMA250 also supports multiple-byte read operations.

In SPI 4-wire configuration CSB (chip select low active), SCK (serial clock), SDI (serial data input), and SDO (serial data output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI and SDO are driven at the falling edge of SCK and should be captured at the rising edge of SCK.

The basic write operation waveform for 4-wire configuration is depicted in figure 11. During the entire write cycle SDO remains in high- impedance state.

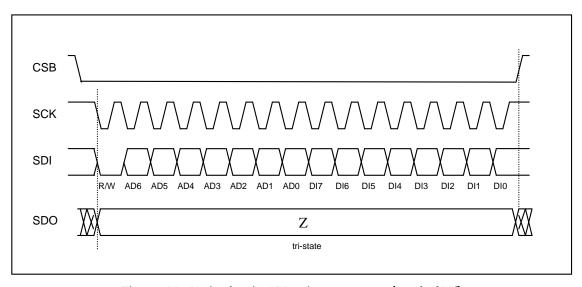


Figure 11: 4-wire basic SPI write sequence (mode '11')

The basic read operation waveform for 4-wire configuration is depicted in figure 12:

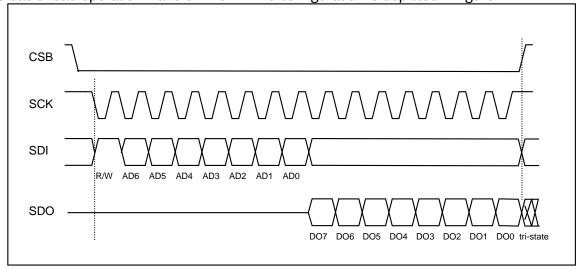


Figure 12: 4-wire basic SPI read sequence (mode '11')



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The data bits are used as follows:

Bit0: Read/Write bit. When 0, the data SDI is written into the chip. When 1, the data SDO from the chip is read.

Bit1-7: Address AD(6:0).

Bit8-15: when in write mode, these are the data SDI, which will be written into the address. When in read mode, these are the data SDO, which are read from the address.

Multiple read operations are possible by keeping CSB low and continuing the data transfer. Only the first register address has to be written. Addresses are automatically incremented after each read access as long as CSB stays active low.

The principle of multiple read is shown in figure 13:

			C	ontro	ol byt	е						Data	byte							Data	byte							Data	byte				
Start	RW		Re	gister	adre	ss (02	2h)			Da	ata re	gister	- adre	ess 02	2h	Data register - adress 03h Data register - adress 04h				Data register - adress 03h			Stop										
CSB																																	CSB
=	1	0	0	0	0	0	1	0	Х	Х	Χ	Х	Χ	Х	Х	Χ	Х	Х	Χ	Х	Х	Х	Х	Х	Х	Χ	Х	Х	Х	Х	Х	Х	=
0																																	1

Figure 13: SPI multiple read

In SPI 3-wire configuration CSB (chip select low active), SCK (serial clock), and SDI (serial data input and output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI is driven (when used as input of the device) at the falling edge of SCK and should be captured (when used as the output of the device) at the rising edge of SCK.

The protocol as such is the same in 3-wire configuration as it is in 4-wire configuration. The basic operation waveform (read or write access) for 3-wire configuration is depicted in figure 14:

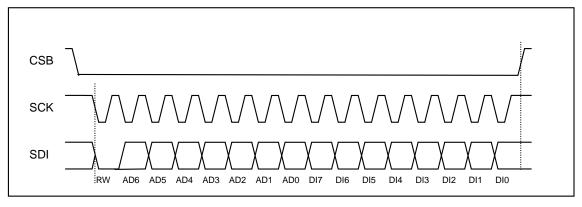


Figure 14: 3-wire basic SPI read or write sequence (mode '11')



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# 6.2 Inter-Integrated Circuit (I<sup>2</sup>C)

The I<sup>2</sup>C bus uses SCL (= SCx pin, serial clock) and SDA (= SDx pin, serial data input and output) signal lines. Both lines are connected to  $V_{\text{DDIO}}$  externally via pull-up resistors so that they are pulled high when the bus is free.

