Dual LED Flash Driver with I²C-Compatible Interface

ADP1655

FEATURES

Ultracompact solution
- Small 2 mm × 1.5 mm 12-ball WLCSP package
- Tiny, low profile 2.2 µH power inductor
- LED current source for local LED grounding and low EMI
- Synchronous 2 MHz PWM boost convertor, no external diode

High efficiency: 88% peak
- Reduces high levels of input battery current during flash
- Limits battery current drain in torch mode

I²C programmable
- Currents up to 400 mA in flash mode for two LEDs
- Currents up to 500 mA in flash mode for one LED with 5% accuracy
- Currents up to 160 mA in torch mode with 10% accuracy
- Peak inductor current limit
- Flash timer

Control
- I²C-compatible control registers
- External STROBE pin
- External direct TORCH pin
- TX_MASK input to prevent high input battery current levels

Safety
- Thermal overload protection
- Flash timeout
- Inductor fault detection
- Output overvoltage
- Short circuit protection
- Soft start reduces inrush input current

APPLICATIONS

Camera-enabled cellular phones and smart phones
Digital still cameras, camcorders, and PDAs

GENERAL DESCRIPTION

The ADP1655 is a very compact, highly efficient, dual white LED flash driver for high resolution camera phones, which improves picture and video quality in low light environments. The device integrates a 2 MHz synchronous inductive boost convertor, an I²C-compatible interface and a 500 mA current source. The high switching frequency enables the use of a tiny, low profile 2.2 µH power inductor, and the current source permits LED cathode grounding for thermally enhanced, low EMI and compact layouts.

The efficiency is high over the entire battery voltage range to maximize the input power to LED power conversion and minimize battery curre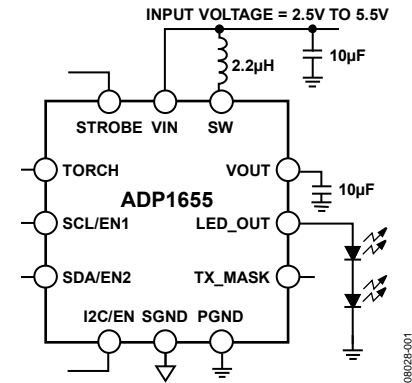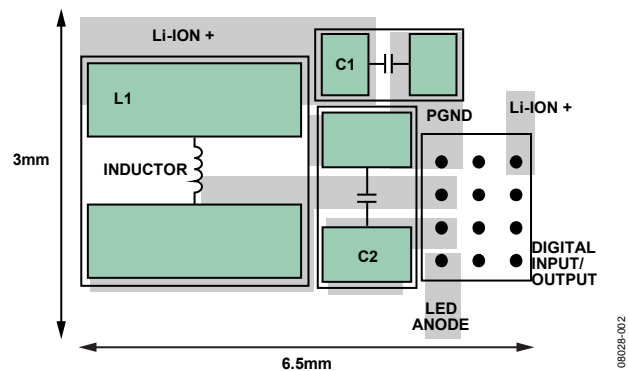nt draw during flash events. In addition, a Tx-mask input permits the flash LED current to reduce quickly and, therefore, the battery current reduces quickly, during a GSM power amplifier current burst.

The I²C-compatible interface enables the programmability of timers, currents, and status bit readback for operation monitoring and safety control.

The ADP1655 comes in a compact 12-ball 0.5 mm pitch WLCSP package and is specified over the full −40°C to +125°C junction temperature range.

FUNCTIONAL BLOCK DIAGRAM

Figure 1.

Figure 2. PCB Layout

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ADP1655

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REVISION HISTORY

5/09—Revison 0: Initial Version

SPECIFICATIONS

V_{IN} = 3.6 V, T_J = −40°C to +125°C for minimum/maximum specifications and T_A = 25°C for typical specifications, unless otherwise noted.

Table 1.

Parameter[1]	Conditions	Min	Typ	Max	Unit
SUPPLY					
Input Voltage Range		2.7		5.5	V
Undervoltage Lockout Threshold	V_{IN} falling	2.3	2.4	2.5	V
Hysteresis		50	100	150	mV
Shutdown Current	T_J = −40°C to +85°C, current into VIN pin, V_{IN} = 2.7 V to 4.5 V		0.3	1	µA
Standby Current	T_J = −40°C to +85°C, current into VIN pin, V_{IN} = 2.7 V to 4.5 V		3	10	µA
I2C/EN = SCL/EN1 = SDA/EN2 = 1.8 V					
Operating Quiescent Current	Torch mode, two LEDs, LED current = 40 mA		5.3		mA
SW Switch Leakage	T_J = −40°C to +85°C			1	µA
INPUTS					
Input Logic Low Voltage				0.54	V
Input Logic High Voltage		1.26			V
TORCH, STROBE, TX_MASK Pull-Down			350		kΩ
SCL/EN1, SDA/EN2 Pull-Down	I2C/EN = 0 V		350		kΩ
TORCH Glitch Filtering Delay	From TORCH rising edge to device start	6.3	9	11.7	ms
LED DRIVER					
LED Current					
Assist Light, Torch	I2C/EN = 0, one LED		80		mA
	I2C/EN = 0, two LEDs		40		mA
	I2C/EN = 1, assist light value setting = 0 (000 binary)		20		mA
	I2C/EN = 1, assist light value setting = 7 (111 binary)		160		mA
Flash	I2C/EN = 0, one LED		500		mA
	I2C/EN = 0, two LEDs		320		mA
	I2C/EN = 1, flash value setting = 0 (0000 binary)		200		mA
	I2C/EN = 1, one LED, flash value setting = 15 (1111 binary)		500		mA
	I2C/EN = 1, two LEDs, flash value setting = 10 to 15 (1010 to 1111 binary)		400		mA
LED Current Accuracy	I_{LED} = 320 mA to 500 mA	−5		+5	%
	I_{LED} = 60 mA to 320 mA	−5		+10	%
	I_{LED} = 20 mA to 60 mA	−5		+20	%
LED Current Source Headroom[2]	Flash typical, 400 mA LED current		290		mV
	Torch 160 mA		190		
LED_OUT Ramp-Up Time				1	ms
LED_OUT Ramp-Down Time				0.5	ms
Maximum Timeout For Flash			850		ms
Timer Accuracy		−7.5		+7.5	%
SWITCHING REGULATOR					
Switching Frequency		1.85	2	2.15	MHz
Minimum Duty Cycle			9.0		%
N-FET Resistance			135		mΩ
P-FET Resistance			290		mΩ

Parameter[1]	Conditions	Min	Typ	Max	Unit
SAFETY FEATURES					
Thermal Shutdown Threshold					
T_J Rising			150		°C
T_J Falling			140		°C
Overvoltage Threshold		9.0	9.5	10.1	V
Coil Peak Current Limit	Peak current value setting = 0 (00 binary)	1.13	1.25	1.38	A
	Peak current value setting = 1 (01 binary)	1.35	1.5	1.65	A
	Peak current value setting = 2 (10 binary)	1.58	1.75	1.93	A
	Peak current value setting = 3 (11 binary)	1.8	2.0	2.2	A
LED_OUT Short-Circuit Detection Comparator Reference Voltage			1.2	1.3	V
LED Counting Comparator Threshold Voltage					
	LED value setting = 0 (00 binary)		4.3		V
	LED value setting = 1 (01 binary)		4.6		V
	LED value setting = 2 (10 binary)		4.0		V
	LED value setting = 3 (11 binary)		4.9		V

[1] All limits at temperature extremes are guaranteed via correlation using standard statistical quality control (SQC).
[2] Two LEDs are used for this parameter.

RECOMMENDED SPECIFICATIONS: INPUT AND OUTPUT CAPACITANCE AND INDUCTANCE

Table 2.

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
CAPACITANCE	C_{MIN}					
Input		T_A = −40°C to +125°C	4.0			µF
Output		T_A = −40°C to +125°C	4.0		20	µF
MINIMUM AND MAXIMUM INDUCTANCE	L	T_A = −40°C to +125°C	1.5		2.8	µH

I²C-COMPATIBLE INTERFACE TIMING SPECIFICATIONS

Table 3.

Parameter[1]	Min	Max	Unit	Description
f_{SCL}		400	kHz	SCL clock frequency
t_{HIGH}	0.6		µs	SCL high time
t_{LOW}	1.3		µs	SCL low time
$t_{SU, DAT}$	100		ns	Data setup time
$t_{HD, DAT}$	0	0.9	µs	Data hold time
$t_{SU, STA}$	0.6		µs	Setup time for repeated start
$t_{HD, STA}$	0.6		µs	Hold time for start/repeated start
t_{BUF}	1.3		µs	Bus free time between a stop and a start condition
$t_{SU, STO}$	0.6		µs	Setup time for stop condition
t_R	$20 + 0.1\,C_B$[2]	300	ns	Rise time of SCL and SDA
t_F	$20 + 0.1\,C_B$	300	ns	Fall time of SCL and SDA
t_{SP}	0	50	ns	Pulse width of suppressed spike
C_B		400	pF	Capacitive load for each bus line

[1] Guaranteed by design.
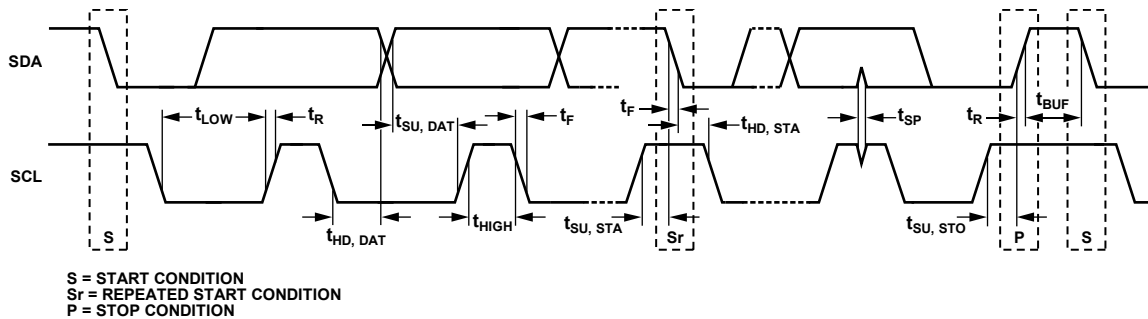[2] C_B is the total capacitance of one bus line in picofarads.

S = START CONDITION
Sr = REPEATED START CONDITION
P = STOP CONDITION

Figure 3. I²C-Compatible Interface Timing Diagram

ABSOLUTE MAXIMUM RATINGS

Table 4.

Parameter	Rating
VIN, SDA/EN2, SCL/EN1, I2C/EN, STROBE, TORCH, TX_MASK to SGND	−0.3 V to +6 V
LED_OUT, SW, VOUT to SGND	−0.3 V to +12 V
PGND to SGND	−0.3 V to +0.3 V
VOUT to LED_OUT	−0.3 V to +6 V
Ambient Temperature Range (T_A)	−40°C to +85°C
Junction Temperature Range (T_J)	−40°C to +125°C
Storage Temperature	JEDEC J-STD-020
ESD Human Body Model	±2000 V
ESD Charged Device Model	±1000 V
ESD Machine Model	±200 V

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL DATA

The ADP1655 may be damaged if the junction temperature limits are exceeded. Monitoring T_A does not guarantee that T_J is within the specified temperature limits. In applications with high power dissipation and poor thermal resistance, the maximum T_A may have to be derated. In applications with moderate power dissipation and low PCB thermal resistance, the maximum T_A can exceed the maximum limit as long as the T_J is within specification limits. T_J of the device is dependent on the T_A, the power dissipation (PD) of the device, and the junction-to-ambient thermal resistance (θ_{JA}) of the package. Maximum T_J is calculated from the T_A and PD using the following formula:

$$T_J = T_A + (PD \times \theta_{JA})$$

THERMAL RESISTANCE

θ_{JA} of the package is based on modeling and calculation using a 4-layer board. θ_{JA} is highly dependent on the application and board layout. In applications where high maximum power dissipation exists, attention to thermal board design is required. The value of θ_{JA} may vary, depending on PCB material, layout, and environmental conditions. The specified value of θ_{JA} is based on a 4-layer, 4 in × 3 in, 2 1/2 oz copper board, per JEDEC standards. For more information, see the AN-617 Application Note, *MicroCSP™ Wafer Level Chip Scale Package.*

θ_{JA} is specified for a device mounted on a JEDEC 2S2P PCB.

Table 3. Thermal Resistance

Package Type	θ_{JA}	Unit
12-Ball WLCSP	75	°C/W

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.
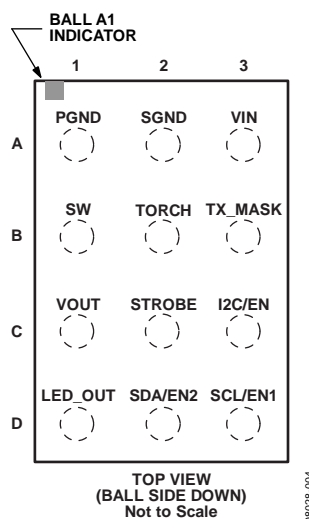
PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Figure 4. Pin Configuration

Table 5. Pin Function Descriptions

| Pin No.	Mnemonic	Type	Description
A1	PGND	Ground	Ground for Internal Switching FET.
A2	SGND	Ground	Connect this pin at a single point to the power ground.
A3	VIN	Supply	Connect the battery between VIN and PGND. Bypass VIN with a 10 µF, 6.3 V or greater X5R/X7R capacitor.
B1	SW	Output	Connect a 2.2 µH inductor between SW and the battery.
B2	TORCH	Digital Input	This pin enables the torch, provided that the device is not in flash or assist light mode.
B3	TX_MASK	Digital Input	Connect a digital signal to the TX_MASK pin. When the logic level is driven high during a flash event the current is reduced to the torch level.
C1	VOUT	Output	VOUT senses the output voltage of the boost converter and provides the input voltage to the LED current source. The VOUT pin features a comparator to detect an overvoltage condition if the LED string is open circuited. Connect a 10.0 µF capacitor between VOUT and PGND.
C2	STROBE	Digital Input/ Output	The STROBE input is used to synchronize the timing of the camera module to the LED driver in I2C-compatible interface mode. In 2-bit logic interface mode, this acts as an output, indicating the number of LEDs attached. STROBE = high indicates two LEDs, whereas STROBE = low indicates one LED.
C3	I2C/EN	Digital Input	A logic low selects the 2-bit logic interface, whereas logic high selects I2C-compatible interface. If I2C/EN is low and SDA/EN2 and SCL/EN1 are low, the driver enters shutdown mode with consumption < 1 µA.
D1	LED_OUT	Output	White LED Anode Connection. Connect LED_OUT to the anode of the white LED. LED_OUT is internally connected to a programmable PMOS current source, which regulates the LED current.
D2	SDA/EN2	Digital Input/ Output	Data Input/Output (SDA). In 2-bit logic interface mode, SDA/EN2 is the second input bit of the digital interface. Second Input Bit (EN2). In I2C mode, SDA is the data input/output of the I2C-compatible interface.
D3	SCL/EN1	Digital Input	Clock Input (SCL). In 2-bit logic interface mode, SCL/EN1 is the first input bit of the digital interface. First Input Bit (EN1). In I2C mode, SCL is the clock input of the I2C-compatible interface.