The I²C interface of the BMA250 is compatible with the I²C Specification UM10204 Rev. 03 (19 June 2007), available at http://www.nxp.com. The BMA250 supports I²C standard mode and fast mode, only 7-bit address mode is supported. For  $V_{DDIO}$  = 1.2V to 1.8V the guaranteed voltage output levels are slightly relaxed as described in the Parameter Specification (table 1).

The default I<sup>2</sup>C address of the device is 0011000b (0x18). It is used if the SDO pin is pulled to 'GND'. The alternative address 0011001b (0x19) is selected by pulling the SDO pin to ' $V_{DDIO}$ '.

The timing specification for I<sup>2</sup>C of the BMA250 is given in table 74:

Table 74: I<sup>2</sup>C timings

Parameter	Symbol	Condition	Min	Max	Units
Clock Frequency	f <sub>SCL</sub>			400	kHz
SCL Low Period	t <sub>LOW</sub>		1.3		
SCL High Period	t <sub>HIGH</sub>		0.6		
SDA Setup Time	t <sub>SUDAT</sub>		0.1		
SDA Hold Time	t <sub>HDDAT</sub>		0.0		
Setup Time for a repeated Start Condition	t <sub>SUSTA</sub>		0.6		
Hold Time for a Start Condition	t <sub>HDSTA</sub>		0.6		— μs
Setup Time for a Stop Condition	t <sub>SUSTO</sub>		0.6		
Time before a new Transmission can start	t <sub>BUF</sub>		1.3		



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Figure 15 shows the definition of the I<sup>2</sup>C timings given in table 74:

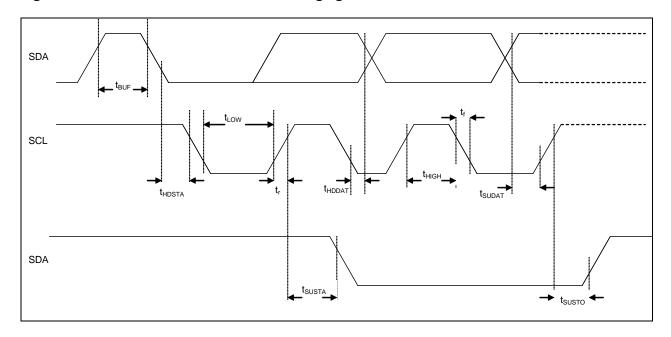


Figure 15: I<sup>2</sup>C timing diagram

The I2C protocol works as follows:

**START**: Data transmission on the bus begins with a high to low transition on the SDA line while SCL is held high (start condition (S) indicated by I<sup>2</sup>C bus master). Once the START signal is transferred by the master, the bus is considered busy.

**STOP**: Each data transfer should be terminated by a Stop signal (P) generated by master. The STOP condition is a low to HIGH transition on SDA line while SCL is held high.

**ACK**: Each byte of data transferred must be acknowledged. It is indicated by an acknowledge bit sent by the receiver. The transmitter must release the SDA line (no pull down) during the acknowledge pulse while the receiver must then pull the SDA line low so that it remains stable low during the high period of the acknowledge clock cycle.

In the following diagrams these abbreviations are used:

Start

Р	Stop
ACKS	Acknowledge by slave
ACKM	Acknowledge by master
NACKM	Not acknowledge by master

RW Read / Write

A START immediately followed by a STOP (without SCK toggling from logic "1" to logic "0") is not supported. If such a combination occurs, the STOP is not recognized by the device.

S



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#### I<sup>2</sup>C write access:

I<sup>2</sup>C write access can be used to write a data byte in one sequence.

The sequence begins with start condition generated by the master, followed by 7 bits slave address and a write bit (RW = 0). The slave sends an acknowledge bit (ACK = 0) and releases the bus. Then the master sends the one byte register address. The slave again acknowledges the transmission and waits for the 8 bits of data which shall be written to the specified register address. After the slave acknowledges the data byte, the master generates a stop signal and terminates the writing protocol.