TYPICAL PERFORMANCE CHARACTERISTICS
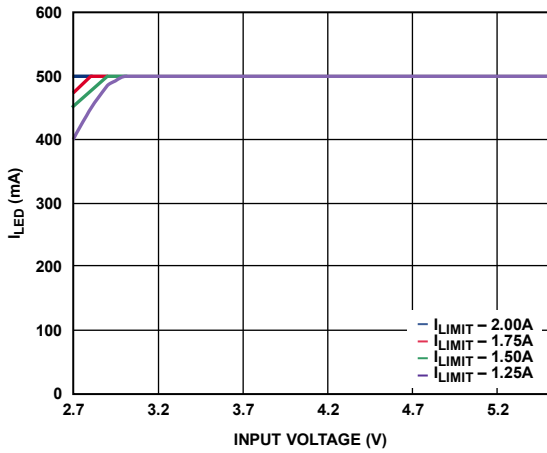
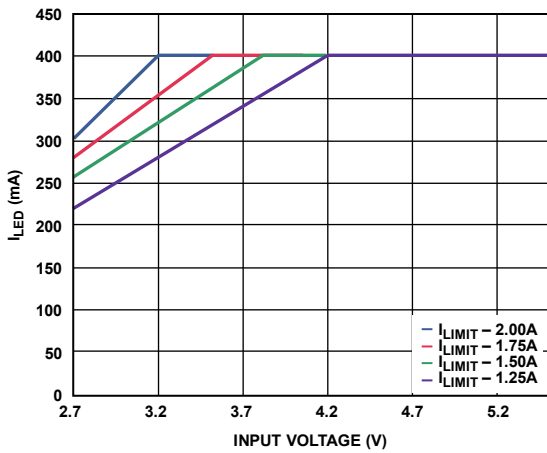Figure 5. Maximum Current vs. Input Voltage, One LED

Figure 6. Maximum Current vs. Input Voltage, Two LEDs;
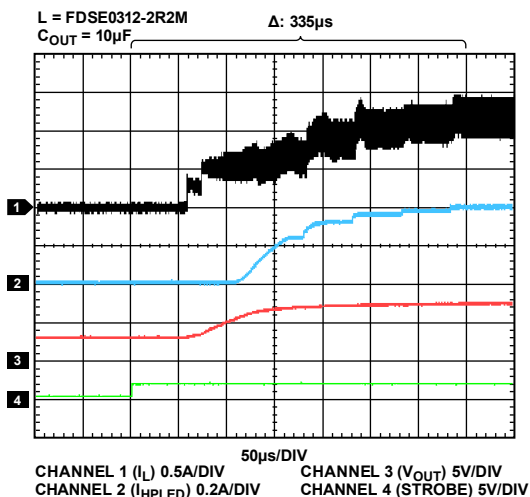LED Forward Voltage (Vf) = 4.3 V for each LED

CHANNEL 1 (I_L) 0.5A/DIV   CHANNEL 3 (V_OUT) 5V/DIV
CHANNEL 2 (I_HPLED) 0.2A/DIV   CHANNEL 4 (STROBE) 5V/DIV

Figure 7. Startup, Two LEDs Flash Mode, I_LED = 400 mA, V_IN = 3.2 V

CHANNEL 1 (I_L) 0.5A/DIV   CHANNEL 3 (V_OUT) 5V/DIV
CHANNEL 2 (I_HPLED) 0.2A/DIV   CHANNEL 4 (STROBE) 5V/DIV
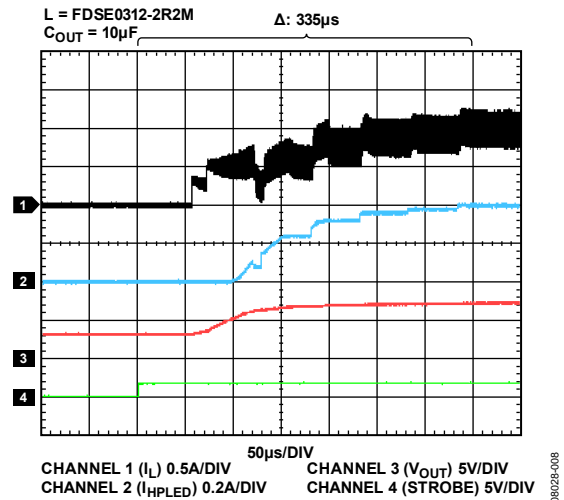
Figure 8. Startup, Two LEDs Flash Mode, I_LED = 400 mA, V_IN = 3.6 V

CHANNEL 1 (I_L) 0.5A/DIV   CHANNEL 3 (V_OUT) 5V/DIV
CHANNEL 2 (I_HPLED) 0.2A/DIV   CHANNEL 4 (SCL) 5V/DIV
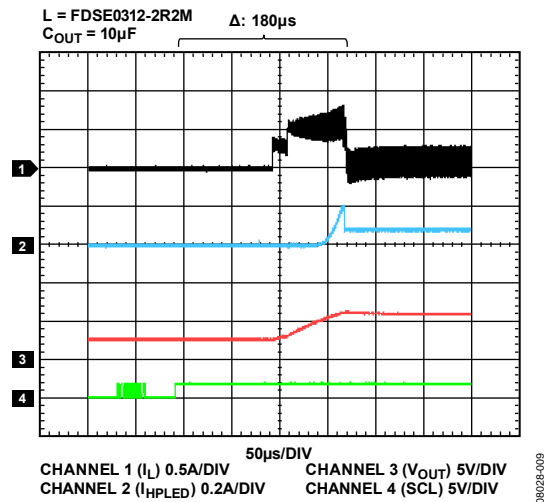
Figure 9. Startup, Two LEDs Assist Light Mode, I_LED = 40 mA, V_IN = 3.2 V

CHANNEL 1 (I_L) 0.5A/DIV   CHANNEL 3 (V_OUT) 5V/DIV
CHANNEL 2 (I_HPLED) 0.2A/DIV   CHANNEL 4 (SCL) 5V/DIV

Figure 10. Startup, Two LEDs Torch Mode, I_LED = 40 mA, V_IN = 3.6 V
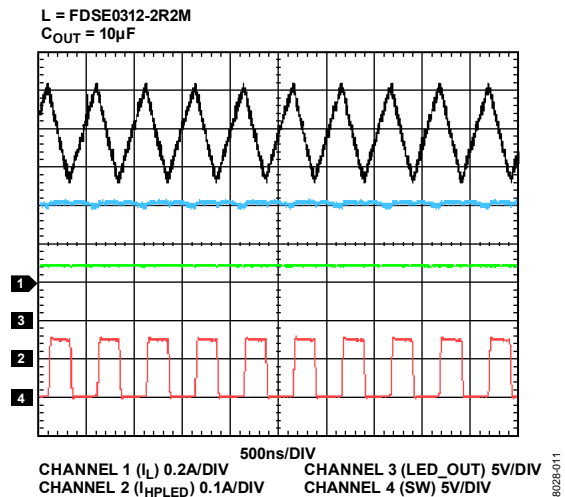
Figure 11. Inductor Current, Two LEDs Flash Mode, I_{LED} = 400 mA, V_{IN} = 3.6 V
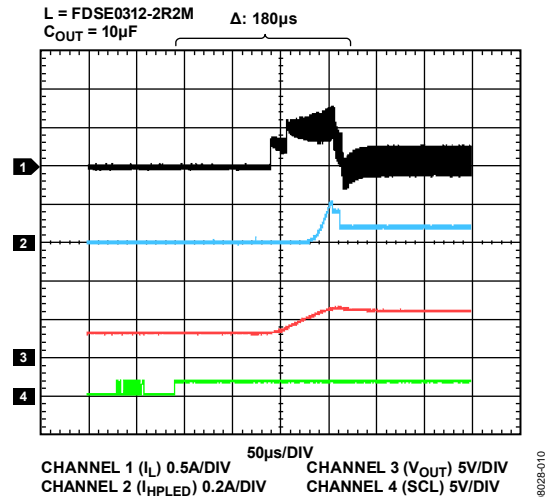

Figure 12. Inductor Current, Two LEDs Torch Mode, I_{LED} = 40 mA, V_{IN} = 3.6 V

Figure 13. LED Current Accuracy vs. Output Current

Figure 14. Efficiency P_{LED}/P_{IN}, Two High Power White LEDs in Series

Figure 15. Efficiency P_{LED}/P_{IN}, One High Power White LED
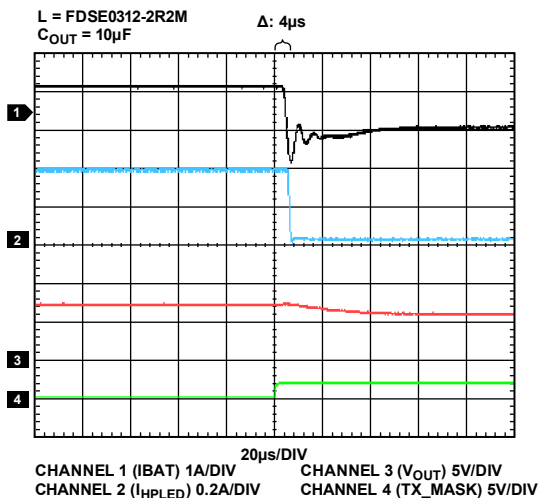
Figure 16. Tx Masking Response, TX_MASK = 0 V to 1.8 V,
I_{LED} = 40 mA to 400 mA, V_{IN} = 3.2 V

L = FDSE0312-2R2M
C_OUT = 10µF
Δ: 260µs

CHANNEL 1 (IBAT) 1A/DIV   CHANNEL 3 (V_OUT) 5V/DIV
CHANNEL 2 (I_HPLED) 0.2A/DIV   CHANNEL 4 (TX_MASK) 5V/DIV
100µs/DIV
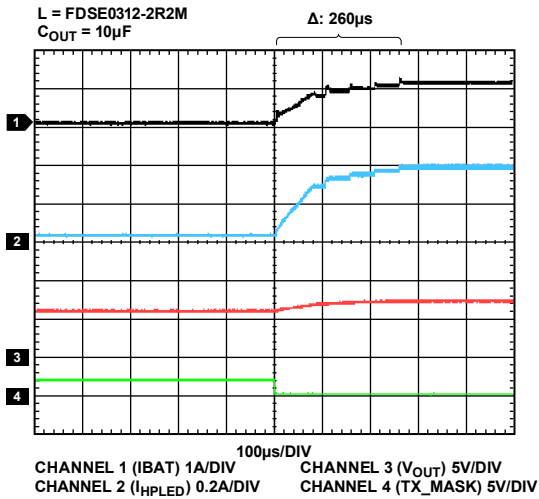
Figure 17. Tx Masking Response, TX_MASK = 1.8 V to 0 V,
I_LED = 40 mA to 400 mA, V_IN = 3.2 V

L = FDSE0312-2R2M
C_OUT = 10µF
Δ: 15mA

CHANNEL 1 (V_IN) 0.5V/DIV
CHANNEL 2 (I_HPLED) 20mA/DIV
100µs/DIV
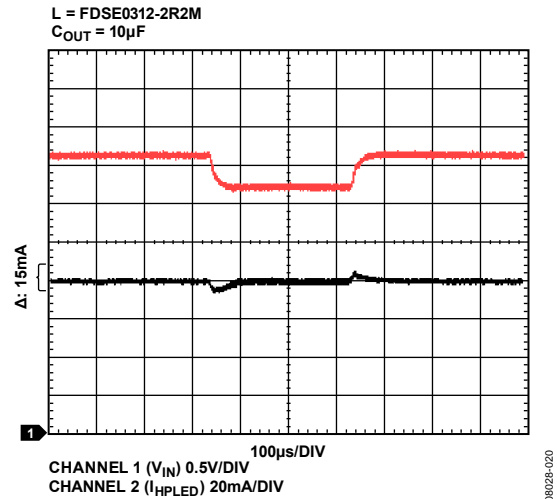
Figure 20. Line Transient, V_IN = 3.2 V to 3.6 V, I_LED = 400 mA

CHANNEL 1 (I_HPLED) 0.1A/DIV   CHANNEL 3 (SCL) 5V/DIV
CHANNEL 2 (STROBE) 1V/DIV
200ms/DIV
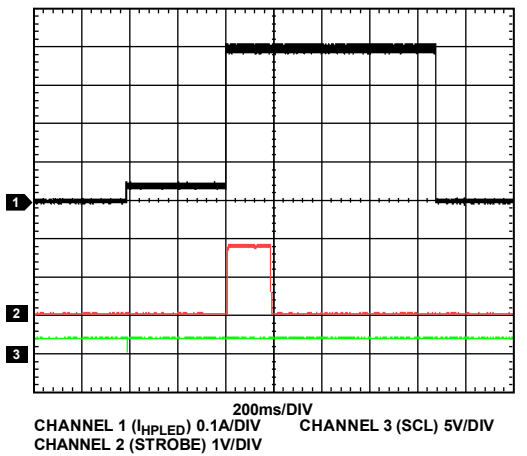
Figure 18. Assist Light and Flash, STROBE Edge Sensitive Mode, Two LEDs,
Timer = 850 ms, I_LED = 40 mA to 400 mA, V_IN = 3.6 V
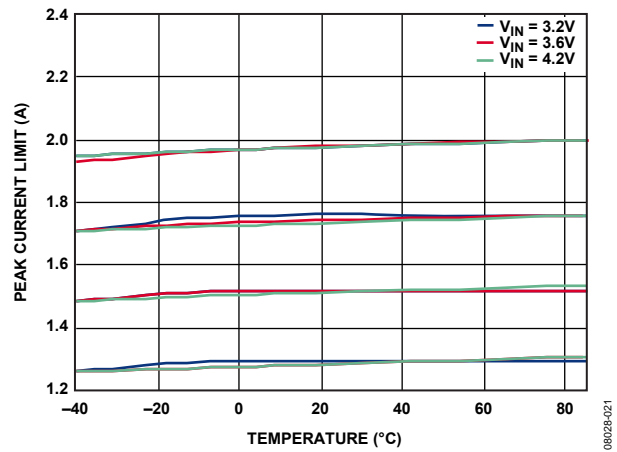
Figure 21. Coil Peak Current Limit vs. Temperature, Output Mode Register =
00, 01, 10, and 11 (Binary)

CHANNEL 1 (I_HPLED) 0.1A/DIV   CHANNEL 3 (SCL) 5V/DIV
CHANNEL 2 (STROBE) 1V/DIV
200ms/DIV
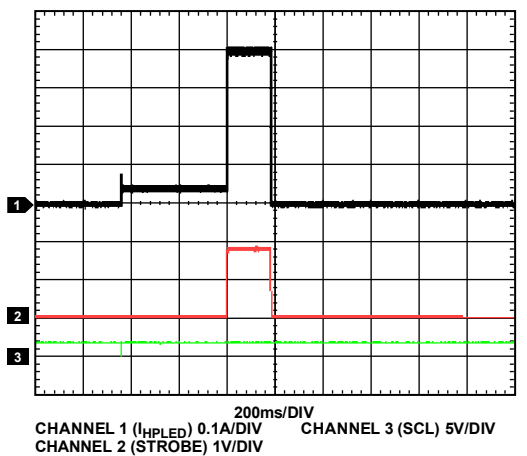
Figure 19. Assist Light and Flash, STROBE Level Sensitive Mode, Two LEDs,
I_LED = 40 mA to 400 mA, V_IN = 3.6 V
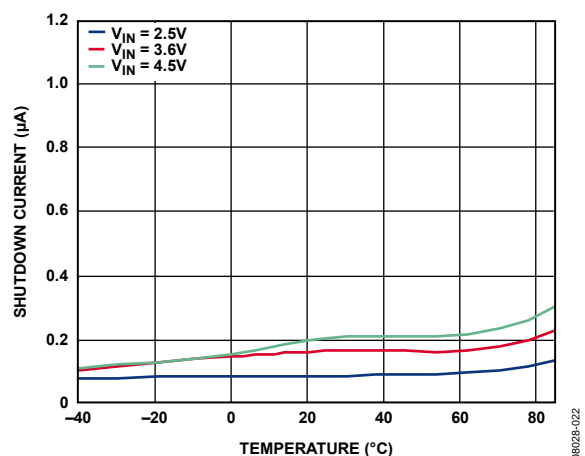
Figure 22. Shutdown Current vs. Temperature vs. V_IN
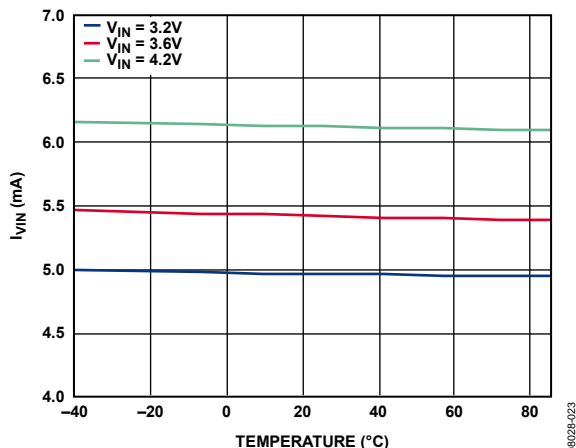
Figure 23. Operating Quiescent Current vs. Temperature, Torch Mode

Figure 24. Standby Current vs. Temperature vs. V_{IN},
I2C/EN = SCL/EN1 = SDA/EN2 = 1.8 V

Figure 25. Switching Frequency vs. Temperature vs. V_{IN}

Figure 26. LED Regulation, Set at 40 mA,
Current Set Register = 001 (Binary)

Figure 27. LED Regulation, Set at 400 mA,
Current Set Register = 1010 (Binary)

THEORY OF OPERATION

The ADP1655 is a high power, white LED driver ideal for driving white LEDs for use as a camera flash. The ADP1655 includes a boost converter and a current regulator suitable for powering one or two high power, white LEDs.