Example of an I<sup>2</sup>C write access:

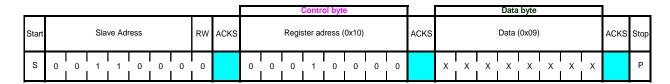


Figure 16: I<sup>2</sup>C write

#### I<sup>2</sup>C read access:

I<sup>2</sup>C read access also can be used to read one or multiple data bytes in one sequence.

A read sequence consists of a one-byte I<sup>2</sup>C write phase followed by the I<sup>2</sup>C read phase. The two parts of the transmission must be separated by a repeated start condition (Sr). The I<sup>2</sup>C write phase addresses the slave and sends the register address to be read. After slave acknowledges the transmission, the master generates again a start condition and sends the slave address together with a read bit (RW = 1). Then the master releases the bus and waits for the data bytes to be read out from slave. After each data byte the master has to generate an acknowledge bit (ACK = 0) to enable further data transfer. A NACKM (ACK = 1) from the master stops the data being transferred from the slave. The slave releases the bus so that the master can generate a STOP condition and terminate the transmission.

The register address is automatically incremented and, therefore, more than one byte can be sequentially read out. Once a new data read transmission starts, the start address will be set to the register address specified in the latest I<sup>2</sup>C write command. By default the start address is set at 0x00. In this way repetitive multi-bytes reads from the same starting address are possible.

In order to prevent the  $I^2C$  slave of the device to lock-up the  $I^2C$  bus, a watchdog timer (WDT) is implemented. The WDT observes internal  $I^2C$  signals and resets the  $I^2C$  interface if the bus is locked-up by the BMA250. The activity and the timer period of the WDT can be configured through the bits (0x34) i2c wdt en and (0x34) i2c wdt sel.

Writing '1' ('0') to (0x34)  $i2c\_wdt\_en$  activates (de-activates) the WDT. Writing '0' ('1') to (0x34)  $i2c\_wdt\_se$  selects a timer period of 1 ms (50 ms).



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# Example of an I2C read access:

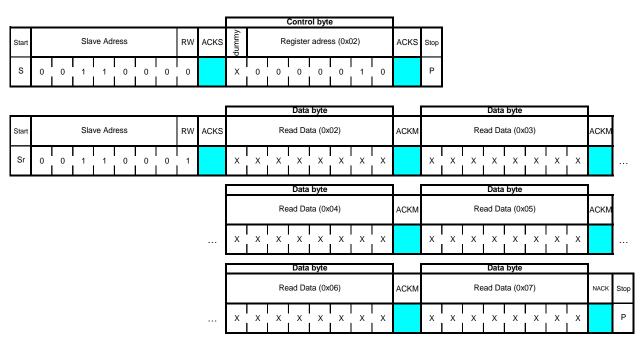


Figure 17: I2C multiple read



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# 7. Pin-out and connection diagram

## 7.1 Pin-out

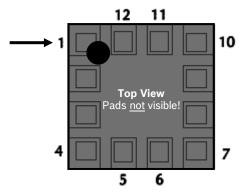


Figure 18: Pin-out top view

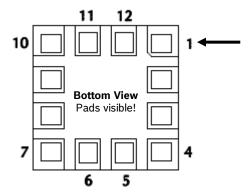


Figure 19 Pin-out bottom view

Table 75: Pin description

Pin#	Name	I/O Type	Description		Connect to	
				in SPI 4W	In SPI 3W	in I <sup>2</sup> C
1	SDO	Digital out	Serial data output in SPI Address select in I <sup>2</sup> C mode see chapter 6.2	SDO	DNC (float)	GND for default addr.
2	SDx	Digital I/O	SDA serial data I/O in I <sup>2</sup> C SDI serial data input in SPI 4W SDA serial data I/O in SPI 3W	SDI	SDA	SDA
3	VDDIO	Supply	Digital I/O supply voltage (1.2V 3.6V)	V <sub>DDIO</sub>	$V_{DDIO}$	V <sub>DDIO</sub>
4	NC			GND	GND	GND
5	INT1	Digital out	Interrupt output 1	INT1	INT1	INT1
6	INT2	Digital out	Interrupt output 2	INT2	INT2	INT2
7	VDD	Supply	Power supply for analog & digital domain (1.62V 3.6V)	V <sub>DD</sub>	$V_{DD}$	$V_{DD}$
8	GNDIO	Ground	Ground for I/O	GND	GND	GND
9	GND	Ground	Ground for digital & analog	GND	GND	GND
10	CSB	Digital in	Chip select for SPI mode	CSB	CSB	DNC (float)
11	PS	Digital in	Protocol select (GND = SPI, $V_{DDIO}$ = I <sup>2</sup> C, float = $\mu$ C-less). Pin must not float unless dedicated mode is used, see chapter 4.2.2	GND	GND	$V_{DDIO}$
12	SCx	Digital in	SCK for SPI serial clock SCL for I <sup>2</sup> C serial clock	SCK	SCK	SCL