The ADP1655 responds to a 2-pin control interface that can operate in two separate pin-selectable modes: tying the I2C/EN pin high enables the I²C interface; tying the I2C/EN pin low enables a 2-bit logic interface.

WHITE LED DRIVER

The ADP1655 drives a synchronous boost converter to power one or two series-connected, high power LEDs. The white LED driver regulates the high power LED current for accurate brightness control. The ADP1655 uses an integrated PFET current regulator.

When the white LED is turned on, the step-up converter output voltage slew is limited to prevent excessive battery current while charging the output capacitor. The output voltage of the boost converter is sensed at VOUT. If the output voltage exceeds the 9.5 V (typical) limit, the white LED driver turns off and indicates that a fault condition has occurred through the system registers. This feature prevents damage due to an overvoltage if the white LED string fails with an open-circuit condition.

Setting the LED regulation currents depends on the 2-pin control interface used.

ASSIST LIGHT AND TORCH MODES

The ADP1655 features a programmable assist light mode that provides continuous LED current. The STROBE pin or the 2-bit logic interface can be used to transition from assist light mode directly to flash mode. The TORCH pin provides an alternative means of accessing a continuous LED current mode of operation. Both assist light and torch modes deliver the same current, which is programmable via the I²C-compatible interface.

Figure 28. Detailed Block Diagram

2-BIT LOGIC INTERFACE MODE (I2C/EN = 0)

In 2-bit logic interface mode, the two control pins, EN1 and
EN2, select whether the part is disabled or operating in assist
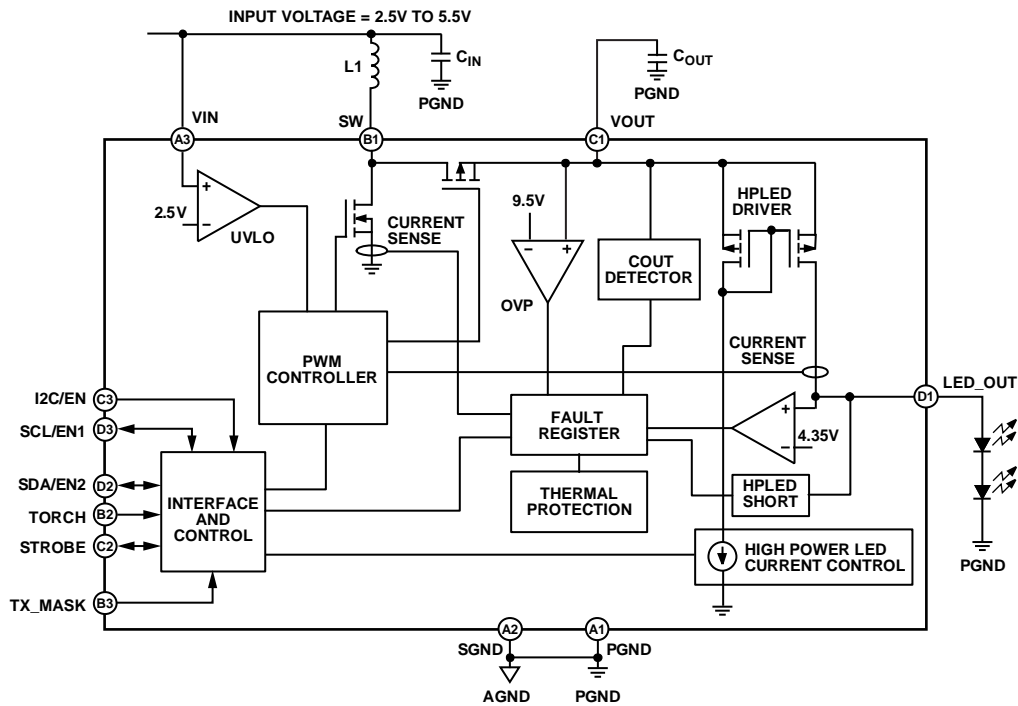light mode or flash mode, as outlined in Table 6. Additionally,
the TORCH pin selects torch mode.

Figure 29 illustrates state transitions of 2-bit logic mode con-
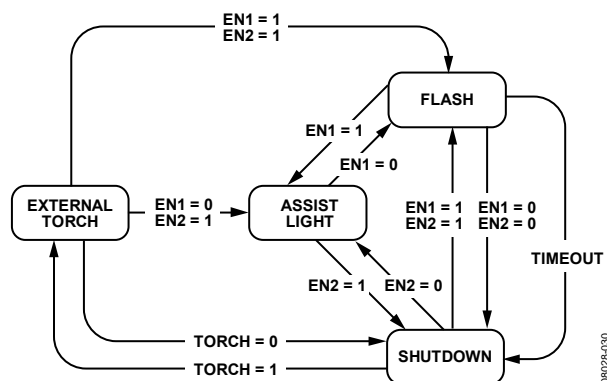trolled by digital inputs EN1, EN2, TORCH, and TX_MASK.

Figure 29. 2-Bit Logic Mode State Transitions (I2C/EN = 0)

When the ADP1655 is in flash mode, the TX_MASK pin can
be used to reduce the battery load. The device remains in flash
mode, but the LED driver output current is reduced to the assist
light level.

Table 6. 2-Bit Logic Interface Mode Selection

Mode	I2C/EN	EN1	EN2	TORCH	Output Current
Shutdown	0	0	0	0	0 mA
Torch	0	0	0	1	One LED: 80 mA Two LEDs: 40 mA
Assist light	0	0	1	X	One LED: 80 mA Two LEDs: 40 mA
Reserved	0	1	0	X	0 mA
Flash	0	1	1	X	One LED: 500 mA Two LEDs: 320 mA

I²C INTERFACE MODE (I2C/EN = 1)

The ADP1655 includes an I²C-compatible serial interface for
control of the LED current, as well as for a readback of system
status registers. The I²C chip address is 0x60 in write mode and
0x61 in read mode.

Table 7. I²C Interface Mode Selection

Mode	I2C/EN	SCL	SDA	TORCH	Output Current
Standby	1	X	X	0	0 mA
Torch	1	X	X	1	20 mA to 160 mA[1,2]
Assist light	1	X	X	X	20 mA to 160 mA[2]
Flash	1	X	X	X	200 mA to 500 mA[2]

[1] Torch mode has to be enabled from Register 0x04.
[2] The output current value depends on the register settings.

Registers values are reset to the default values when VIN supply
falls below the undervoltage (UVLO) level.

Figure 30 illustrates the I²C write sequence to a single register.
The subaddress content selects which of the five ADP1655
registers is written to first. The ADP1655 sends an acknowl-
edgement to the master after the 8-bit data byte has been written.
The ADP1655 increments the subaddress automatically and starts
receiving a data byte to the following register until the master
sends an I²C stop as shown in Figure 31. Figure 32 shows the
I²C read sequence of a single register. ADP1655 sends the data
from the register denoted by the subaddress and increments
the subaddress automatically, sending data from the next reg-
ister until the master sends an I²C stop condition as shown in
Figure 33.

State transitions between standby, assist light, flash, and
external torch modes are described in the State Transitions
section and Figure 34.

The register definitions are shown in the I²C Register Map
section. The lowest bit number (0) represents the least
significant bit, and the highest bit number (7) represents
the most significant bit.

0 = WRITE

MASTER STOP

CHIP ADDRESS | ADP1655 ACK | SUBADDRESS | ADP1655 ACK | ADP1655 RECEIVES DATA | ADP1655 ACK

Figure 30. I²C Single Register Write Sequence

0 = WRITE

MASTER STOP

CHIP ADDRESS | ADP1655 ACK | SUBADDRESS REGISTER N | ADP1655 ACK | ADP1655 RECEIVES DATA TO REGISTER N | ADP1655 ACK | ADP1655 RECEIVES DATA TO REGISTER N + 1 | ADP1655 ACK | ADP1655 RECEIVES DATA TO LAST REGISTER | ADP1655 ACK
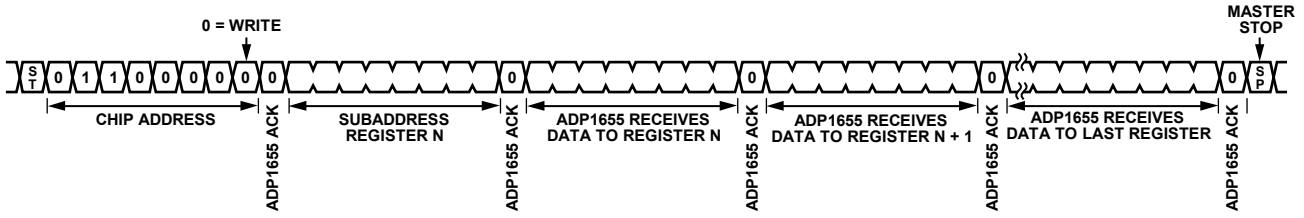
Figure 31. I²C Multiple Register Write Sequence

0 = WRITE

1 = READ

MASTER STOP

CHIP ADDRESS | ADP1655 ACK | SUBADDRESS | ADP1655 ACK | CHIP ADDRESS | ADP1655 ACK | ADP1655 SENDS DATA | MASTER ACK
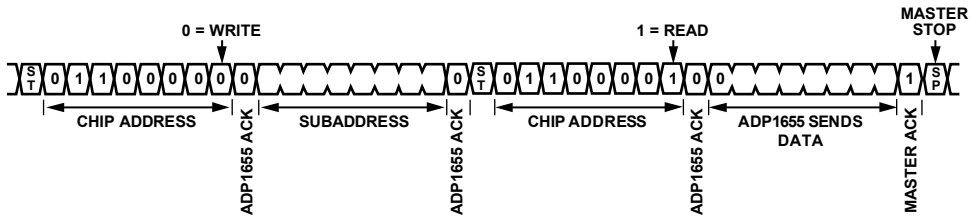
Figure 32. I²C Single Register Read Sequence

0 = WRITE

1 = READ

MASTER STOP

CHIP ADDRESS | ADP1655 ACK | SUBADDRESS REGISTER N | ADP1655 ACK | CHIP ADDRESS | ADP1655 ACK | ADP1655 SENDS DATA OF REGISTER N | MASTER ACK | ADP1655 SENDS DATA OF REGISTER N + 1 | MASTER ACK | ADP1655 SENDS DATA OF LAST REGISTER | MASTER ACK
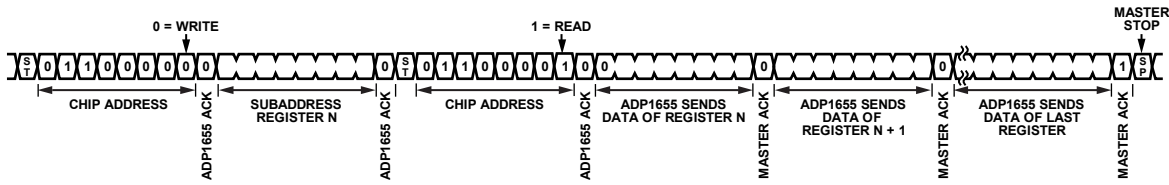
Figure 33. I²C Multiple Register Read Sequence

STATE TRANSITIONS

When the ADP1655 is in flash mode, the TX_MASK pin can be used to reduce the battery load. The device remains in flash mode, but the LED driver output current is reduced to the assist light level. In Figure 34, if the flash was triggered by the strobe pin in level-sensitive mode, a timeout triggers a timeout fault, as defined in the Safety Features section.
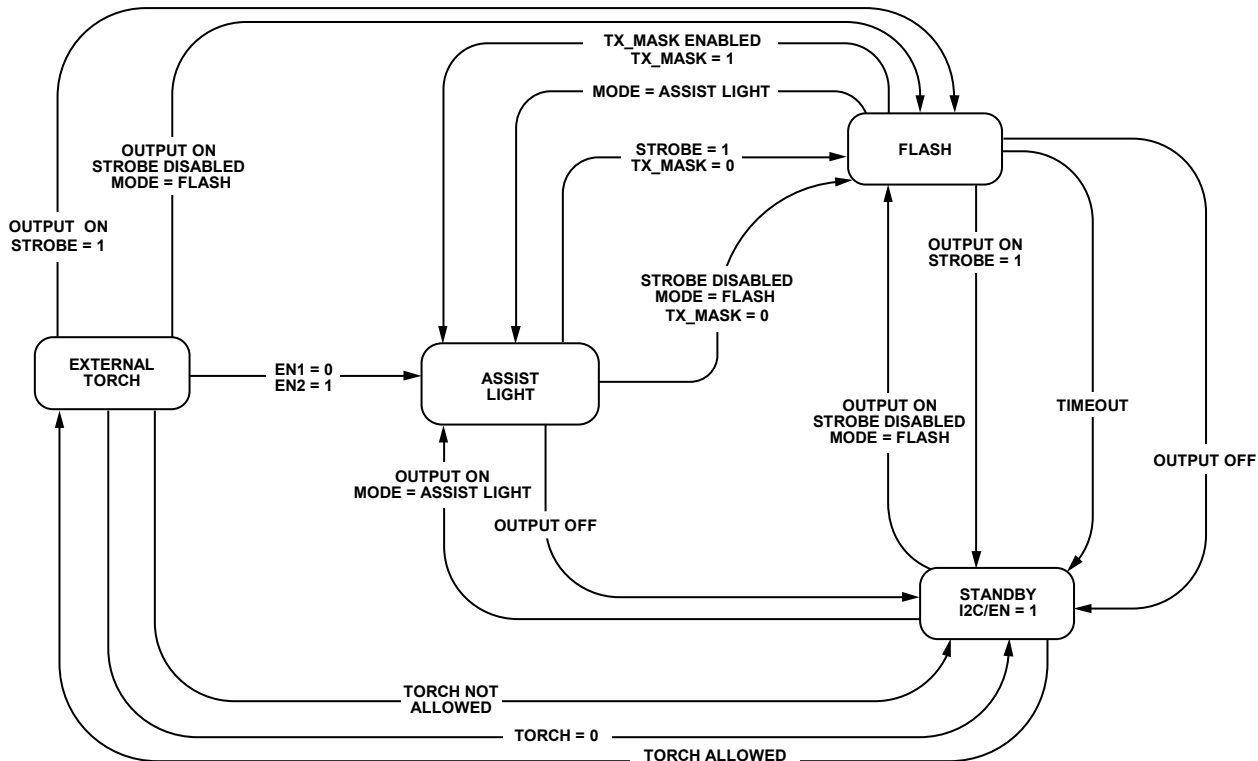
Figure 34. I²C Interface Mode: State Transitions

ADP1655

I²C REGISTER MAP

The lowest bit number (0) represents the least significant bit, and the highest bit number (7) represents the most significant bit.
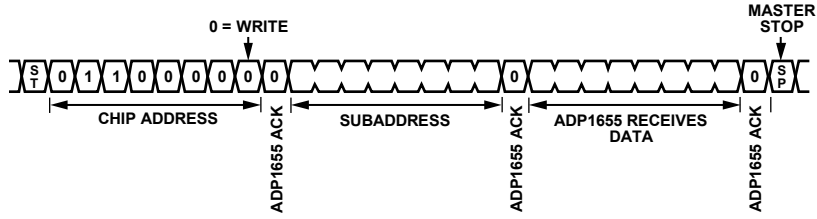
Table 8. Design Information Register (Register 0x00)

Bit	R/W	Reset State
7:0	R	00100001

Table 9. Version Register (Register 0x01)

Bit	R/W	Reset State
7:0	R	00000001

Table 10. VREF and Timer Register (Register 0x02)

Bit	R/W	Description
7:6	R/W	Reserved
5:4	R/W	Number of LEDs detection comparator reference level 00 = 4.3 V (default) 01 = 4.6 V 10 = 4.0 V 11 = 4.9 V
3:0	R/W	Flash timer value setting 0000 = 100 ms 0001 = 150 ms 0010 = 200 ms 0011 = 250 ms 0100 = 300 ms 0101 = 350 ms 0110 = 400 ms 0111 = 450 ms 1000 = 500 ms 1001 = 550 ms 1010 = 600 ms 1011 = 650 ms 1100 = 700 ms 1101 = 750 ms 1110 = 800 ms 1111 = 850 ms (default)

Table 11. Current Set Register (Register 0x03)