# 7.2 Connection diagram 4-wire SPI

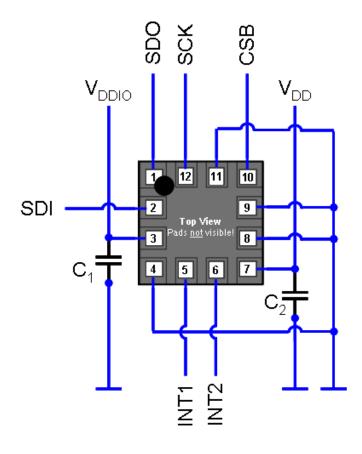
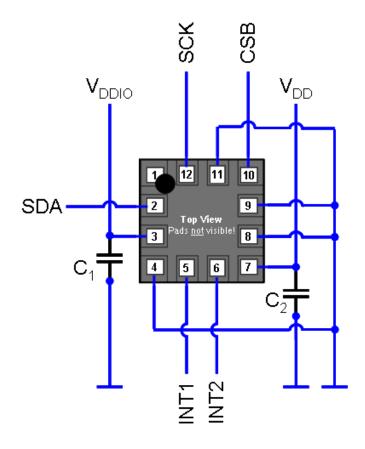


Figure 20: 4-wire SPI connection



# 7.3 Connection diagram 3-wire SPI



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Figure 21: 3-wire SPI connection

# 7.4 Connection diagram I<sup>2</sup>C

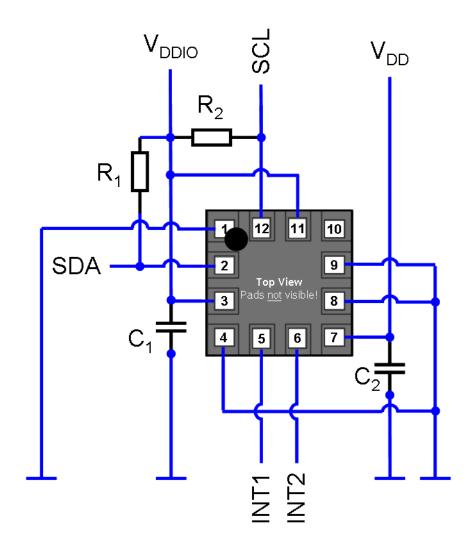


Figure 22: I<sup>2</sup>C connection

Note: the recommended value for  $C_1$ ,  $C_2$  is 100 nF.



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# 8. Package

#### 8.1 Outline dimensions

The sensor housing is a standard LGA package. It is compliant with JEDEC Standard MO-229 Type VGGD-3. Its dimensions are the following.