Bit	R/W	Description
7:4	R/W	Flash current value setting 0000 = 200 mA 0001 = 220 mA 0010 = 240 mA 0011 = 260 mA 0100 = 280 mA 0101 = 300 mA 0110 = 320 mA (default for two LEDs) 0111 = 340 mA 1000 = 360 mA 1001 = 380 mA 1010 = 400 mA 1011 = 420 mA 1100 = 440 mA 1101 = 460 mA 1110 = 480 mA 1111 = 500 mA (default for one LED)
3		N/A
2:0	R/W	Torch and assist light current value setting 000 = 20 mA 001 = 40 mA (default) 010 = 60 mA 011 = 80 mA 100 = 100 mA 101 = 120 mA 110 = 140 mA 111 = 160 mA

Table 12. Output Mode Register (Register 0x04)

Bit	R/W	Description
7:6	R/W	Inductor peak current limit setting 00 = 1.25 A 01 = 1.5 A 10 = 1.75 A (default) 11 = 2.0 A
5	R/W	0 = edge sensitive 1 = level sensitive (default)
4	R/W	0 = TORCH not allowed 1 = TORCH allowed (default)
3	R/W	0 = LED_OUT off (default) 1 = LED_OUT on
2	R/W	0 = STROBE disabled 1 = STROBE enabled (default)
1:0	R/W	Configures LED output mode 00 = standby mode (default) 01 = reserved 10 = assist light mode 11 = flash mode

Table 13. Fault Information Register (Register 0x05)

Bit	R/W	Description
7	R	0 = no fault (default) 1 = overvoltage or C_{OUT} fault
6	R	0 = no fault (default) 1 = short-circuit fault
5	R	0 = no fault (default) 1 = overtemperature fault
4	R	0 = no fault (default) 1 = timeout 850 ms fault
3	R/W	0 = one LED 1 = two LEDs (default)
2	R	Reserved
1	R	0 = no fault (default) 1 = current limit fault
0	R	Reserved

Table 14. Input Control Register (Register 0x06)

Bit	R/W	Description
7:3		Reserved
2	R/W	0 = Strobe 0 triggers flash in level sensitive mode, Strobe 1 > 0 triggers flash in edge sensitive mode 1 = Strobe 1 triggers flash in level sensitive mode, Strobe 0 > 1 triggers flash in edge sensitive mode (default)
1	R/W	0 = TX_MASK function disabled 1 = TX_MASK function allowed (default)
0	R	Reserved

SAFETY FEATURES

For critical system conditions, such as output overvoltage, flash timeout, LED output short circuit, and overtemperature conditions, the ADP1655 has built-in safety mechanisms. If one of the fault conditions occurs, the device shuts down and a corresponding flag is set in the fault information register (Register 0x05). In I²C interface mode, the system baseband processor can read the fault information register through the I²C interface to determine the nature of the fault condition and, consequently, the fault flag is cleared. The device is disabled until the fault information register is cleared.

In 2-bit logic interface mode, the I²C register readback is not available. To clear a fault, set EN1, EN2, and TORCH low.

OVERVOLTAGE FAULT

The ADP1655 contains a comparator at the VOUT pin that monitors the voltage between VOUT and SGND. If the voltage exceeds 9.5 V (typical), the ADP1655 shuts down. In I²C mode, Bit 7 in the fault information register is read back as high. The ADP1655 is disabled until the fault is cleared, ensuring protection against an open circuit.

OUTPUT CAPACITOR FAULT

If no output capacitor is present at the VOUT pin when the ADP1655 is enabled for a flash, torch, or assist light event, the part shuts down and Bit 7 in the fault information register is read back as high. The ADP1655 is disabled until the fault is cleared. The output capacitor detection scheme does not cause the VOUT pin to rise above the overvoltage threshold even though the overvoltage flag (Bit 7) in the fault information register (Register 5) is set. The overvoltage and output capacitor fault flags share a single register bit to reduce the required number of registers.

TIMEOUT FAULT

If the 2-bit logic interface is used, the maximum duration for flash being enabled (EN1/EN2 = 1) is preset to 850 ms. If EN1 and EN2 remain high for longer than 850 ms, ADP1655 is disabled until the fault is cleared (EN1, EN2, and TORCH low).

In I²C mode, if strobe mode is enabled (Register 0x04, Bit 2), strobe is set to level sensitive mode (Register 0x04, Bit 5), and if strobe remains high for longer than 850 ms, the timeout fault bit, Register 0x05, Bit 4), is read back as high. The ADP1655 is disabled until the fault is cleared.

OVERTEMPERATURE FAULT

If the junction temperature of the ADP1655 rises above 150°C, a thermal protection circuit shuts down the device. In I²C mode, Bit 5 of the fault information register is read back as high. The ADP1655 is disabled until the fault is cleared.

SHORT-CIRCUIT FAULT

The LED_OUT pin features short-circuit protection that disables the ADP1655 if it detects a short circuit to ground at the LED_OUT pin. The ADP1655 monitors the LED voltage when the LED driver is enabled. If the LED_OUT pin remains below the short-circuit detection threshold during startup, a short circuit is detected. Bit 6 of the fault information register is read back as high. The ADP1655 is disabled until the fault is cleared.

CURRENT LIMIT

The internal switch limits battery current by ensuring that the peak inductor current does not exceed the programmed limit (current limit is set by Bit 6 and Bit 7 in the output mode register, Register 0x04). If the peak inductor current exceeds the limit, the part shuts down and Bit 1 of the fault information register is read back as high. The ADP1655 is disabled until the fault is cleared.

AMOUNT OF LED DETECTION

The ADP1655 is able to detect the amount of LED connected in series between the LED_OUT pin and the PGND potential. In I²C mode, the detection is enabled with Bit 3 in the output mode register. The part uses an 80 mA LED driver current setting to detect the LED forward voltage (Vf) with a voltage comparator at the start of a flash, torch, or assist light event. If the detected forward voltage is higher than 4.3 V (typical), Bit 3 of the fault information register is read back as high.

INPUT UNDERVOLTAGE

The ADP1655 includes an input undervoltage lockout circuit. If the battery voltage drops below the 2.4 V (typical) input UVLO threshold, the ADP1655 shuts down. In this case, information in all registers is lost, and when power is reapplied, a power-on reset circuit resets the registers to their default conditions.

APPLICATIONS INFORMATION

EXTERNAL COMPONENT SELECTION

Selecting the Inductor

The ADP1655 boost converter increases the battery voltage to allow driving of one or two LEDs, whose combined voltage drop is higher than the battery voltage plus the current source headroom voltage. This allows the converter to regulate the LED current over the entire battery voltage range and with a wide variation of LED forward voltage.

The inductor saturation current should be greater than the sum of the dc input current and half the inductor ripple current. A reduction in the effective inductance due to saturation increases the inductor current ripple. Suggested inductors are shown in Table 15.

Table 15. Suggested Inductors

Vendor	Value (µH)	Part No.	DCR (mΩ)	ISAT (A)	Dimensions L × W × H (mm)
Toko	2.2	FDSE0312	160	3.1	3 × 3 × 1.2
Toko	2.0	DE2812C	67	1.8	3.0 × 3.2 × 1.22
Coilcraft	2.2	LPS3010	220	1.4	3 × 3 × 1.0
Coilcraft	2.2	LPS3314	100	1.5	3 × 3 × 1.4

Selecting the Input Capacitor

The ADP1655 requires an input bypass capacitor to supply transient currents while maintaining constant input and output voltages. The input capacitor carries the input ripple current, allowing the input power source to supply only the dc current. Use an input capacitor with a sufficient ripple current rating to handle the inductor ripple. Increased input capacitance reduces the amplitude of the switching frequency ripple on the battery. Because of the dc bias characteristics of ceramic capacitors, a 0603, 6.3 V X5R/X7R, 10 µF ceramic capacitor is preferable.

Higher value input capacitors help to reduce the input voltage ripple and improve transient response. Maximum input capacitor current is calculated using the following equation:

$$I_{CIN} \geq I_{LOAD(MAX)} \sqrt{\frac{V_{OUT}(V_{IN} - V_{OUT})}{V_{IN}}}$$

To minimize supply noise, place the input capacitor as close to the VIN pin of the ADP1655 as possible. As with the output capacitor, a low ESR capacitor is suggested. A list of suggested input capacitors is shown in Table 16.

Table 16. Suggested Input Capacitors

Vendor	Value	Part No.	Dimensions L × W × H (mm)
Murata	10 µF, 6.3 V	GRM188R60J106ME47	1.6 × 0.8 × 0.8
TDK	10 µF, 6.3 V	C1608JB0J106K	1.6 × 0.8 × 0.8
Tayio Yuden	10 µF, 6.3 V	JMK107BJ106MA	1.6 × 0.8 × 0.8

Selecting the Output Capacitor

The output capacitor maintains the output voltage and supplies the LED current during NFET power switch on period. It also stabilizes the loop. A 10.0 µF, 16 V X5R/X7R ceramic capacitor is suggested.

Note that dc bias characterization data is available from capacitor manufacturers and should be taken into account when selecting input and output capacitors. 16 V capacitors are recommended for most two-LED designs. Designs with 1 mm height restrictions can also use 0603 case size, 16 V capacitors in parallel. A list of suggested output capacitors is shown in Table 17.

Table 17. Suggested Output Capacitors

Vendor	Value	Part No.	Dimensions L × W × H (mm)
Murata	10.0 µF, 10 V	GRM21BR71A106KE51	2 × 1.25 × 1.25
Murata	10.0 µF, 16 V	GRM31CR61C106KA88	3.2 × 1.6 × 1.6
Tayio Yuden	10.0 µF, 16 V	EMK212BJ106KG	2 × 1.25 × 1.25

Higher output capacitor values reduce the output voltage ripple and improve load transient response. When choosing this value, it is also important to account for the loss of capacitance due to output voltage dc bias.

Ceramic capacitors are manufactured with a variety of dielectrics, each with different behavior over temperature and applied voltage. Capacitors must have a dielectric that ensures the minimum capacitance over the necessary temperature range and dc bias conditions. X5R or X7R dielectrics with a voltage rating of 10.0 V or 16 V are suggested for best performance. Y5V and Z5U dielectrics are not suggested for use with any dc-to-dc converter because of their poor temperature and dc bias characteristics.

The worst-case capacitance accounting for capacitor variation over temperature, component tolerance, and voltage is calculated using the following equation:

$$C_{EFF} = C_{OUT} \times (1 - TEMPCO) \times (1 - TOL)$$

where:
C_{EFF} is the effective capacitance at the operating voltage.
$TEMPCO$ is the worst-case capacitor temperature coefficient.
TOL is the worst-case component tolerance.

In this example, TEMPCO over −40°C to +85°C is assumed to be 15% for an X5R dielectric, TOL is assumed to be 10%, and C_{OUT} is 9.528 μF at 1.8 V, as shown in Figure 35.

Substituting these values in the equation yields

$$C_{EFF} = 9.528 \text{ μF} \times (1 - 0.15) \times (1 - 0.1) = 7.288 \text{ μF}$$
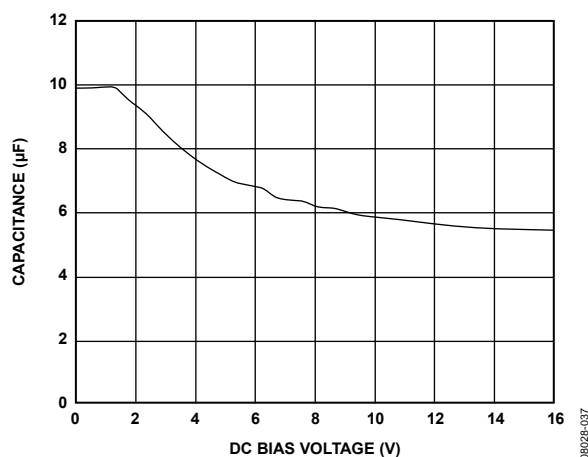

Figure 35. DC Bias Characteristic of a 16 V, 10 μF Ceramic Capacitor

To guarantee the performance of the ADP1655, it is imperative that the effects of dc bias, temperature, and tolerances on the behavior of the capacitors be evaluated for each application.

The peak-to-peak output voltage ripple for the selected output capacitor and inductor values is calculated using the following equation:

$$V_{RIPPLE} = \frac{V_{IN}}{(2\pi \times f_{SW}) \times 2 \times L \times C_{OUT}} = \frac{I_{RIPPLE}}{8 \times f_{SW} \times C_{OUT}}$$

Capacitors with lower equivalent series resistance (ESR) are preferred to guarantee low output voltage ripple, as shown in the following equation:

$$ESR_{COUT} \leq \frac{V_{RIPPLE}}{I_{RIPPLE}}$$

The effective capacitance needed for stability, which includes temperature and dc bias effects, is 4 μF.

PCB LAYOUT

Poor layout can affect performance, causing electromagnetic interference (EMI) and electromagnetic compatibility (EMC) problems, ground bounce, and voltage losses. Poor layout can also affect regulation and stability. A good layout is implemented using the following rules and shown in Figure 36:

- Place the inductor, input capacitor, and output capacitor close to the IC using short tracks. These components carry high switching frequencies and large tracks act as antennas.

- Route the output voltage path away from the inductor and SW node to minimize noise and magnetic interference.
- Maximize the size of ground metal on the component side to help with thermal dissipation.
- Use a ground plane with several vias connecting to the component side ground to further reduce noise interference on sensitive circuit nodes.
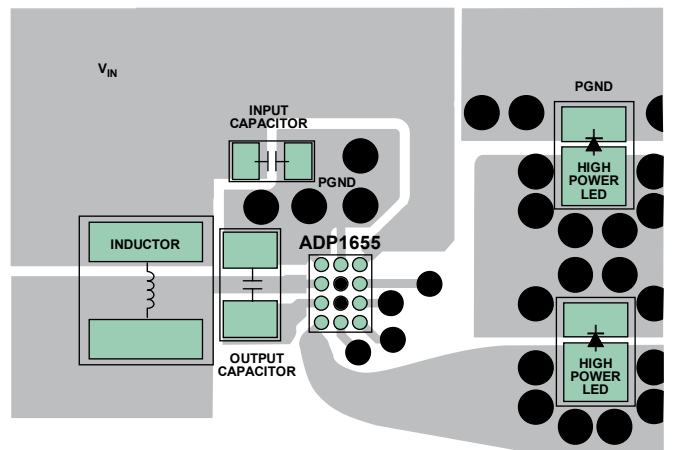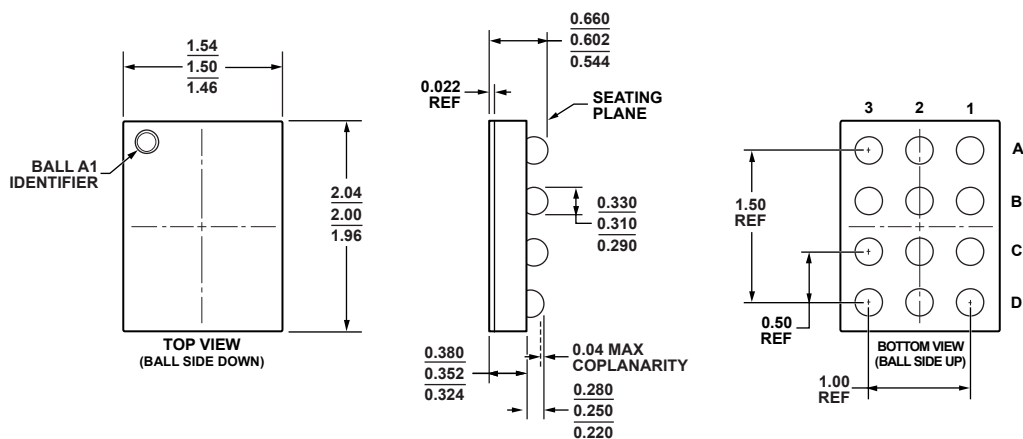

Figure 36. Example Layout of the ADP1655 Driving Two White LEDs

OUTLINE DIMENSIONS

Figure 37. 12-Ball Wafer Level Chip Scale Package [WLCSP]
(CB-12-4)
Dimensions shown in millimeters

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option	Branding
ADP1655ACBZ-R7[1]	−40°C to +125°C	12-Ball Wafer Level Chip Scale Package [WLCSP]	CB-12-4	LAM
ADP1655-EVALZ[1]		Evaluation Board		

[1] Z = RoHS Compliant Part.

NOTES

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