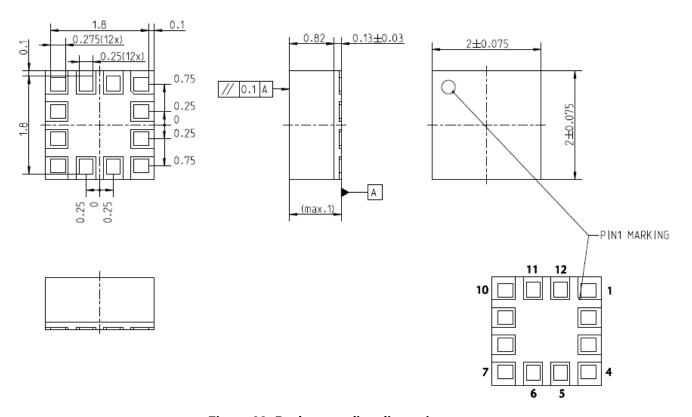


Figure 23: Package outline dimensions

# 8.2 Sensing axes orientation

If the sensor is accelerated in the indicated directions, the corresponding channel will deliver a positive acceleration signal (dynamic acceleration). If the sensor is at rest and the force of gravity is acting along the indicated directions, the output of the corresponding channel will be negative (static acceleration).

Example: If the sensor is at rest or at uniform motion in a gravity field according to the figure given below, the output signals are:

- ± 0g for the X channel
- ± 0g for the Y channel
- + 1g for the Z channel

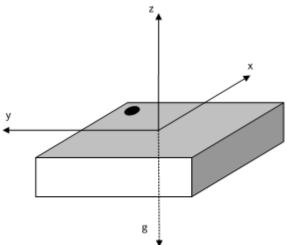


Figure 24: Orientation of sensing axis

The following table lists all corresponding output signals on X, Y, and Z while the sensor is at rest or at uniform motion in a gravity field under assumption of a  $\pm 2g$  range setting and a top down gravity vector as shown above.

Table 76: Output signals depending on sensor orientation

Sensor Orientation (gravity vector √)	•	•	•		upright	thgirqu
Output Signal X	Og / OLSB	1g/256LSB	Og / OLSB	-1g/-256LSB	0g / OLSB	0g / OLSB
Output Signal Y	-1g/-256LSB	0g / OLSB	+1g / 256LSB	0g / OLSB	0g / OLSB	0g / OLSB
Output Signal Z	Og / OLSB	0g / OLSB	Og / OLSB	0g / OLSB	1g/256LSB	-1g/-256LSB

# 8.3 Landing pattern recommendation

For the design of the landing patterns, we recommend the following dimensioning:

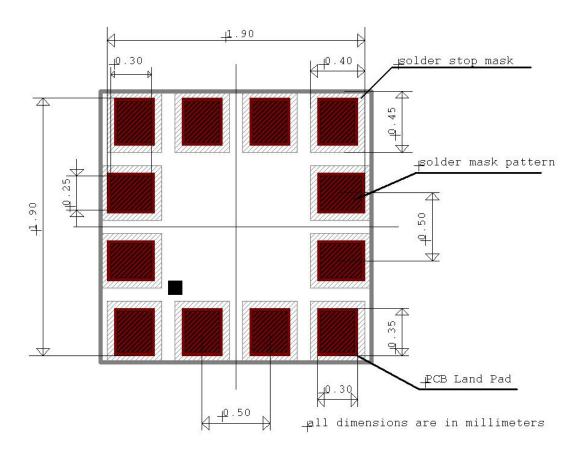


Figure 25: Landing patterns relative to the device pins, dimensions are in mm



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# 8.4 Marking

# 8.4.1 Mass production samples

**Table 77: Marking of mass production samples** 

Lal	beling	Name	Symbol	Remark
		Lot counter	ссс	3 alphanumeric digits, variable to generate mass production trace-code
	CCC	Product number	Т	1 alphanumeric digit, fixed to identify product type, T = "8"
	TL	Sub-con ID	L	1 alphanumeric digit, variable to identify sub-con (L = "A" or L = "U" or L = "P")
		Pin 1 identifier	•	

# 8.4.2 Engineering samples

**Table 78: Marking of engineering samples** 

Labeling	Name	Symbol	Remark		
	Eng. sample ID	N	1 alphanumeric digit, fixed to identify engineering sample, N = "e"		
XXN CC	Sample ID	XX	2 alphanumeric digits, variable to generate trace-code		
	Counter ID	СС	2 alphanumeric digits, variable to generate trace-code		
	Pin 1 identifier	•			

# 8.5 Soldering guidelines

The moisture sensitivity level of the BMA250 sensors corresponds to JEDEC Level 1, see also

- IPC/JEDEC J-STD-020C "Joint Industry Standard: Moisture/Reflow Sensitivity Classification for non-hermetic Solid State Surface Mount Devices"
- IPC/JEDEC J-STD-033A "Joint Industry Standard: Handling, Packing, Shipping and Use of Moisture/Reflow Sensitive Surface Mount Devices".

The sensor fulfils the lead-free soldering requirements of the above-mentioned IPC/JEDEC standard, i.e. reflow soldering with a peak temperature up to 260°C.

Profile Feature	Pb-Free Assembly
Average Ramp-Up Rate (Ts <sub>max</sub> to Tp)	3° C/second max.
Preheat  - Temperature Min (Ts <sub>min</sub> )  - Temperature Max (Ts <sub>max</sub> )  - Time (ts <sub>min</sub> to ts <sub>max</sub> )	150 °C 200 °C 60-180 seconds
Time maintained above:  - Temperature (T <sub>L</sub> )  - Time (t <sub>L</sub> )	217 °C 60-150 seconds
Peak/Classification Temperature (Tp)	260 °C
Time within 5 °C of actual Peak Temperature (tp)	20-40 seconds
Ramp-Down Rate	6 °C/second max.
Time 25 °C to Peak Temperature	8 minutes max.

Note 1: All temperatures refer to topside of the package, measured on the package body surface.

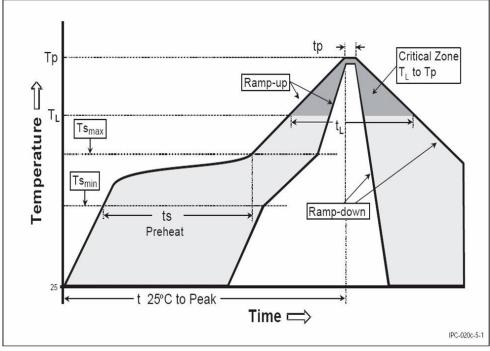


Figure 26: Soldering profile

# 8.6 Handling instructions

Micromechanical sensors are designed to sense acceleration with high accuracy even at low amplitudes and contain highly sensitive structures inside the sensor element. The MEMS sensor can tolerate mechanical shocks up to several thousand g's. However, these limits might be exceeded in conditions with extreme shock loads such as e.g. hammer blow on or next to the sensor, dropping of the sensor onto hard surfaces etc.

We recommend to avoid g-forces beyond the specified limits during transport, handling and mounting of the sensors in a defined and qualified installation process.

This device has built-in protections against high electrostatic discharges or electric fields (e.g. 2kV HBM); however, anti-static precautions should be taken as for any other CMOS component. Unless otherwise specified, proper operation can only occur when all terminal voltages are kept within the supply voltage range. Unused inputs must always be tied to a defined logic voltage level.

# 8.7 Tape and reel specification

The BMA250 is shipped in a standard cardboard box. The box dimension for 1 reel is:  $L \times W \times H = 35 \text{cm} \times 35 \text{cm} \times 6 \text{cm}$  BMA250 quantity: 10,000pcs per reel, please handle with care.

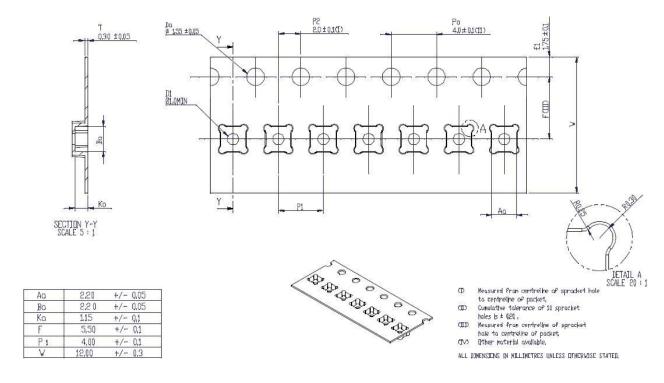


Figure 27: Tape and reel dimensions in mm



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#### 8.7.1 Orientation within the reel

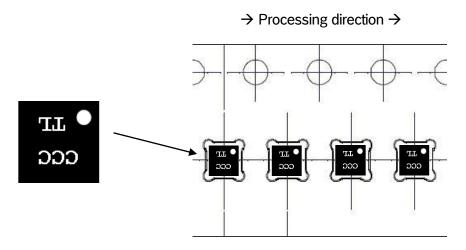


Figure 28: Orientation of the BMA250 devices relative to the tape

#### 8.8 Environmental safety

The BMA250 sensor meets the requirements of the EC restriction of hazardous substances (RoHS) directive, see also:

Directive 2002/95/EC of the European Parliament and of the Council of 27 January 2003 on the restriction of the use of certain hazardous substances in electrical and electronic equipment.

#### 8.8.1 Halogen content

Results of chemical analysis indicate that the BMA250 contains less than 900ppm (by weight) of Fluorine, Chlorine, Iodine and Bromine (i.e. < 50ppm per each substance). Therefore the BMA250 can be regarded as halogen-free. For more details on the analysis results please contact your Bosch Sensortec representative.

## 8.8.2 Internal package structure

Within the scope of Bosch Sensortec's ambition to improve its products and secure the mass product supply, Bosch Sensortec qualifies additional sources (e.g. 2<sup>nd</sup> source) for the LGA package of the BMA250.

While Bosch Sensortec took care that all of the technical packages parameters are described above are 100% identical for all sources, there can be differences in the chemical content and the internal structural between the different package sources.

However, as secured by the extensive product qualification process of Bosch Sensortec, this has no impact to the usage or to the quality of the BMA250 product.



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#### 9. Legal disclaimer

## 9.1 Engineering samples

Engineering Samples are marked with "e". Samples may vary from the valid technical specifications of the product series contained in this data sheet. They are therefore not intended or fit for resale to third parties or for use in end products. Their sole purpose is internal client testing. The testing of an engineering sample may in no way replace the testing of a product series. Bosch Sensortec assumes no liability for the use of engineering samples. The Purchaser shall indemnify Bosch Sensortec from all claims arising from the use of engineering samples.

#### 9.2 Product use

Bosch Sensortec products are developed for the consumer goods industry. They may only be used within the parameters of this product data sheet. They are not fit for use in life-sustaining or security sensitive systems. Security sensitive systems are those for which a malfunction is expected to lead to bodily harm or significant property damage. In addition, they are not fit for use in products which interact with motor vehicle systems.

The resale and/or use of products are at the purchaser's own risk and his own responsibility. The examination of fitness for the intended use is the sole responsibility of the Purchaser.

The purchaser shall indemnify Bosch Sensortec from all third party claims arising from any product use not covered by the parameters of this product data sheet or not approved by Bosch Sensortec and reimburse Bosch Sensortec for all costs in connection with such claims.

The purchaser must monitor the market for the purchased products, particularly with regard to product safety, and inform Bosch Sensortec without delay of all security relevant incidents.

## 9.3 Application examples and hints

With respect to any examples or hints given herein, any typical values stated herein and/or any information regarding the application of the device, Bosch Sensortec hereby disclaims any and all warranties and liabilities of any kind, including without limitation warranties of non-infringement of intellectual property rights or copyrights of any third party. The information given in this document shall in no event be regarded as a guarantee of conditions or characteristics. They are provided for illustrative purposes only and no evaluation regarding infringement of intellectual property rights or copyrights or regarding functionality, performance or error has been made.



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# 10. Document history and modification

Revision	Chapter	Description of modification/changes	Date
0.8		Document release	17 December 2010
0.9	1	Update table 1	26 January 2011
	4.2.2	Added missing table numbers	
	4.3	Update table 7	
	4.4.1	Update range register 0x0F	
1.0		Document rev. 1.0 update / no changes	03 March 2011
1.05	4.8.3	Typo correction, int1_od, int2_od	17 June 2011
	5.11	Typo correction, register 0x25	
	5.11	Typo correction in table 53 description	
1.10	5.2	Typo correction register map	02 November 2011
1.15	4.8.7	Update orientation interrupt	31 May 2012
	6.2	Update I2C address selection	

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